Operating Characteristics of Crosscorrelator With or Without Sample Mean Removal

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Preface

This research was conducted under NUSC Project No. A75205, Subproject No. ZR0000101, "Applications of Statistical Communication Theory to Acoustic Signal Processing," Principal Investigator Dr. Albert H. Nuttall (Code 33), Program Manager CAPT Z. L. Newcomb, Naval Material Command (MAT-05B). Also, this research was conducted under NUSC Project No. C12012, "Sonar Systems Improvement Program," Principal Investigator J. DePrimo (Code 3291), Sponsoring Activity Naval Sea Systems Command, PMS-409, AN/BQQ-5 Program Manager J. Archer.

The Technical Reviewer for this report was Ronald R. Kneipfer (Code 3212).

Reviewed and Approved: 14 August 1984

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New London Laboratory, Naval Underwater Systems Center,
# Operating Characteristics of Crosscorrelator with or Without Sample Mean Removal

**Title**: Operating Characteristics of Crosscorrelator with or Without Sample Mean Removal

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**Abstract**: The characteristic function of the output of a crosscorrelator, with the sample means removed from each channel, is derived in closed form. More generally, if scaled versions of the sample means are subtracted prior to multiplication of the channel inputs and summation, a closed form for the characteristic function of the correlator output is derived. These results are used to plot the exact operating characteristics of the crosscorrelator, as functions of the threshold, the general scaling factors applied to the sample means, the number of channels, etc.
of terms, \( N \), summed to yield the output, the actual means at the inputs, and the signal-to-noise ratios of the random signal components at each of the system inputs. Programs for the various cases considered are documented and exercised. Comparisons are made with a Gaussian approximation, which can be used to extend the results to larger values of \( N \) than considered here, if needed. Asymptotic results for the exceedance distribution functions have also been derived, but they are not too useful for large \( N \).
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OPERATING CHARACTERISTICS OF CROSSCORRELATOR
WITH OR WITHOUT SAMPLE MEAN REMOVAL

INTRODUCTION

The detection of weak signals in two channels is often accomplished by crosscorrelating the two waveforms and comparing with a threshold. For the case where a large number of independent products are added to yield the correlator output, the central limit theorem is often employed, with questionable validity for low false alarm probabilities, i.e., large thresholds. Also, this approximation may not be valid for intermediate numbers of terms added.

Here we wish to get exact operating characteristics for the crosscorrelator, namely detection probability vs. false alarm probability, even for probabilities as low as $10^{-10}$. In particular, we desire results for an arbitrary number of products summed, for any degree of correlation between corresponding individual samples of the two channel inputs, and for any input signal and noise power levels.

Furthermore, it sometimes happens that the two input channels contain dc components, which can be considered either desirable or otherwise, depending on the application. Here we will consider these dc components as nuisance terms and will subtract them out prior to crosscorrelation. More precisely, since the actual values of the dc components in each channel will generally be unknown, we will estimate them via the sample means (over the available record lengths) and subtract these estimates from the available data. This subtraction feature creates new random variables, all of which are statistically dependent on each other, and thereby significantly complicates the analysis. Nevertheless, this crosscorrelation of the sample ac components of the input channels is encountered in practical situations, and in one recent study [1], it was in fact the generalized likelihood ratio detector under two different realistic scenarios. Accordingly, it merits study and accurate quantitative evaluation of performance capability.
More generally, we consider subtraction of scaled versions of the sample means of each channel prior to multiplication and summation. Then as special cases, we can investigate the crosscorrelator with or without sample mean removal, or any intermediate case of interest.

The major analytical result here is a closed form for the characteristic function of the correlator output, in the most compact form involving only two rooting operations and one exponential. Although this processor could be analyzed by the general method given in [2], in terms of the eigenvalues and eigenvectors of a correlation matrix, it would be less accurate and considerably more time consuming, even with computer aid, especially for a large number of terms summed. The actual numerical procedure adopted here for proceeding from the characteristic function to the exceedance distribution functions (false alarm and detection probabilities) is that given in [3], and utilized to advantage in [2,3,4].
PROBLEM DEFINITION

INPUT STATISTICS

The two channel inputs to the crosscorrelator are synchronously sampled in time, yielding random variables \( \{u_n\}_N \) and \( \{v_n\}_N \), where \( N \) is the total number of data samples taken in each channel. These random variables are Gaussian with the following statistics:

\[
\begin{align*}
\text{means} & : \quad \overline{u}_n = u_u, \quad \overline{v}_n = u_v, \\
\text{variances} & : \quad (u_n - \overline{u}_n)^2 = \sigma_u^2, \quad (v_n - \overline{v}_n)^2 = \sigma_v^2, \\
\text{covariances} & : \quad (u_n - \overline{u}_n)(v_n - \overline{v}_n) = \rho \sigma_u \sigma_v,
\end{align*}
\]

(An overbar denotes a statistical average.) That is, the means and variances in each channel, although different, do not change with time, and the degree of correlation between channels is constant. Also

\[
\begin{align*}
\text{if } m \neq n, & \quad u_m \text{ is statistically independent of } u_n, \\
\text{if } m \neq n, & \quad v_m \text{ is statistically independent of } v_n, \\
\text{if } m \neq n, & \quad u_m \text{ is statistically independent of } v_n.
\end{align*}
\]

However, \( u_n \) and \( v_n \) are statistically dependent on each other, for all \( n \), to the extent \( \rho \) indicated in (1).

A SIGNAL AND NOISE MODEL

To better fix the mathematical definitions above, consider in this subsection the following possible signal and noise model:

\[
\begin{align*}
\{u_n\}_N & : \quad u_n = u_u + u_s(n) + u_d(n) \\
\{v_n\}_N & : \quad v_n = u_v + v_s(n) + v_d(n),
\end{align*}
\]

for \( 1 \leq n \leq N \).
where random signal components $u_s(n), v_s(n)$ are zero-mean and partially correlated with each other:

$$
\bar{u_s(n)} = 0, \bar{v_s(n)} = 0, \\
\bar{u_s^2(n)} = S_u, \bar{v_s^2(n)} = S_v, \bar{u_s(n)} \bar{v_s(n)} = \rho_s (S_u S_v)^{1/2}
$$

for all $n$. (4)

Thus $S_u, S_v$ are the powers of the random signal components in each channel.

