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**Strong-Inversion Model of Threshold Voltage in  
Modulation Doped Field-Effect Transistors:  
The Role of Unintentional Acceptors**

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A strong-inversion model, in the depletion layer approximation, of the threshold voltage for modulation doped field-effect transistors (MODFETs) has been developed. We show that for a range of unintentional acceptor doping densities from $10^{13}$ to $10^{15}$ cm <sup>-3</sup> in typical molecular beam epitaxy (MBE)-grown structures, threshold voltages may increase by as much as 200 mV. We isolate the contributions of AlGaAs and GaAs to the threshold voltage. We have verified our model by comparing predicted threshold voltages with those measured in typical MBE-grown MODFET structures.					
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PREFACE

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## I. INTRODUCTION

Since the original charge-control model for the I-V characteristics of modulation doped field-effect transistors (MODFETs) was developed,<sup>1</sup> a number of models have been proposed.<sup>2-11</sup> In most of these models the density of unintentional acceptor doping, which results as a by-product of the molecular beam epitaxy (MBE) growth process, is ignored when one calculates the electrical characteristics of the device. None of the models includes a precise description of the threshold voltage in MODFET structures; rather, the models all cite  $V_{\text{off}}$ ,<sup>1</sup> which is usually considered to be independent of unintentional acceptor doping, as an estimate of the threshold voltage.

In the discussion that follows, we show (1) that the threshold voltage may be defined in the strong-inversion, depletion layer approximation, and (2) that the density of unintentional acceptor doping plays an important role in determining this voltage.

## II. STRONG-INVERSION MODEL

The band structure of a typical AlGaAs(n)/GaAs heterojunction with a Schottky barrier at the gate and a spacer layer at the interface under bias is shown in Fig. 1. In most charge-control models, the AlGaAs layer is assumed to be totally depleted during operation<sup>1-3,7,9,10,11</sup> and unintentional acceptor doping in both the AlGaAs and the GaAs layer is ignored. Even in the more complicated descriptions, in which the incomplete ionization of the AlGaAs is included, unintentional acceptor doping is usually ignored altogether<sup>4,5,7,8</sup> or is included only in the GaAs layer.<sup>6</sup> First, we will concern ourselves with the net doping density in each of the regions of the band structure shown in Fig. 1.

### A. CHARGE DENSITIES AT THRESHOLD

In the following discussion we assume that all the unintentional acceptors  $N_a$  and all the AlGaAs donors are ionized. In the doped AlGaAs layer the charge density is given by

$$\rho_d = q(N_d - N_a) \quad (1)$$

where  $q$  is the elemental charge,  $N_d$  is the donor concentration in the doped AlGaAs layer, and  $N_a$  is the unintentional acceptor doping density throughout the structure.

In the spacer layer  $a$ , the charge density is due to the unintentional acceptor doping density; i.e.,

$$\rho_a = -qN_a \quad (2)$$

The region from the interface ( $d + a$ ) to the edge of the depletion layer ( $W + d + a$ ) has the following charge density:

