Design of a Low-Cost, Lightweight, Passively Cooled, Narrowband, SWIR Camera for Space-Based Imaging

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Prepared by
R. J. RUDY, Y. DOTAN, J. H. HECHT,
D. J. MABRY, and M. G. SIVJEE
Space Science Applications Laboratory
Laboratory Operations

D. W. WARREN
Sensor Systems Subdivision
Electronic Systems Division

Prepared for
SPACE AND MISSILE SYSTEMS CENTER
AIR FORCE SPACE COMMAND
2430 E. El Segundo Boulevard
Los Angeles Air Force Base, CA 90245

Engineering and Technology Group

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Michael Zambrana
SMC/AXE
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The Aerospace Corporation
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El Segundo, CA 90245-4691

**Space and Missile Systems Center**
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**This document describes the conceptual design for a spaceborne imager operating in the short-wave infrared (SWIR). The instrument was designed to image a single emission line of OH from the night sky from a satellite in low Earth orbit, but much of the design is relevant to other imaging applications. The imager has a 23.4° FOV in the cross-track direction with a resolution of 1.6 mrad. It functions like a line scanner but incorporates a two-dimensional (256 x 256) InGaAs focal-plane array (FPA) to attain high signal-to-noise. It employs a set of telecentric foreoptics to accommodate narrowband imaging, but a simpler design would serve for broadband measurements. It uses a small radiator to passively cool the FPA to < 200K and direct radiation by the support assembly to reduce the temperature of the optics. A novel approach that accomplishes TDI (time, delay, integrate) off of the focal plane obviates the need for a step/stare mirror. The design has no moving parts, weighs less than 6 kg, and consumes less than 15 W of power.**

**Space instrumentation, Infrared imaging, Infrared radiometry**
Acknowledgments

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1. Background

The history of this project is very brief. Approximately 6 weeks prior to the proposal deadline, Aerospace was approached about providing an instrument for a NASA Small Explorer (SMEX) class mission. The instrument needed to image the nighttime mesospheric hydroxyl (OH) emission at 1.4 μm from a 550-km orbit. The instrument was severely weight limited (<6 kg) and had tight restrictions on both power (15 W) and cost ($2.6 M). It needed to provide a reasonable field of view (FOV) at good spatial resolution (<2 km) and high signal-to-noise (S/N ≥ 100) with significant processing capability to keep the data flow manageable. In addition to observing the OH emission in a single bright emission line, the system also needed to sample the background emission in a second nearby band that was free of OH. The design that resulted from these requirements is described below.
2. Selection of the Focal-Plane Array

The choice of the focal-plane array (FPA) was the first, and in some ways most fundamental, decision to be made. Many aspects of the optical design flowed from the linear dimensions and pixel pitch of the FPA. The stringent performance dictated by the mission requirements limit the commercially available FPAs to only a few units. The 1.4 \( \mu \)m operating wavelength precluded the use of intrinsic silicon (CCDs or CMOS) detectors. InSb was ruled out because its very long cut-off wavelength (>5 \( \mu \)m) leads to overwhelming large dark currents and background fluxes at the operating temperatures envisioned.