Also, the random noise disturbances $u_d(n), v_d(n)$ in (3) are zero-mean and independent of each other:

$$
\bar{u_d(n)} = 0, \bar{v_d(n)} = 0, \\
\bar{u_d^2(n)} = D_u, \bar{v_d^2(n)} = D_v, \bar{u_d(n)} \bar{v_d(n)} = 0
$$

for all $n$. (5)

Thus $D_u, D_v$ are the powers of the random noise disturbances in each channel. Finally, except for the statistical dependencies indicated in (4) between $u_s(n)$ and $v_s(n)$, all the 4N random components in (3) are independent of each other.

For this particular signal and noise model in (3)-(5), the master parameters in (1) take the special form

$$
\sigma_u^2 = S_u + D_u, \sigma_v^2 = S_v + D_v, \rho \sigma_u \sigma_v = \rho_s (S_u S_v)^{1/2},
$$

from which there follows

$$
\rho = \rho_s \left( \frac{R_u}{1+R_u} \frac{R_v}{1+R_v} \right)^{1/2},
$$

(7)
where the signal-to-noise ratios (per sample) of the random components in (3) have been defined as

\[ R_u = \frac{S_u}{D_u} = \frac{u_s^2(n)}{u_d^2(n)} , \quad R_v = \frac{S_v}{D_v} = \frac{v_s^2(n)}{v_d^2(n)} \quad \text{for all } n. \quad (8) \]

Thus the parameters \( \sigma_u, \sigma_v, \rho \) in (1) depend only on the statistics of the random components in model (3), and not on the dc components \( u \) and \( v \). Observe that even if \( \rho_s = 1 \) and \( R_u = \infty \), \( \rho \) would still be less than 1; the one noisy channel prevents full correlation between inputs.

**CROSSCORRELATOR OUTPUT**

We define the sample ac components of each channel of the crosscorrelator as

\[
\begin{align*}
\tilde{u}_n &= u_n - \frac{1}{N} \sum_{m=1}^{N} u_m \\
\tilde{v}_n &= v_n - \frac{1}{N} \sum_{m=1}^{N} v_m
\end{align*}
\]

for \( 1 \leq n \leq N \), \( (9) \)

where we have subtracted the corresponding sample means from each and every data sample. Thus \( \{\tilde{u}_n\}_1^N \) and \( \{\tilde{v}_n\}_1^N \) have zero-means and have statistics completely independent of the unknown actual values of input means \( u \), \( v \). However, in trade, we now must deal with a new set of \( 2N \) random variables, all of which are statistically dependent on each other; this is the feature which complicates the ensuing analysis. The test statistic (decision variable) of interest is the crosscorrelator output after sample mean removal,

\[
q = \sum_{n=1}^{N} \tilde{u}_n \tilde{v}_n = \sum_{n=1}^{N} u_n v_n - \frac{1}{N} \sum_{m=1}^{N} u_m \sum_{n=1}^{N} v_n , \quad (10)
\]
which is independent of the actual unknown values of input means \( \mu_u \) and \( \mu_v \). If we knew the input means, we could subtract them directly and not have to resort to sample means.

More generally, we consider the modified channel components

\[
\begin{align*}
\tilde{u}_n &= u_n - \frac{\alpha}{N} \sum_{m=1}^{N} u_m \\
\tilde{v}_n &= v_n - \frac{\beta}{N} \sum_{m=1}^{N} v_m
\end{align*}
\]

for \( 1 \leq n \leq N \) \quad (11)

and the crosscorrelator output

\[
q = \sum_{n=1}^{N} \tilde{u}_n \tilde{v}_n = \sum_{n=1}^{N} u_n v_n - \frac{\gamma}{N} \sum_{m=1}^{N} u_m \sum_{n=1}^{N} v_n ,
\]

instead of (9) and (10). Scale factors \( \alpha \) and/or \( \beta \) in (11) may be unequal to 1; the final parameter \( \gamma \) in (12) is given by

\[
\gamma = \alpha + \beta - \alpha \beta = 1 - (\alpha\beta)(\beta\alpha).
\]

The case of \( \gamma=0 \) in (12) obviously corresponds to the case of no sample mean removal. On the other hand, if either* \( \alpha=1 \) or \( \beta=1 \), then \( \gamma=1 \), and we have removal of the sample mean; i.e., (12) reduces to (10). We shall be interested here in the analysis of the general case represented by (12), for arbitrary \( \gamma \).

* It is demonstrated in appendix A that if scale factor \( \alpha=1 \) but \( \beta\neq1 \), correlator output \( q \) is completely independent of \( \mu_u, \mu_v, \beta \).
CHARACTERISTIC FUNCTION OF CROSSCORRELATOR OUTPUT

DERIVATION

We express the collection of random variables in (1) and (2) in column matrix form according to

\[ U = [u_1 \ u_2 \ \ldots \ u_N]^T, \quad V = [v_1 \ v_2 \ \ldots \ v_N]^T, \quad (14) \]

where superscript \( T \) denotes transpose. The crosscorrelator output \( q \) in (12) can then be written as quadratic form

\[ q = U^T Q V, \quad (15) \]

where: \( N \times N \) matrix

\[ Q = I - \frac{1}{N} 1 1^T, \quad (16) \]

\( I \) is the \( N \times N \) identity matrix, and

\[ 1 = [1 \ 1 \ \ldots \ 1]^T \quad (17) \]

is a \( N \times 1 \) column matrix of ones.