$$\rho_w(z) = -q[N_a + n(z)] \quad (3)$$

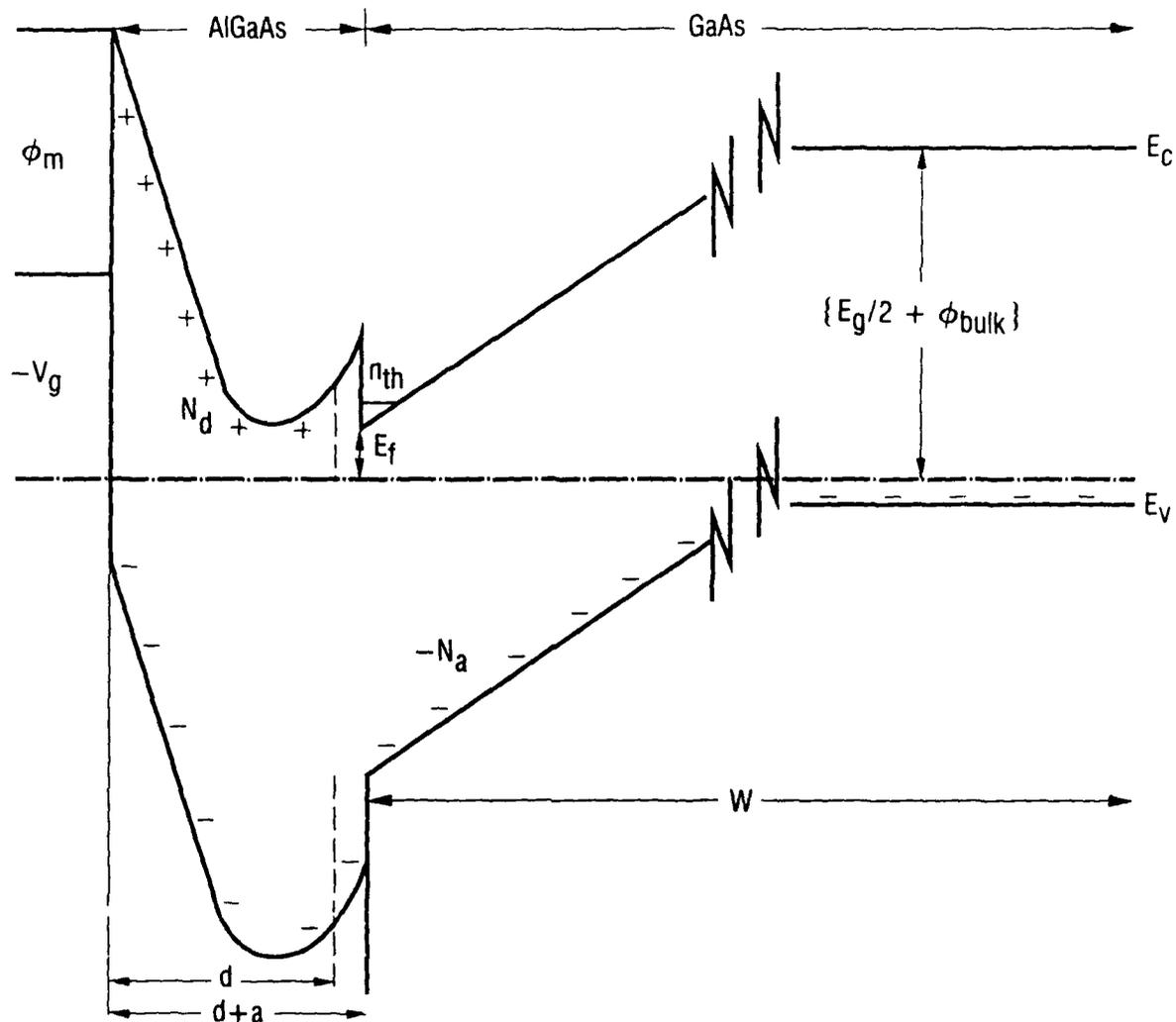


Fig. 1. The Band Diagram of a Typical AlGaAs/GaAs MODFET with Schottky Gate, under Bias  $V_g$ , at Threshold. At threshold the AlGaAs donors  $N_d$  and unintentional acceptors  $N_a$  are assumed to be completely ionized. The charge density  $n_a$  in the channel at threshold,  $n_{th}$ , is equal to  $N_a z_{ave}$  in our strong-inversion model. The Fermi level  $E_f$  lies below the bottom of the two-dimensional well at the interface. Band bending from the interface to the edge of the depletion region is the difference between the potential of the conduction band in the bulk,  $E_g/2 + \phi_{bulk}$ , and the potential at the interface,  $E_f$ .

where  $n(z)$  is the channel carrier density.

We assume that the channel carrier density at threshold may be written as

$$n(z) = n_{th} \delta[z - (d + a + z_{ave})] \quad (4)$$

where  $n_{th}$  is the channel carrier density at threshold and  $z_{ave}$  is the average position of the channel charge in the two-dimensional (2-D) channel. At threshold we require that the channel carrier density be

$$n_{th} = N_a z_{ave} \quad (5)$$

Therefore, the charge density in the GaAs layer is

$$\rho_w(z) = -qN_a \{1 + z_{ave} \delta[z - (d + a + z_{ave})]\} \quad (6)$$

Equation (5), for the channel carrier density at threshold, is consistent with the strong-inversion definition of threshold in MOSFETs, where strong inversion occurs when the density of electrons in the channel is equal to the density of acceptors in the bulk. Our definition of strong inversion in MODFETs is the 2-D analog of strong inversion in MOSFETs.