The two detector materials that remained as possible candidates were HgCdTe and InGaAs. Both are trinary compounds that can have the relative proportions of the three materials varied to cover a broad range of wavelengths. Typically, there is a short-wave cutoff at wavelengths < 1 \( \mu \)m due to absorption in the material. The long-wave cutoff is selected by choosing the proper material proportions. Photons below a certain energy (and thus beyond a certain wavelength) will not promote charges into a conduction band where they can be detected. The general rule of thumb is to tailor the bandgap such that the cutoff wavelength is just to the red of the longest wavelength for which a detection is required. This is extremely important in situations where cooling capacity is limited since by maximizing the bandgap the thermally generated charges in the detector material (the so-called dark current) are minimized. Moreover, a larger bandgap and a shorter cutoff wavelength also minimize the detection of unwanted thermal background from the detector housing, optics, filters, etc. Thus for this program with measurement requirements at 1.4 \( \mu \)m, a cutoff wavelength of ~ 1.5 \( \mu \)m would seem ideal. In actuality, neither HgCdTe nor InGaAs exist in a broad assortment of cutoff wavelengths. Other than a few “standard” cutoffs, other wavelengths require a development effort that is beyond the budget of this program. For that reason, we considered both materials with cutoff wavelengths of 1.7 \( \mu \)m, a value for which the vendors either had assets or a good deal of experience in growing reliable detector material. At Rockwell’s recommendation, we selected InGaAs as the detector material since it exhibits smaller dark currents over the range of temperatures that the FPA is likely to operate with comparable or better quantum efficiency than HgCdTe. Table 1 presents the dark current as a function of temperature as derived from some test data provided by Rockwell.

<table>
<thead>
<tr>
<th>Temperature (K)</th>
<th>Dark Current (electrons sec(^{-1}) @ 50 mV bias)</th>
</tr>
</thead>
<tbody>
<tr>
<td>170</td>
<td>0.003</td>
</tr>
<tr>
<td>180</td>
<td>0.1</td>
</tr>
<tr>
<td>190</td>
<td>1.6</td>
</tr>
<tr>
<td>200</td>
<td>16</td>
</tr>
<tr>
<td>210</td>
<td>160</td>
</tr>
<tr>
<td>220</td>
<td>1280</td>
</tr>
</tbody>
</table>
After deciding on InGaAs as the detector material, the single remaining decision was the selection of the multiplexor (MUX). There were two, again supplied by the Rockwell Science Center, that met the low read noise criterion necessary to satisfy the signal-to-noise requirement. These were the Hawaii MUX with 18.5 µm pixels in a 1024 x 1024 format and the PICNIC MUX with 40 µm pixels in a 256 x 256 arrangement. While the former would have produced a more capable instrument, problems with reading the array sufficiently rapidly to keep pace with the satellite motion and the processing required to handle the large number of pixels, plus a significantly greater cost, led us to choose the PICNIC MUX instead. The operation of this resulting InGaAs array on the 256 x 256 PICNIC MUX is discussed further in the electronics section.
3. Optical Design

Once the FPA had been selected the remaining factors that drove the optical design were field of view (FOV), footprint size, and signal-to-noise (S/N). An additional consideration was the necessity to observe with a second background filter in addition to a narrowband filter that isolates a strong OH emission line.

The FOV and footprint size are related by the fact that, for utilization of the full array, the FOV is 256 times the footprint. For this reason the two were considered together. In choosing the FOV, it was desirable to have sufficient spatial coverage in the cross-track (the direction perpendicular to the motion of the spacecraft) to resolve the largest of the atmospheric wave structures that the mission targeted. Coverage in this direction of 23.4° provided 190 km at the level of the OH emission, sufficient to fulfill this goal. This corresponded to a footprint of 0.74 km, well under the 2-km resolution necessary to investigate wave activity on the smallest scales.

Before considering the S/N requirement, a topic that will also be discussed in the electronics section, we consider the requirement for narrowband imaging and how this affects the optical design. The issue is that the effective passband of a filter changes as the angle of the incident light varies. The shift in wavelength, which is always towards shorter wavelengths, can, for small angles, be expressed as:

$$\Delta \lambda = \lambda \theta^2 / 2 n^2,$$

where $\theta$ is the angle from normal, and $n$ is the index of refraction of the substrate.