Since \( U \) and \( V \) are Gaussian, their joint probability density function is, in terms of the parameters in (1),

\[
\begin{equation}
p(U,V) = \left[ 2\pi \sigma_u \sigma_v \left(1-\rho^2\right)^{1/2} \right]^{-N} \exp \left[ -\frac{1}{2(1-\rho^2)} \right. \\
\left. \left\{ \frac{1}{\sigma_u^2} (U - u_u 1)^T (U - u_u 1) + \frac{1}{\sigma_v^2} (V - u_v 1)^T (V - u_v 1) - \frac{2\rho}{\sigma_u \sigma_v} (U - u_u 1)^T (V - u_v 1) \right\} \right] (18)
\end{equation}
\]

The characteristic function of correlator output \( q \) in (15) is then given by the statistical average

7
\[ f_q(\xi) = \exp(i\xi q) = \exp(i\xi U^T Q V) = \]
\[ \left[ 2\pi\sigma_u\sigma_v \right]^{-1/2} \int dU dV \ p(U, V) \exp \left[ i\xi U^T Q V - \frac{1}{2(1-\rho^2)} \right] \]
\[ \cdot \left\{ \frac{1}{\sigma_u^2} \left( \frac{1}{2} (U - \mu_u) (U - \mu_u)^T + \frac{1}{2} (V - \mu_v) (V - \mu_v)^T - \frac{2\rho}{\sigma_u \sigma_v} (U - \mu_u) (V - \mu_v)^T \right) \right\}. \tag{19} \]

At this point, in order to evaluate this 2N-fold integral, we employ the general integral result (B-2) and (B-6) in appendix B, identifying the matrices there as

\[ A = \frac{1}{\sigma_u^2(1-\rho^2)} I, \quad B = \frac{1}{\sigma_v^2(1-\rho^2)} I, \quad C = i\xi Q + \frac{\rho}{\sigma_u \sigma_v (1-\rho^2)} I, \]
\[ D = \frac{\sigma_v \mu_u - \rho \sigma_u \mu_v}{\sigma_u^2 \sigma_v (1-\rho^2)} I, \quad E = \frac{\sigma_v \mu_u - \rho \sigma_v \mu_u}{\sigma_v^2 (1-\rho^2)} I. \tag{20} \]

We also need the following auxiliary results for special matrix forms; namely, for arbitrary scalars \( c_1, c_2 \), the matrix determinant

\[ \det(c_1 I + c_2 11^T) = c_1^{N-1}(c_1 + Nc_2), \tag{21} \]

and the matrix inverse

\[ (c_1 I + c_2 11^T)^{-1} = \frac{1}{c_1} I - \frac{c_2}{c_1(c_1 + Nc_2)} 11^T. \tag{22} \]

Employment of appendix B and (20)-(22) then yields, after a very considerable amount of effort, a closed form for the characteristic function in (19) (in its most compact form)
The square roots in (23) are principal value, being +1 at $\xi=0$. This characteristic function has four branch points and two essential singularities which overlap two of the branch points; the complexity of this characteristic function of $q$ precludes tractable analytical results for the probability density function or exceedance distribution function of the correlator output, except in very special cases. Nevertheless, since the characteristic function in (23) is easily numerically evaluated with computer aid, it readily lends itself to the procedure presented in [2,3]. A program for the evaluation of the cumulative and exceedance distribution functions corresponding to characteristic function (23)-(24) is given in appendix C for arbitrary values of

\begin{align*}
N, & \quad \text{number of terms summed} \\
\gamma, & \quad \text{scale factor in sample mean removal} \\
\mu_u, & \quad \text{mean in } u\text{-channel} \\
\mu_v, & \quad \text{mean in } v\text{-channel} \\
\sigma_u, & \quad \text{standard deviation in } u\text{-channel} \\
\sigma_v, & \quad \text{standard deviation in } v\text{-channel} \\
\rho, & \quad \text{correlation coefficient between channels.}
\end{align*}
A sample plot of the cumulative and exceedance distribution functions for a typical selection of numerical values for the above parameters is also presented in appendix C.

CUMULANTS OF CORRELATOR OUTPUT

By taking the natural logarithm of the characteristic function in (23) and expanding in a power series in $\xi$, the cumulants of random variable $q$ can be extracted:

$$K_q(n) = \frac{1}{2}(n-1)! [N-1+(1-\gamma)^n] \left( \sigma_u \sigma_v \right)^n (S^n + D^n) +$$

$$+ \frac{1}{n!} N(1-\gamma)^n \left( \sigma_u \sigma_v \right)^{n-1} u_u u_v (S^{n-1} + D^{n-1}) +$$

$$+ \frac{1}{4n!} N(1-\gamma)^n \left( \sigma_u \sigma_v \right)^{n-2} (\sigma_u^2 \sigma_v^2 + \sigma_u \sigma_v^2 + \sigma_v \sigma_u^2) (S^{n-2} - D^{n-2}) , \quad (25)$$

where here

$$S = \rho + 1, \quad D = \rho - 1 . \quad (26)$$

In particular, the mean and variance of $q$ are available by using $n=1$ and $2$ respectively in (25):

$$\mu_q = (N-\gamma) \rho \sigma_u \sigma_v + N(1-\gamma) u_u u_v ,$$

$$\sigma_q^2 = (N-2\gamma + \gamma^2)(1+\rho^2) \sigma_u^2 \sigma_v^2 + N(1-\gamma)^2 (\sigma_u^2 \sigma_v^2 + \sigma_u \sigma_v^2 + 2\rho \sigma_u \sigma_v u_u u_v) . \quad (27)$$

SPECIAL CASE OF $\gamma=1$, SAMPLE MEAN REMOVAL

For $\gamma=1$, the general characteristic function in (23) reduces to

$$f_q(\xi; \gamma=1) = \left( 1 - i \xi E_1 + \xi^2 E_2 \right)^{-\frac{N-1}{2}} , \quad (28)$$
where $E_1$ and $E_2$ are still given by (24), and are independent of means $\mu_u$ and $\mu_v$, as shown earlier. The cumulants in (25) reduce to

$$\chi_q(n) = \frac{1}{2}(n-1)!\frac{(N-1)(\sigma_u\sigma_v)^n}{[(\rho+1)^n + (\rho-1)^n]},$$

(29)

and in particular, the mean and variance of $q$ are

$$\mu_q = (N-1)p\sigma_u\sigma_v,$$

$$\sigma_q^2 = (N-1)(1+p^2)\sigma_u^2\sigma_v^2.$$  

(30)

**SPECIAL CASE OF $\gamma=0$, SAMPLE MEAN NOT REMOVED**

For $\gamma=0$, the characteristic function in (23) reduces to

$$f_q(\xi;\gamma=0) = (1 - iE_1 + \xi^2E_2)^{-N/2} \exp\left[ \frac{i\xi G_1^{(0)} + i\xi G_2^{(0)}}{1 - iE_1 + \xi^2E_2} \right].$$