#### B. THRESHOLD VOLTAGE

The potential  $V$  across the structure may be derived in a convenient form by solving Poisson's equation in the following way:

$$V(z) = V_a(z) + V_b(z) \quad (7)$$

where  $V_a(z)$  is the solution to the homogeneous equation and  $V_b(z)$  is the solution to the inhomogeneous equation with homogeneous boundary conditions. The homogeneous equation yields

$$V_a(z) = -(V_1 - V_2)(z - z_1)/(z_2 - z_1) + V_1 \quad (8)$$

where  $V_1$  is the potential at  $z_1$  and  $V_2$  is the potential at  $z_2$ . The inhomogeneous equation may be solved to yield

$$V_b(z) = - [1/\epsilon(z_2 - z_1)] \left[ (z - z_2) \int_{z_1}^z (z' - z_1) \rho(z') dz' + (z - z_1) \int_z^{z_2} (z' - z_2) \rho(z') dz' \right] \quad (9)$$

where  $\epsilon$  is the permittivity. Therefore, the potential at any point  $z$  in the structure may be written as

$$V(z) = V_1 - (V_1 - V_2)[z - z_1]/(z_2 - z_1) - [1/\epsilon(z_2 - z_1)] \left[ (z - z_2) \int_{z_1}^z (z' - z_1) \rho(z') dz' + (z - z_1) \int_z^{z_2} (z' - z_2) \rho(z') dz' \right] \quad (10)$$

We can obtain the field  $E$  at any position  $z$  by differentiating Eq. (10):

$$E(z) = (V_1 - V_2)/(z_2 - z_1) + [1/\epsilon(z_2 - z_1)] \left[ \int_{z_1}^z (z' - z_1) \rho(z') dz' + \int_z^{z_2} (z' - z_2) \rho(z') dz' \right] \quad (11)$$

Evaluating Eq. (11) from the gate to the interface and solving for the potential at the gate yields

$$V_G = z_i E_i + V_{i-} - (1/\epsilon) \int_0^{z_i} \rho(z') z' dz' \quad (12)$$

where  $V_G$  is the potential at the gate and  $E_i$  is the field at the interface. Evaluating Eq. (11) from the interface ( $d + a$ ) to the edge of the depletion layer ( $W + d + a$ ) and solving for the potential at the edge of the depletion layer yields

$$V_W = -WE_i + V_{i+} + (1/\epsilon) \int_{z_i}^{W+z_i} (z' - z_2)\rho(z')dz' \quad (13)$$

where  $V_W$  is the potential at the edge of the depletion layer relative to the Fermi level. Subtracting Eq. (13) from Eq. (12) and solving for the potential at the gate yields

$$V_G = (W + z_i)E_i + (V_{i-} - V_{i+}) + V_W - (1/\epsilon) \left\{ \int_0^{z_i} \rho(z')z'dz' + \int_z^{W+z_i} [z' - (W + z_i)]\rho(z')dz' \right\} \quad (14)$$

which upon evaluation of the last term may be simplified to yield

$$V_G = (V_{i-} - V_{i+}) + V_W - (1/\epsilon) \int_0^{W+z_i} \rho(z')z'dz' \quad (15)$$

where  $(V_{i+} - V_{i-})$  is the conduction band offset, and  $z_i$  is equal to the sum of the doped AlGaAs layer and the spacer layer, ( $d + a$ ). The potential at the gate  $V_G$  is the difference between the applied potential  $V_g$  and the Schottky-barrier height at the gate  $\phi_m$ . Making these substitutions and solving for the applied gate voltage at threshold  $V_{th}$  yields

$$V_{th} = \phi_m = \Delta E_c + V_W - (1/\epsilon) \left[ \int_0^d \rho_d z dz + \int_d^{d+a} \rho_a z dz + \int_{d+a}^{W+d+a} \rho_W(z) z dz \right] \quad (16)$$

where  $\Delta E_c$  is the conduction band offset at the AlGaAs/GaAs interface.