Table 2 presents shifts for a range of incidence angles for a bandpass centered at 1.4344 $\mu$m, the location of the OH line of interest. The values were computed for an index of refraction of 1.57, a

<table>
<thead>
<tr>
<th>Angle of Incidence (°)</th>
<th>Wavelength Shift (Å)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0.0</td>
<td>0.0</td>
</tr>
<tr>
<td>1.0</td>
<td>0.87</td>
</tr>
<tr>
<td>2.0</td>
<td>3.48</td>
</tr>
<tr>
<td>2.86</td>
<td>7.11</td>
</tr>
<tr>
<td>4.0</td>
<td>13.9</td>
</tr>
<tr>
<td>8.0</td>
<td>55.7</td>
</tr>
<tr>
<td>10.0</td>
<td>87.0</td>
</tr>
<tr>
<td>11.7</td>
<td>119</td>
</tr>
</tbody>
</table>
value representative of several common glasses. Since the anticipated width of the bandpass is only \( \sim 13\text{Å} \), an angle of incidence greater than 4° will shift the bandpass off of the emission line. Since the 23.4° FOV implies a maximum, off-normal angle of \( 23.4°/2 = 11.7° \), it is apparent from Table 2 that the optical design could not simply place the filter in front of the optics. In fact, the situation is considerably more restrictive than this. Because it is necessary to compare very precisely different regions of the FOV with each other, each region of the FOV needs to be transmitted by the filter in precisely the same manner.

The solution to both these problems is to form a telecentric image of the field for passage through the filter. Telecentric images have the property that light bundles from all parts of the fields are parallel. This means that every portion of the field is affected by the filter in the same way. The problem relating to the angle of incidence is addressed by expanding the size of the telecentric image. This has the corresponding effect of making the spread of the light bundles (the f-cones) small. For f/10 beams, the largest off-normal angle is 2.86°. From Table 2 it can be seen that this produces a wavelength shift of only a little over 7Å, which is acceptable for a 13Å bandpass.

Figure 1 shows the optical design. The first three lenses (to the left side) provide the light-gathering aperture necessary to meet the signal-to-noise requirements as well as encompassing the large FOV. They also form the telecentric image of the FOV. The filters (not shown), which are co-planar and butted together, are positioned very near the telecentric focus. As noted above, by producing a comparatively large image at the telecentric focus, the f-cones for each part of the field are kept narrow.

The remaining five lenses (to the right of the telecentric focus) de-magnify and re-image the field on the FPA. The reduction in image size is considerable, taking the field that is approximately 2 in. on a side and re-imaging it on a 1-cm-square detector. The FOV is split evenly between the two filters, with each covering the full width of the track (i.e., a given point in the OH layer is imaged sequentially by the two filters as the spacecraft moves the location through the FOV). Figure 2 shows the spot sizes produced by this optical design. As can be seen, for both wavelengths and for all portions of the field, the spot sizes are much smaller than a single pixel.

![Figure 1. Optical layout SWIR imager.](image_url)
Figure 2. Spot diagrams for the entire optical train. The boxes show the size of a 40-μm pixel. The blue shows the spots at the wavelength of the OH line (1.4344 μm); the red shows some small additional spread at the wavelength of the background filter (1.4503 μm).

The optics that produced these very small spots are all composed of a single glass—LAH58. It is a very dense glass and was selected to see if the optical design could be implemented in a heavy material and still meet the weight requirements. If the material needed to be changed there is now every assurance that we remain within the limit.
4. Mechanical Design

The mechanical design had to accommodate the optical design by allowing for the precise positioning and mounting of the optics while housing and cooling the detector. Moreover, it had to accomplish this while meeting the stringent weight allotment. Figure 3 shows the mechanical design. The all-refractive optics make possible a very simple design in which all optical elements are mounted along a
single axis in a tube-like structure. To facilitate construction, the structure has three separate sections. The first contains the lenses that form the intermediate telecentric image and the filters. The second supports the five lenses that re-image the field on the detector. The final portion houses the detector, its mount, and the attendant cabling and provides for the thermal link to the radiator. Wherever possible, the optics are baffled to limit stray light. The tube structure has a wall thickness of 60 mils.