(31)

where

$$E_1 = 2p\sigma_u\sigma_v, \quad E_2 = \sigma_u^2(1-p^2),$$

$$G_1^{(0)} = Nu_u\mu_v, \quad G_2^{(0)} = \frac{1}{2}N(\sigma_u^2\mu_v^2 + \sigma_v^2\mu_u^2 - 2p\sigma_u\sigma_v\mu_u\mu_v).$$

(32)

The cumulants are obtained by setting $\gamma=0$ in (25), and in particular, the mean and variance of correlator output $q$ are

$$\mu_q = N(p\sigma_u\sigma_v + \mu_u\mu_v),$$

$$\sigma_q^2 = N\left[(1+p^2)\sigma_u^2\sigma_v^2 + \sigma_u^2\mu_v^2 + \sigma_v^2\mu_u^2 + 2p\sigma_u\sigma_v\mu_u\mu_v\right].$$

(33)
INTERRELATIONSHIP OF TWO SPECIAL CASES

Let the general characteristic function in (23) be denoted by $f_q(\xi; N, \gamma, \mu_u, \mu_v)$. We have already seen the expression for $f_q(\xi; N, 1, \mu_u, \mu_v)$ in (28). At the same time, from (23) and (24), there follows

\[ f_q(\xi; N-1, 0, 0, 0) = \left(1 - \gamma E_1 + \gamma^2 E_2\right)^{-\frac{N-1}{2}}, \tag{34} \]

which is identical to (28). That is,

\[ f_q(\xi; N, 1, \mu_u, \mu_v) = f_q(\xi; N-1, 0, 0, 0). \tag{35} \]

Thus the characteristic functions of the two following random variables are identical:

1. Sum of $N$ terms with sample mean removal, and the true means arbitrary,

2. Sum of $N-1$ terms without sample mean removal, but the true means zero. \tag{36}

The removal of the sample means has eliminated the dependence of the correlator output on the unknown means but has reduced the number of degrees of freedom by 1.

SPECIALIZATION TO THE SIGNAL AND NOISE MODEL

For general scaling factor $\gamma$ and arbitrary input means $\mu_u$, $\mu_v$, and for the model introduced earlier in (3)-(6), the general characteristic function of the correlator output is still given by (23), but with the parameters in (24) now specialized to the form
\[ E_1 = 2\rho_s (S_u S_v)^{1/2}, \quad E_2 = D_u D_v + D_u S_v + D_v S_u + S_u S_v (1-\rho_s^2), \]

\[ F_1 = E_1 (1-\gamma), \quad F_2 = E_2 (1-\gamma)^2, \]

\[ G_1 = N (1-\gamma) u_u u_v, \quad G_2 = \frac{1}{2} N (1-\gamma)^2 [(S_u + D_u) u_v^2 + (S_v + D_v) u_u^2 - 2\rho_s (S_u S_v)^{1/2} u_u u_v] \quad \text{(37)} \]

The general n-th cumulant is still given by (25); however, the use of (6) allows for determination in terms of the fundamental quantities of the signal and noise model, namely \( S_u, S_v, D_u, D_v, \rho_s \) defined in (4)-(5). In particular, the mean and variance of correlator output \( q \) are

\[ u_q = (N-\gamma) \rho_s (S_u S_v)^{1/2} + N (1-\gamma) u_u u_v, \]

\[ \sigma_q^2 = (N-2\gamma+\gamma^2) [D_u D_v + D_u S_v + D_v S_u + (1+\rho_s^2) S_u S_v] + N (1-\gamma)^2 [(S_u + D_u) u_v^2 + (S_v + D_v) u_u^2 + 2\rho_s (S_u S_v)^{1/2} u_u u_v]. \quad \text{(38)} \]
ANALYTIC RESULTS FOR $\gamma=1$, SAMPLE MEAN REMOVAL

In this section and the next, we will confine attention solely to the case of scale factor $\gamma=1$. The characteristic function of the crosscorrelator output $q$ follows from (28) and (24) as

$$f_q(\xi) = \left[1 - i\xi E_1 + \xi^2 E_2\right] - \frac{N-1}{2} = \left[1 - i\xi 2\rho \sigma_u \sigma_v + \xi^2 \sigma_u^2 \sigma_v^2 (1-\rho^2)\right] - \frac{N-1}{2} =$$

$$= \left\{\left[1 + i\xi \sigma_u \sigma_v (1-\rho)\right] \left[1 - i\xi \sigma_u \sigma_v (1+\rho)\right]\right\} - \frac{N-1}{2} \quad \text{for } \gamma=1 , \quad (39)$$

where we must have $N>2$. We observe, for later numerical use in appendix D, that since $|1 \pm i\xi b| = (1 + \xi^2 b^2)^{1/2}$ is monotonically increasing for $\xi \geq 0$, then $|f_q(\xi)|$ is monotonically decreasing for all $\xi \geq 0$ and any $N, \sigma_u, \sigma_v, \rho$.

GENERAL PROBABILITY RESULTS

The cumulants of $q$ have already been listed in (29) and (30). The probability density function corresponding to characteristic function (39) is given by [5, 6.99 12]

$$p_q(u) = \left[\Gamma\left(\frac{N-1}{2}\right) \pi^{1/2} (1-\rho^2)^{1/2} \sigma_u \sigma_v\right]^{-1} \left(\frac{|u|}{2\sigma_u \sigma_v}\right)^{N-1} *$$

$$* K_{N-1} \left(\frac{|u|}{\sigma_u \sigma_v (1-\rho^2)}\right) \exp\left(-\frac{\rho u}{\sigma_u \sigma_v (1-\rho^2)}\right) \quad \text{for all } u, \quad \gamma=1 , \quad (40)$$

where $K_{N-1}(z)$ is a modified Bessel function of the second kind [6, section 9.6]. If the number of terms added, $N$, to yield correlator output $q$, is odd, simple relations for the probability density function in (40) can be obtained [6, 10.2.15 and 10.1.9, last equation]; letting $n = \frac{N-3}{2}$ for $N$ odd, we find the exact result
\[ p_q(u) = \frac{(1-p^2)^n}{2\sigma_u^4} \sum_{m=0}^{n} \frac{(2n-m)!}{(n-m)!} \exp\left(\frac{\rho u - |u|}{\sigma_u \sigma_v(1-p^2)}\right) \]

for all \( u; n = \frac{N-3}{2}, \ N = 3, 5, 7, \ldots \) \( (41) \)