In this analysis the dielectric constants of GaAs and AlGaAs are taken to be identical. The potential at the depletion layer edge  $V_W$  is the negative of the sum of the potential at the bottom of the conduction band and the band bending from the interface to the depletion layer edge at strong inversion,  $\phi_{bb}$ ; i.e.,

$$V_W = -(E_f + \phi_{bb}) \quad (17)$$

In the depletion layer approximation  $\phi_{bb}$  may be calculated by integrating over the charge density from the depletion layer edge to the interface:

$$\phi_{bb} = -(1/\epsilon) \int_{W+d+a}^{d+a} \int_{W+d+a}^{z'} \rho_W(z) dz dz' \quad (18)$$

Using Eq. (6) for the charge density in the GaAs region, the band bending from the interface to the edge of the depletion layer is

$$\phi_{bb} = (q/\epsilon)(N_a W^2/2)(1 - 2z_{ave}^2/W^2) \quad (19)$$

Effecting the integrations in Eq. (16), using Eqs. (1), (2), (6), and (19), yields

$$V_{th} = \phi_m - \Delta E_c - E_f + (q/\epsilon) \left\{ -(N_d - N_a)d^2/2 + N_a W(d+a) [1 + \mathcal{O}(z_{ave}/W)] \right\} \quad (20)$$

for the threshold voltage where terms of order  $z_{ave}/W$  will be neglected.

In the typical triangular-well approximation with one subband, the difference between the Fermi energy and the bottom of the 2-D well, at threshold, can be shown to be<sup>1</sup>

$$E_f = - (kT/q) \ln[\exp(\pi n_{th}^2/m_1 kT) - 1] - E_0 \quad (21)$$

where  $k$  is the Boltzmann constant,  $T$  is absolute temperature,  $h$  is the Planck constant divided by  $2\pi$ ,  $m_1$  is the longitudinal effective mass of the carriers, and  $E_0$  is the energy level of the subband.

At threshold the exponential may be expanded to yield

$$E_f = -(kT/q) \ln(\pi n^2 n_{th} / m_1 kT) - E_0 \quad (22)$$

The triangular-well approximation yields the following for the energy of the first subband:

$$E_0 = 9n^2/8m_1 (4E_1 m_1 / n^2)^{2/3} \quad (23)$$

Integrating the charge density from the interface to the edge of the depletion layer yields the field at the interface:

$$E_i = (q/\epsilon) \{N_a W [1 + O(z_{ave}/W)]\} \quad (24)$$

where the terms of order  $z_{ave}/W$  will be ignored. The triangular-well approximation gives the following for the average channel width:

$$z_{ave} = 3/(4E_1 m_1 / n^2)^{1/3} \quad (25)$$

The depletion width may be determined from Eq. (19):

$$W = [2(\epsilon/q)\phi_{bb}/N_a]^{1/2} \quad (26)$$

The band bending  $\phi_{bb}$  may be written in terms of the GaAs band gap, the bulk potential  $\phi_{bulk}$ , and the Fermi potential (see Fig. 1):

$$\phi_{bb} = E_g/2 + \phi_{bulk} - E_f \quad (27)$$

where  $E_g$  is the band gap of GaAs and  $\phi_{\text{bulk}}$  is the bulk potential in the GaAs. The bulk potential can be calculated by invoking charge neutrality in the bulk. For a p-type material of acceptor density  $N_a$ , this yields

$$\phi_{\text{bulk}} = (kT/q)\ln(N_a/n_i) \quad (28)$$

where  $n_i$  is the intrinsic carrier density of GaAs ( $\approx 1.79 \times 10^6 \text{ cm}^{-3}$  at room temperature).

Equation (27) for band bending in MODFETs differs from that in MOSFETs in which the band bending is twice the bulk potential. This difference is due to the 2-D nature of the MODFET structure. To compare the band bending in these two structures, we can write Eq. (27) in the following form:

$$\phi_{\text{bb}} = 2\phi_{\text{bulk}} [1 + (E_g/2 - \phi_{\text{bulk}} - E_f)/2\phi_{\text{bulk}}] \quad (29)$$

For high acceptor concentrations ( $\approx 10^{17} \text{ cm}^{-3}$ ) the MODFET band bending exceeds the MOSFET band bending by less than 5%, whereas at low acceptor densities ( $\approx 10^{13} \text{ cm}^{-3}$ ) the MODFET band bending is almost 40% greater. This difference is reflected in the depletion layer width  $W$ . At high acceptor concentrations the MODFET depletion layer is only about 2% larger than the MOSFET depletion layer for the same doping densities. However, at low acceptor concentrations the width of the MODFET depletion layer exceeds that of the MOSFET depletion layer by  $\approx 18\%$ .

The substitution of Eqs. (22) through (28) into Eq. (20) yields a threshold voltage that is a function of the geometric parameters  $d$  and  $a$ , the donor density  $N_d$ , and the unintentional acceptor doping density  $N_a$ .