Weight restrictions dictated that the tube structure be aluminum, beryllium, or composite. Cost narrowed this to aluminum. Table 3 lists the design weights. Note that the total is < 6 kg.

As noted in the FPA section and evident in Table 1, the FPA must be cooled below ambient temperature to keep the dark current tractable. Weight, cost, and power restrictions preclude the use of a cryocooler. Temperatures near the desired range (≤200K) are just barely obtainable with a two-stage thermoelectric cooler and this remains a viable alternative. We have selected radiative (passive) cooling as our baseline, however. Assuming that a radiator is shielded from both the sun and the Earth and has an unobstructed view of deep space, it is warmed only by zodiacal light and diffuse galactic emission, and its temperature may fall as low as ~30K. For objects in low Earth orbit, however, the principal limitation on radiator temperature becomes heat loading from the spacecraft, both conductive and radiative. If the radiator is attached to the spacecraft by a material of low thermal conductivity, radiation from the spacecraft becomes the dominant heating mechanism. Table 4 presents the predicted focal-plane temperatures as a function of the radiative coupling between the radiator and the spacecraft. It assumes a radiator with an area of 50 in.², an effective emissivity of 0.8 radiating to space (this low value allows for some degradation over the life of the mission), a spacecraft temperature of 300K, <1 mW of heat dissipation by the FPA, 46 mW of radiative loading on the FPA and cooling link, and mounting of the radiator to the spacecraft with a 1-in.-dia tube of G10 with a 50-mil wall thickness and a 1-in. length. Conduction through the latter is second to the radiative load from the spacecraft in heating the radiator, but in all cases contributes less than 30% of the load.

<table>
<thead>
<tr>
<th>Component</th>
<th>lbs.</th>
<th>kg</th>
</tr>
</thead>
<tbody>
<tr>
<td>Optics (no filter)</td>
<td>1.136</td>
<td>0.515</td>
</tr>
<tr>
<td>Filter (5mm thick)</td>
<td>0.134</td>
<td>0.060</td>
</tr>
<tr>
<td>Electronics box</td>
<td>5.280</td>
<td>2.395</td>
</tr>
<tr>
<td>cards</td>
<td>2.700</td>
<td>1.225</td>
</tr>
<tr>
<td>enclosure</td>
<td>2.580</td>
<td>1.170</td>
</tr>
<tr>
<td>connectors</td>
<td>0.1</td>
<td>0.045</td>
</tr>
<tr>
<td>cable (conn./shielding)</td>
<td>0.15</td>
<td>0.068</td>
</tr>
<tr>
<td>Detector mount assembly</td>
<td>0.260</td>
<td>0.118</td>
</tr>
<tr>
<td>Radiator assembly</td>
<td>0.250</td>
<td>0.113</td>
</tr>
<tr>
<td>Aperture Door Assy</td>
<td>0.200</td>
<td>0.091</td>
</tr>
<tr>
<td>Structure and mounts</td>
<td>4.256</td>
<td>1.930</td>
</tr>
<tr>
<td><strong>Total</strong></td>
<td>11.75</td>
<td>5.34</td>
</tr>
</tbody>
</table>

10
Table 4. FPA Temperature as a Function of Radiator Emissivity. This is the emissivity of the side facing the spacecraft. It determines the radiative coupling between the spacecraft and the radiator, which is the principal heat load on the radiator.

<table>
<thead>
<tr>
<th>Emissivity</th>
<th>FPA Temperature (K)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0.05</td>
<td>161.9</td>
</tr>
<tr>
<td>0.075</td>
<td>172.9</td>
</tr>
<tr>
<td>0.10</td>
<td>181.7</td>
</tr>
<tr>
<td>0.15</td>
<td>195.3</td>
</tr>
<tr>
<td>0.20</td>
<td>205.6</td>
</tr>
</tbody>
</table>