For example, for \( N=3 \), we have \( n=0 \), yielding

\[ p_q(u) = \frac{1}{2\sigma_u^4} \exp\left(\frac{\rho u - |u|}{\sigma_u \sigma_v(1-p^2)}\right) \] for all \( u \). \( (42) \)

The corresponding cumulative distribution function for \( N=3 \) is

\[ p_q(u) = \int_{-\infty}^{u} dt \ p_q(t) = \frac{1-p}{2} \exp\left(\frac{-\rho u}{\sigma_u \sigma_v(1-p)}\right) \] for \( u \leq 0 \), \( (43A) \)

while the exceedance distribution function is

\[ 1 - p_q(u) = \int_{u}^{+\infty} dt \ p_q(t) = \frac{1+p}{2} \exp\left(\frac{-u}{\sigma_u \sigma_v(1+p)}\right) \] for \( u \geq 0 \). \( (43B) \)

This dichotomy, of presenting the cumulative distribution function for negative arguments, and the exceedance distribution function for positive arguments, turns out to be notationally convenient and physically meaningful and will be adopted throughout this report.

Although closed form expressions for the exceedance distribution function corresponding to probability density function (40) are not available for general \( N \), the use of [6, 9.7.2] on (40) leads to the dominant term in the asymptotic expansion of the exceedance distribution function:

\[ 1 - p_q(u) \sim \frac{1+p}{2\Gamma\left(\frac{N-3}{2}\right)} \left(\frac{u}{2\sigma_u \sigma_v}\right)^{\frac{N-3}{2}} \exp\left(\frac{-u}{\sigma_u \sigma_v(1+p)}\right) \] as \( u \rightarrow +\infty \). \( (44) \)

For \( N=3 \), this is precise; see (43B).
POSSIBLE NORMALIZATIONS OF $q$

If we define a normalized random variable

$$x = \frac{q}{\sqrt{E_2}} = \frac{q}{\sigma_u \sigma_v (1 - \rho^2)^{1/2}},$$

then the characteristic function of $x$ is given by (39) as

$$f_x(\xi) = f_q(\frac{\xi}{\sqrt{E_2}}) = \left[1 - i(1 - \rho^2) \xi^2 + \xi^2\right]^{-\frac{N-1}{2}} =$$

$$= \left[1 - i(1 - \rho^2) \xi^2 + \xi^2\right]^{-\frac{N-1}{2}},$$

which has only two fundamental parameters, namely, $N$ and $\rho$.

A second possibility is the random variable defined by

$$y = \frac{q}{\sigma_u \sigma_v},$$

for which characteristic function

$$f_y(\xi) = f_q\left(\frac{\xi}{\sigma_u \sigma_v}\right) = \left[1 - i(1 - \rho^2) \xi^2 + \xi^2\right]^{-\frac{N-1}{2}},$$

also depends only on $N$ and $\rho$. However, neither of the normalizations, (45) and (47), are of interest to us here; an alternative normalization and reasons for its selection are given below.
SPECIALIZATION TO THE SIGNAL AND NOISE MODEL

For the model presented earlier in (3)-(8), the original $E_1$, $E_2$ parameters in (24) take the form already given in the upper line of (37). Let a normalized random variable, relative to the additive random noise disturbances, be defined according to

$$h = \frac{q}{(D_u D_v)^{1/2}}; \quad (49)$$

see (5). This normalization for the particular signal model (3) is different from both $x$ and $y$ in the general case above. The reason we employ $h$ is that the normalization depends only on the power of the additive noise disturbances, and not on the signal strengths or correlation coefficients; this is consistent with a system which monitors the noise-only background and sets a threshold for a desired false alarm probability.

The characteristic function of the normalized random variable $h$ in (49) is given by

$$f_h(q) = f_q\left(q/(D_u D_v)^{1/2}\right) = \left[1 - i\xi - \frac{E_1}{(D_u D_v)^{1/2}} + \xi^2 \frac{E_2}{D_u D_v}\right]^{-\frac{N-1}{2}} =$$

$$= \left[1 - i\xi 2\alpha + \xi^2 (\beta^2 - \alpha^2)\right]^{-\frac{N-1}{2}}, \quad (50)$$

where we define auxiliary parameters here as

$$\alpha = \rho_s (R_u R_v)^{1/2}, \quad \beta = [(1+R_u)(1+R_v)]^{1/2}. \quad (51)$$

Here we used (39), (37), and (8). This characteristic function in (50) depends on the four fundamental parameters $N$, $\rho_s$, $R_u$, $R_v$, where the latter two quantities are the signal-to-noise ratios per sample of the random components of model (3); see (8).
Reference to (40) reveals that the probability density function of \( h \) corresponding to characteristic function (50) is given by

\[
p_h(u) = \left[ \frac{(N-1)}{2} \pi^{1/2} (\beta^2 - \alpha^2)^{1/2} \right]^{-1} \left( \frac{|u|}{2\beta} \right)^{1/2} \ast \left( \frac{|u|}{2\beta} \right)^{N-1} \ast \frac{K_n}{2} - 1 \left( \frac{u}{\beta^2 - \alpha^2} \right) \exp \left( \frac{\alpha u}{\beta^2 - \alpha^2} \right) \text{ for all } u.
\]

(52)

For \( N \) odd, alternative forms are available from (41), if desired. The asymptotic behavior of the exceedance distribution function of \( h \) follows in a manner similar to that used for (44):

\[
1 - p_h(u) \sim \frac{1+\alpha/\beta}{2\pi \left( \frac{N-1}{2} \right)} \left( \frac{u}{2\beta} \right)^{N-3} \exp \left( \frac{-u}{\beta+\alpha} \right) \text{ as } u \to +\infty.
\]

(53)

The cumulants of \( h \) follow from (29), (26), and (6)-(8):

\[
\chi_h(n) = \frac{1}{2}(n-1)! \left( (\alpha+\beta)^n + (\alpha-\beta)^n \right),
\]

and in particular, the mean and variance of \( h \) are

\[
\mu_h = (N-1)\alpha = (N-1)\rho_S (R_u R_v)^{1/2},
\]

\[
\sigma_h^2 = (N-1)(\alpha^2 + \beta^2) = (N-1)[1 + R_u + R_v + R_u R_v (1 + \rho_S^2)].
\]

(55)

The two parameters, \( \alpha \) and \( \beta \), are given here by (51), in terms of the fundamental quantities \( R_u, R_v, \rho_S \) of the signal and noise model.