### III. THRESHOLD VOLTAGE VERSUS ACCEPTOR DENSITY

In Fig. 2 we show the threshold voltage versus the unintentional acceptor concentration  $N_a$  for an AlGaAs doping density  $N_d$  of  $1 \times 10^{18} \text{ cm}^{-3}$  for the structure shown in Fig. 1. The doped AlGaAs layer  $d$  is 400 Å wide, the spacer layer  $a$  is 30 Å wide, and the GaAs layer is assumed to be semi-infinite. We have isolated the relevant terms in Eq. (20) in order to compare the contributions of each.

At low acceptor densities the threshold voltage is dominated by a combination of (1) the constant term due to the difference between the Schottky-barrier height and the conduction band offset ( $\phi_m - \Delta E_c$ ); (2) the contribution of channel charge to the Fermi energy,  $(kT/q) \ln(\pi^2 n_{th}^2 / m_1 kT)$ ; and (3) the contribution of doped AlGaAs,  $-(N_d - N_a)d^2/2$ . The contribution of the first quantum-level energy,  $E_0$ , and the contribution proportional to the depletion layer width,  $N_a W(d + a)$ , are near zero at acceptor densities below  $\approx 10^{14} \text{ cm}^{-3}$ .

As the acceptor density increases, all nonconstant contributions to the threshold voltage increase. Even though the depletion layer width  $W$  decreases with increasing acceptor concentration, the depletion layer charge  $-qN_a W$  increases, thereby increasing the contribution of the depletion layer to the threshold voltage. As the field at the interface increases as a result of increased charge in the depletion layer, the energy level in the 2-D well also increases as the two-thirds power of the depletion-layer charge, thus also contributing to an increase in the threshold voltage.

Because the channel charge at threshold increases with the acceptor density  $N_a z_{ave}$ , this contribution also tends to increase the threshold voltage. For all but the highest acceptor concentrations, the contribution of the AlGaAs layer to the threshold voltage is relatively constant; only at acceptor concentrations comparable to the donor density does this term begin to increase appreciably.