While a value of < 0.05 for the emissivity should be readily obtainable with a gold overcoating to the aluminum plate radiator, a radiative barrier composed of a highly reflective plate will be inserted between the radiator and the spacecraft. This serves the same purpose as multi-layer-insulation (MLI) and will be helpful in achieving the very low effective emissivity. To reduce any thermally generated background from the optics themselves, the temperature of the entire assembly is lowered to ~250K using passive cooling. For this reason the instrument is mounted in a shaded region on the outside wall of the spacecraft.
5. Electronics

The electronics generate the clocks and bias voltages that operate the FPA; and amplify, digitize, and record the analog signals that are produced by it. Moreover, the electronics provide for an effective, off-chip TDI (time, delay, and integrate) that is necessary to achieve the requisite signal-to-noise. The implementation of all of these functions is accomplished via four separate, custom circuit cards that are shown schematically in Figure 4.

Cards 1 and 3 provide the interface with the FPA—card 3 creates the timing signals and voltages necessary to operate and sequence the array while card 1 accepts the output signal and performs the amplification before passing the signals to the single analog-to-digital converter (ADC). The decision to multiplex the four FPA outputs to a single ADC was necessary because of the large power consumption of the ADC. Card number 4 conditions and isolates the instrument power from the spacecraft power while card 2 communicates the measurements to the spacecraft for telemetry to Earth.

```
1 Analog Preprocessor 8x6" 350 5.5 W 0.7 W
2 Processor/Controller 8x6" 350 3.6 W
3 FPA Driver 8x6" 350 3.7 W 0.35 W
4 Dgtl Pwr Sply 8x6" 350 4.65 W
Total Power 9.2+4.65 = 13.85 W 9.2 W 4.65 W
```

Figure 4. Schematic of the electrical design.
after first performing the crucial TDI processing. The latter removes the need for expensive "step and stare" mechanisms in front of the optics and is described in some detail in the following paragraph.

True TDI is associated with CCDs (charge coupled devices) and results from the ability to transfer the photo-generated charge from one pixel to another. The rate at which the charge is transferred can be chosen to offset image motion, thereby allowing the integration time on an area of interest in the FOV to be as long as the time it takes that area to transit the array. Because the PICNIC array (like most infrared arrays) has separate circuitry for each pixel that is not linked to the other pixels, TDI as implemented on CCDs is not possible. What is possible, and what is intended for this instrument, is to read the array at the frame rate that matches the motion of the image from one row of the array to the next and to coadd these frames in such a manner as to keep track of the image motion and preserve the resolution of a single pixel. Since this will involve the coaddition of from 64 to 96 frames, the "trick" is to minimize the read noise. To do so, we make use of the fact that the read noise is associated almost entirely with the reset process (the process by which a pixel is switched to ground to clear the charge that has accumulated) and not with the actual sampling of the array. For this reason, it is our intent to reset the array only a single time during the 64 to 96 frames that constitute a measurement cycle. This is, of course, only possible if the photo-charge does not exceed the integration capacity of a pixel. The light levels anticipated fill only a quarter to a third of this capacity so the observing conditions are comfortably within that limit. Without the reset, however, the photo-charge continues to accrue. Thus the signal measured for a given pixel during a given frame must have its value from the prior frame subtracted from it.
6. Summary

This document has presented the conceptual design for a very lightweight (<6 kg), low-power (<15 W), passively cooled, low-cost (~ $2.5M), spaceborne SWIR imager. The instrument was designed for a NASA proposal to image a single emission line of OH from the night sky from a satellite in low Earth orbit, but much of the design has broad applicability. The imager has a 23.4° FOV in the cross-track direction with a resolution of 1.6 mrad. The heart of the imager is a modern 256 x 256 InGaAs FPA. It is operated in a “pushbroom” fashion, but instead of employing a single line of detectors, utilizes most of its rows to enhance signal-to-noise. A sophisticated set of electronics performs the necessary shifting, sorting, and coadding necessary to preserve the spatial resolution of a single pixel.
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