REDUCTION TO IDENTICAL SIGNAL COMPONENTS

At this point, we will further specialize the results for the signal and noise model in the above subsection. We presume that
\[ R_u = R_v = R \quad \text{and} \quad \rho_s = 1, \quad (56A) \]
giving, from (51),

\[ \alpha = R, \quad \beta = 1 + R; \quad (56B) \]
that is, the signal-to-noise ratios in the two channels are equal, and the two channel signals are fully correlated. This corresponds physically to a case where the random signal components in (3) are identical, \( u_s(n) = v_s(n) \), and the independent random noise disturbances have the same power level. This situation will hold for the rest of this section and all of the next section where the graphical results are presented.

Equations (50) and (56B) then yield the characteristic function for normalized random variable \( h \) in (49) as

\[
\begin{align*}
f_h(\xi) &= \left[1 - i\xi 2R + \xi^2(1+2R)\right]^{-\frac{N-1}{2}} \\
&= \left[(1 + i\xi)(1 - i\xi(1+2R))\right]^{-\frac{N-1}{2}}. \quad (57)
\end{align*}
\]
The cumulants in (54) reduce to

\[
\chi_h(n) = \frac{1}{2}(n-1)! (N-1)[(1+2R)^n+(-1)^n], \quad (58)
\]
and in particular, the mean and variance become

\[
u_h = (N-1)R, \quad \sigma_h^2 = (N-1)(1+2R+2R^2). \quad (59)
\]

All the above statistical descriptions depend only on the two parameters \( R \), the per-sample signal-to-noise ratio, and \( N \), the number of terms added to yield the correlator output.
The probability density function for $h$ follows from (52) as

$$p_h(u) = \left[ \frac{1}{\Gamma\left(\frac{N-1}{2}\right)} \pi^{1/2} (1+2R)^{1/2} \right]^{-1} \left( \frac{u}{2(1+R)} \right)^{N-1} \times$$

$$\times K_{\frac{N}{2}-1} \left( \frac{1+R}{1+2R} |u| \right) \exp\left( \frac{Ru}{1+2R} \right) \quad \text{for all } u,$$

(60)

and the asymptotic exceedance distribution function from (53):

$$1 - P_h(u) = \frac{1+2R}{2(1+R)\Gamma\left(\frac{N-1}{2}\right)} \left( \frac{u}{2(1+R)} \right)^{N-3} \exp\left( -\frac{u}{1+2R} \right) \quad \text{as } u \to +\infty.$$  

(61)

An important word of caution must be mentioned at this point: when $N$ is large, (61) is inadequate for evaluating small false alarm and detection probabilities, since the succeeding terms in the asymptotic expansion contribute significantly. For example, when $R=0$, the maximum value of the dominant term (61) occurs when $u = (N-3)/2$ which, for $N=128$, yields false alarm probability $3.86E-21$, a value far below those of interest. Thus (61) has limited applicability, being best for small $N$; in fact, the first correction term to (61) yields the multiplicative factor

$$1 + \frac{1+2R}{1+R} \frac{(N-3)(N+3+4R)}{8u}.$$  

(62)

It indicates that, for large $N$, $u$ must be of the order of $N^2$ in order for the dominant term (61) to be fairly accurate.

Although for $N$ odd, an alternative closed form to the probability density function (60) of $h$ is available from (41), the exceedance distribution function will generate a double sum and be rather cumbersome for large $N$. On the other hand, the characteristic function in (57) decays rapidly with $N$ when $N$ is large and yields very nicely to the numerical approach given in [2,3]. The only difficult cases are in fact those for small $N$; accordingly, some analytic results for $N = 2, 3, 4, 5$ will now be presented, based on characteristic function (57).
GENERAL DISTRIBUTION INTEGRALS

Suppose a random variable $y$ has characteristic function $f_{y}(\xi)$. The cumulative distribution function of $y$ can be written as a contour integral:

$$P_{y}(u) = -\frac{1}{12\pi} \int_{C_{+}} d\xi \frac{f_{y}(\xi)}{\xi} \exp(-iu\xi) \quad \text{for all } u,$$

where $C_{+}$ is a contour along the real axis of the complex $\xi$-plane, with an upward indentation at the origin $\xi=0$, to avoid the pole of the integrand there.

Similarly, the exceedance distribution function of random variable $y$ can be expressed as:

$$1 - P_{y}(u) = \frac{1}{12\pi} \int_{C_{-}} d\xi \frac{f_{y}(\xi)}{\xi} \exp(-iu\xi) \quad \text{for all } u,$$

where $C_{-}$ is a contour along the real $\xi$ axis, with a downward indentation at $\xi=0$.

For $u<0$, both contours can be closed in the upper-half $\xi$-plane, since the exp term furnishes rapid decay there. Similarly, for $u>0$, both contours can be closed in the lower-half $\xi$-plane, to realize exponential decay on the circular arcs tending to infinity.