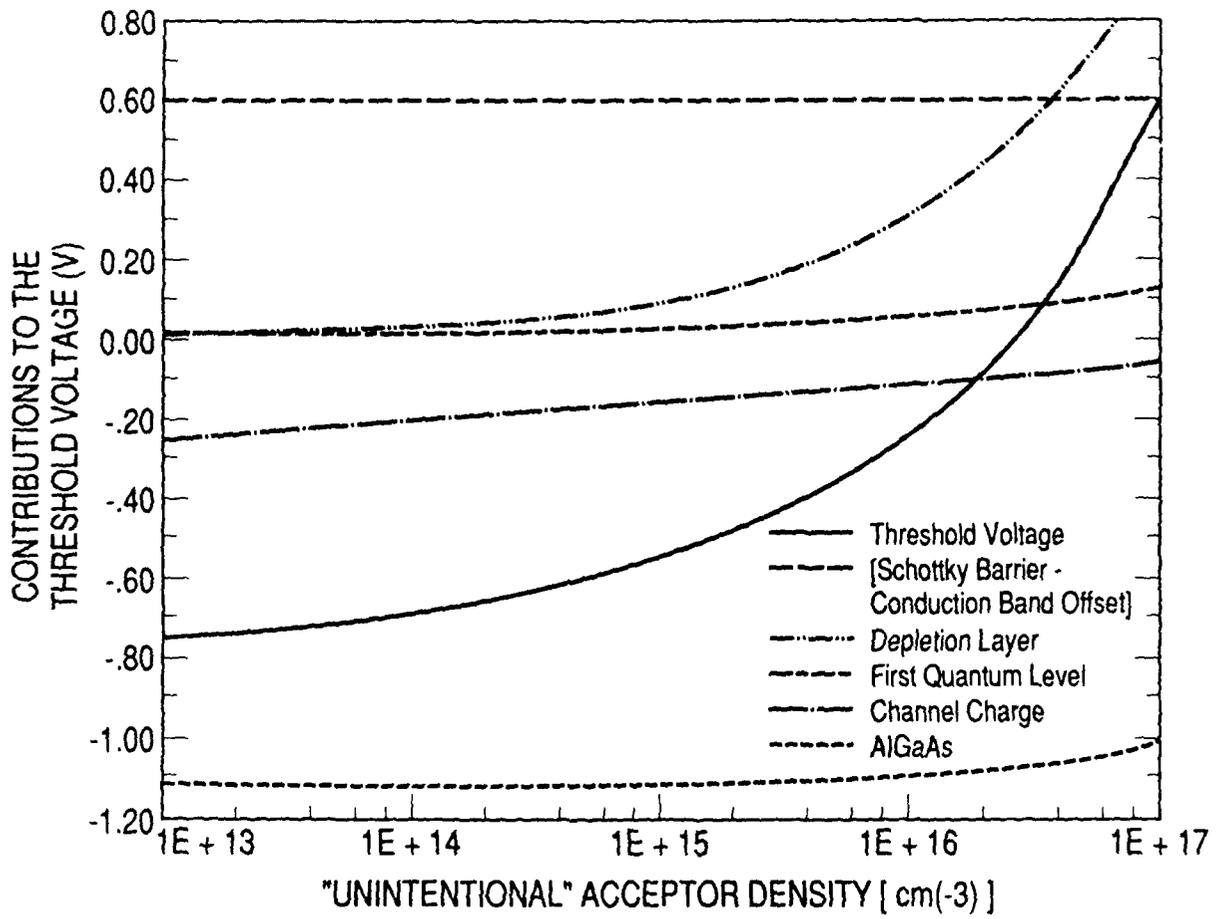


Fig. 2. The Threshold Voltage and Contributions Due to (1) the Height of the Schottky Barrier and the Conduction Band Offset, (2) the Charge of the Depletion Layer, (3) the Position of the First Quantum Level, (4) the Charge Density of the Channel, and (5) the Charge of the AlGaAs Layer versus "Unintentional" Acceptor Doping

As seen in Fig. 2 for unintentional doping densities in the range encountered for typical MBE growth conditions ( $10^{13}$  to  $10^{15}$   $\text{cm}^{-3}$ ), the threshold voltage can increase by as much as 200 mV over this range. At doping densities above this range, the increase in the charge of the depletion layer dominates the increase in the threshold voltage and the threshold voltage rises rapidly.

We have measured the threshold voltage of structures having the geometry cited above with a donor doping density of  $1 \times 10^{18}$   $\text{cm}^{-3}$  and an acceptor doping density of  $5 \times 10^{14}$   $\text{cm}^{-3}$ . The value of the threshold voltage predicted by our strong-inversion model is -0.620 V (see Fig. 2). This is in reasonable agreement with the measured values, which range from -0.41 to -0.73 V. Variations in the measured values result from variations in the width of the AlGaAs layer that arise from nonuniform processing. The value of the width of the AlGaAs layer, 430 Å, is a nominal value that gives rise to the discrepancy between our calculated value and the measured values.

#### IV. SUMMARY AND CONCLUSIONS

We have developed a strong-inversion model, in the depletion layer approximation, of the threshold voltage for modulation doped field-effect transistors (MODFETs). The threshold voltage is defined as the voltage at which the channel charge density is equal to the acceptor density times the average channel width. We have experimentally verified the model by comparing predicted threshold voltages with those measured for typical MBE-grown MODFET structures.

We isolate contributions to the threshold voltage that depend on (1) the difference between the height of the Schottky barrier and the offset of the conduction band, (2) the charge of the depletion layer, (3) the energy of the first quantum level, (4) the charge in the 2-D channel, and (5) the density of the AlGaAs donors. At low acceptor densities (below  $\approx 10^{15}$   $\text{cm}^{-3}$ ), the threshold voltage is dominated by the contribution of the AlGaAs layer, the contribution of the channel charge, and the difference between the height of the Schottky barrier and the conduction band offset.

In MODFETs the unintentional acceptor density is usually on the order of  $10^{13}$  to  $10^{15}$   $\text{cm}^{-3}$  and the channel width is on the order of 100 Å. Even for these apparently small acceptor densities, especially in comparison with the typical AlGaAs donor densities of  $\approx 10^{18}$   $\text{cm}^{-3}$ , we show that the threshold voltage may increase by as much as 200 mV over this range of acceptor densities. Above these acceptor densities the threshold voltage increases rapidly as the charge of the depletion layer increases.

Increasing  $N_d$ , the donor density in the AlGaAs layer, would decrease the contribution of AlGaAs to the threshold voltage, thereby offsetting any threshold voltage increase that results from an increase in the unintentional acceptor doping.