DISTRIBUTIONS FOR N=2

From (57), the characteristic function of normalized correlator output $h$ is

$$f_{h}(\xi) = [(1+i\xi)(1-i\xi(1+2R))]^{-1/2},$$

and the probability density function follows from (60) as
\[ p_h(u) = \frac{1}{\pi (1+2R)^{1/2}} K_0 \left( \frac{1+R}{1+2R} |u| \right) \exp \left( \frac{Ru}{1+2R} \right) \quad \text{for all } u . \quad (66) \]

There is no closed form for the indefinite integral of a \( K_0 \) function; see [6, 11.1.8 and 11.1.9]. Instead, we use (65) in (63) and move the contour upwards until it wraps around the branch point at \( \xi = i \) and extends vertically from there; this is in fact the steepest descent direction for the exponential. The contributions of the small and large circular arcs tend to zero as the radii tend to zero and infinity, respectively. Under a change of variable, there follows the cumulative distribution function in the form

\[ P_h(u) = \frac{2}{\pi} \int_0^{\infty} \frac{dt \exp \left[ u \left( \frac{t^2}{1+t^2} \right) \right]}{(1+t^2) \left[ 1+(1+2R)(1+t^2) \right]^{1/2}} \quad \text{for } u < 0 . \quad (67) \]

This is a useful exact result for several reasons: the integrand decays rapidly, has no cusps, and involves only elementary functions which are easily computed; also the integral is a sum of positive quantities and retains significance even for large \( |u| \).

In a similar fashion, if characteristic function (65) is substituted in (64) and the contour moved down and wrapped around the branch point at \( \xi = -i/(1+2R) \) and along the vertical steepest descent direction for the exponential, the exceedance distribution function becomes, upon a change of variable,

\[ 1 - P_h(u) = \frac{2}{\pi} (1+2R) \exp \left( \frac{u}{1+2R} \right) \int_0^{\infty} \frac{dt \exp \left( -ut^2 \right)}{[1+(1+2R)t^2][1+(1+2R)(1+t^2)]^{1/2}} \quad \text{for } u > 0 . \quad (68) \]

This is useful for the same reasons given above.

There is one closed form result possible; namely, for \( u = 0 \), direct integration of probability density function (66) yields [5, 6.611 9]

\[ P_h(0) = \frac{1}{\pi} \arccos \left( \frac{R}{1+R} \right) , \quad 1-P_h(0) = \frac{1}{\pi} \arccos \left( \frac{1-R}{1+R} \right) . \quad (69) \]
DISTRIBUTIONS FOR $N=3$

Use of [6, 10.2.17] on (60) immediately yields probability density function

$$p_n(u) = \begin{cases} \frac{1}{2(1+R)} \exp(u) & \text{for } u \leq 0 \\ \frac{1}{2(1+R)} \exp\left(\frac{-u}{1+2R}\right) & \text{for } u > 0 \end{cases}$$

The cumulative and exceedance distribution functions easily follow as

$$P_n(u) = \frac{1}{2(1+R)} \exp(u) \quad \text{for } u \leq 0 ,$$

$$1 - P_n(u) = \frac{1+2R}{2(1+R)} \exp\left(\frac{-u}{1+2R}\right) \quad \text{for } u > 0 .$$

This latter result corroborates (61) and (62).

DISTRIBUTIONS FOR $N=4$

The only closed form result possible is obtained by direct integration of probability density function (60) to get origin value

$$P_n(0) = \frac{1}{\pi} \left[ \arccos\left(\frac{R}{1+R}\right) - \frac{R(1+2R)1/2}{(1+R)^2} \right] .$$

This follows by use of the integral

$$\int_{0}^{\infty} dx e^{-\alpha x} x K_1(\beta x) = \frac{\beta \arccos(\alpha/\beta)}{(\beta^2-\alpha^2)^{3/2}} - \frac{\alpha}{\beta(\beta^2-\alpha^2)} \quad \text{for } \beta > -\alpha ,$$

which follows from [5, 6.611 9] by applying $\alpha/\alpha \beta$ to both sides.
DISTRIBUTIONS FOR N=5

Use of \([6, 10.2.17]\) on (60) immediately yields probability density function

\[
p_h(u) = \begin{cases} 
\frac{1+2R-(1+R)u}{4(1+R)^3} \exp(u) & \text{for } u \leq 0 \\
\frac{1+2R+(1+R)u}{4(1+R)^3} \exp\left(-\frac{u}{1+2R}\right) & \text{for } u \geq 0 
\end{cases}
\]  

(74)

The cumulative and exceedance distribution functions follow as

\[
p_h(u) = \frac{2+3R-(1+R)u}{4(1+R)^3} \exp(u) \quad \text{for } u \leq 0,
\]

\[
1 - p_h(u) = \frac{1+2R}{4(1+R)^3} \left[(1+2R)(2+R)+(1+R)u\right] \exp\left(-\frac{u}{1+2R}\right) \quad \text{for } u \geq 0.
\]  

(75)

This latter result corroborates (61) and (62). Also, this example was used as a check on the numerical procedure \([3]\) applied directly to the characteristic function, which is used in the following section; the agreement was ten decimals for numerous values of R and u.
SUMMARY OF PARTICULAR CASE CONSIDERED

We first summarize here the particular case that will be considered quantitatively in this section. The input samples are

\[
\begin{align*}
    u_n &= u_u + u_s(n) + u_d(n) \\
    v_n &= u_v + v_s(n) + v_d(n)
\end{align*}
\]

for \(1 \leq n \leq N, \quad (76)\)

where these Gaussian random variables have statistics

\[
\begin{align*}
    \overline{u_s(n)} = \overline{v_s(n)} = \overline{u_d(n)} = \overline{v_d(n)} &= 0, \\
    \overline{u_s^2(n)} &= S_u, \quad \overline{v_s^2(n)} = S_v, \quad \overline{u_s(n)v_s(n)} = (S_u S_v)^{1/2}, \\
    \overline{u_d^2(n)} &= D_u, \quad \overline{v_d^2(n)} = D_v, \quad \overline{u_d(n)v_d(n)} = 0. \quad (77)
\end{align*}
\]

We presume that the simultaneous signal components \(u_s(n), v_s(n)\) in the two channels are fully correlated, that all other random variables are independent, and that the two channel input signal-to-noise ratios

\[
\frac{S_u}{D_u} = \frac{S_v}{D_v} = R. \quad (78)
\]

have a common value \(R\). More general situations have been considered in earlier sections; however, only this special case will be numerically evaluated here.
The normalized crosscorrelator output, with sample mean removal ($\gamma=1$), is

$$h = \frac{1}{(D_u D_v)^{1/2}} \sum_{n=1}^{N} \tilde{u}_n \tilde{v}_n,$$

(79)

where the sample ac components

$$\tilde{u}_n = u_n - \frac{1}{N} \sum_{m=1}^{N} u_m,$$

$$\tilde{v}_n = v_n - \frac{1}{N} \sum_{m=1}^{N} v_m.$$

(80)

The characteristic function of $h$ is given by (57) as

$$f_h(\xi) = \left[ (1+i\xi)(1-i\xi(1+2R)) \right]^{-\frac{N-1}{2}}$$

(81)

and depends only on signal-to-noise ratio $R$ and number of terms $N$. We must have $N \geq 2$.