The formalism developed in Section II has been applied to the threshold as defined by the strong inversion of the channel. This formalism is by no means restricted to the description of the threshold alone. If one

were to proceed from the equation for the gate potential, Eq. (15), with the assumption that the channel density  $n_s$  is much larger than the depletion layer charge  $N_a W$ , the original charge-control results of Delagebeaudeuf and Linh<sup>1</sup> would be recovered.

## REFERENCES

1. D. Delagebeaudeuf and N. T. Linh, "Metal-(n) AlGaAs Two-Dimensional Electron Gas FET," IEEE Trans. Electron Devices ED-29 [6], 955-960 (June 1982).
2. T. J. Drummond, H. Morkoc, K. Lee, and M. Shur, "Model for Modulation-Doped Field-Effect Transistor," IEEE Electron Dev. Lett. EDL-3 [11], 338-341 (November 1982).
3. T. J. Drummond, S. L. Su, W. G. Lyons, R. Fischer, and H. Morkoc, "Design and fabrication of high-transconductance modulation-doped (Al,Ga)As/GaAs FETs," J. Vac. Sci. Technol. B1(2), 186-189 (April - June 1983).
4. K. Lee and M. Shur, "Electron density of the two-dimensional electron gas in modulation-doped layers," J. Appl. Phys. 54(4), 2093-2096 (April 1983).
5. B. Vintner, "Subbands and charge control in a two-dimensional electron gas field-effect transistor," Appl. Phys. Lett. 44(3), 307-309 (1 February 1989).
6. J. Yoshida, "Classical versus Quantum Mechanical Calculation of the Electron Distribution at the n-AlGaAs/GaAs Heterointerface," IEEE Trans. Electron Devices ED-33 [1], 154-156 (January 1986).
7. L. P. Sadwick and K. L. Wang, "A Treatise on the Capacitance-Voltage Relation of High-Electron-Mobility Transistors," IEEE Trans. Electron Devices ED-33 [5], 651-656 (May 1986).
8. S. Subramanian, A. S. Vengurlekar, and A. A. Diwan, "Effect of Shallow and Deep Donors on the Equilibrium Electron Density of the Two-Dimensional Electron Gas in a Modulation-Doped Field-Effect Transistor," IEEE Trans. Electron Devices ED-33 [5], 707-711 (May 1986).
9. A. A. Grinberg, "The effect of the two-dimensional gas degeneracy on the I-V characteristics of the modulation-doped field-effect transistor," J. Appl. Phys. 60(3), 1097-1103 (1 August 1986).
10. Y. M. Kim and P. Roblin, "Two-Dimensional Charge-Control Model for MODFETs," IEEE Trans. Electron Devices ED-33 [11], 1644-1651 (November 1986).
11. W. A. Hughes and C. M. Snowden, "Nonlinear Charge Control in AlGaAs/GaAs Modulation-Doped FETs," IEEE Trans. Electron Devices ED-34 [8], 1617-1625 (August 1987).

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Computer Science Laboratory: Program verification, program translation, performance-sensitive system design, distributed architectures for spaceborne computers, fault-tolerant computer systems, artificial intelligence, microelectronics applications, communication protocols, and computer security.

Electronics Research Laboratory: Microelectronics, solid-state device physics, compound semiconductors, radiation hardening; electro-optics, quantum electronics, solid-state lasers, optical propagation and communications; microwave semiconductor devices, microwave/millimeter wave measurements, diagnostics and radiometry, microwave/millimeter wave thermionic devices; atomic time and frequency standards; antennas, rf systems, electromagnetic propagation phenomena, space communication systems.

Materials Sciences Laboratory: Development of new materials: metals, alloys, ceramics, polymers and their composites, and new forms of carbon; non-destructive evaluation, component failure analysis and reliability; fracture mechanics and stress corrosion; analysis and evaluation of materials at cryogenic and elevated temperatures as well as in space and enemy-induced environments.

Space Sciences Laboratory: Magnetospheric, auroral and cosmic ray physics, wave-particle interactions, magnetospheric plasma waves; atmospheric and ionospheric physics, density and composition of the upper atmosphere, remote sensing using atmospheric radiation; solar physics, infrared astronomy, infrared signature analysis; effects of solar activity, magnetic storms and nuclear explosions on the earth's atmosphere, ionosphere and magnetosphere; effects of electromagnetic and particulate radiations on space systems; space instrumentation.