If $R=0$ and we evaluate the exceedance distribution function corresponding to (81), we then have the false alarm probability. But when $R>0$, the exceedance distribution function corresponding to (81) is the detection probability. In the following, we plot the detection probability vs. the false alarm probability, with signal-to-noise ratio $R$ as a parameter; different values of $N$ are handled in separate plots.

OPERATING CHARACTERISTICS FOR $\gamma=1$

A sample program for evaluation of the cumulative and exceedance distribution functions corresponding to characteristic function (81), and thereby the detection probability vs. false alarm probability operating characteristics of the crosscorrelator with sample mean removal, is given in appendix D. It is heavily based on the technique developed and explained in [3].
In figures 1-14 are presented the operating characteristics for the crosscorrelator with sample mean removal, for values of

$$N = 2, 3, 4, 6, 8, 12, 16, 24, 32, 48, 64, 96, 128, 256,$$  \hspace{1cm} (82)

respectively. The case of \(N=2\) was accomplished by use of (67)-(69); results for \(N=3\) relied on (71); and the remainder for \(N>4\) employed a numerical procedure \[3\] proceeding directly from characteristic function (81) to the exceedance distribution function. False alarm probabilities \(P_F\) in the range \(1E^{-10}\) to \(.5\) and detection probabilities \(P_D\) covering \(1E^{-10}\) to \(.999\) are presented. The abscissa and ordinate on these plots are according to a normal probability transformation, as explained below. Values of signal-to-noise ratio \(R\) are taken as \(R=2^n\), where \(n\) assumes values appropriate for each plot in order to cover the full range of probabilities of interest.

GAUSSIAN APPROXIMATION

Suppose the decision variable of a processor is Gaussian with mean and standard deviation \(m_0, \sigma_0\) respectively when the input signal is absent, and \(m_1, \sigma_1\) when signal is present. Then for threshold \(\lambda\), the false alarm probability and detection probability are

$$P_F = \int_{-\infty}^{+\infty} \frac{1}{\sigma_0} \phi \left( \frac{u-m_0}{\sigma_0} \right) \, du = \Phi \left( \frac{m_0-\lambda}{\sigma_0} \right),$$  \hspace{1cm} (83)

$$P_D = \int_{-\infty}^{+\infty} \frac{1}{\sigma_1} \phi \left( \frac{u-m_1}{\sigma_1} \right) \, du = \Phi \left( \frac{m_1-\lambda}{\sigma_1} \right),$$

respectively, where \(\phi\) and \(\Phi\) are the normalized Gaussian probability density function and cumulative distribution function:

$$\phi(u) = (2\pi)^{-1/2} \exp(-u^2/2), \quad \Phi(u) = \int_{-\infty}^{u} dt \phi(t).$$  \hspace{1cm} (84)

If we let \(\Phi_1\) be the inverse function to \(\Phi\), and define

$$x = \Phi_1(P_F), \quad y = \Phi_1(P_D),$$  \hspace{1cm} (85)

*See page 51 et seq.*
then threshold $A$ can be eliminated from (83) to yield

$$y = \frac{m_1 - m_0 + \sigma_0 x}{\sigma_1}.$$  \hspace{1em} (86)

Equation (85) corresponds to the transformation to normal probability coordinates; thus a plot of $P_D$ vs $P_F$ on normal probability paper is the straight line (86) when the decision variable is Gaussian under both hypotheses of signal absent as well as present.

Reference to (59) reveals that, for our application,

$$m_0 = 0, \quad m_1 = (N-1)R, \quad \sigma_0^2 = N-1, \quad \sigma_1^2 = (N-1)(1+2R+2R^2),$$  \hspace{1em} (87)

since setting signal-to-noise ratio $R=0$ corresponds to hypothesis 0, signal absent. Substitution in (86) yields

$$y = \frac{(N-1)^{1/2} \frac{R+x}{\sigma_1^{1/2}}}{(1+2R+2R^2)};$$  \hspace{1em} (88)

that is, if normalized crosscorrelator output $h$ were Gaussian, the operating characteristics would be straight lines dictated by (88). These straight lines are superposed as dashed lines in figures 12-14 for $N=96, 128, 256$ respectively. Despite the large value of $N=96$ in figure 12, the Gaussian approximation is not that accurate, especially for small false alarm probabilities and large detection probabilities. The exact curve (solid) and the Gaussian approximation (dashed) cross each other, and are labelled at the crossing with the corresponding value of $n$ in signal-to-noise ratio $R=2^n$. For $N=256$ in figure 14, agreement is better and the Gaussian approximation is probably adequate for larger $N$. If not, an additional term or two in an Edgeworth expansion could be investigated with the aid of the cumulants given in (58).
An obvious shortcoming of the Gaussian approximation (88) may be seen immediately:

$$\lim_{R \to \infty} y = \left( \frac{N-1}{2} \right)^{1/2} < 1$$

for any \( x \). \hspace{1cm} (89)

Reference to (85) then yields the interpretation

$$\lim_{R \to \infty} P_D = \Phi \left( \frac{N-1}{2} \right)^{1/2} < 1$$

for any \( P_F \). \hspace{1cm} (90)

That is, as input signal-to-noise ratio \( R \) tends to infinity, the approximate detection probability saturates at a value less than 1, regardless of the false alarm probability. Thus the Gaussian approximation must certainly deteriorate for large \( R \); the exact discrepancy for probabilities of practical interest is displayed in figures 12-14.
ANALYTIC RESULTS FOR $\gamma=0$, SAMPLE MEAN NOT REMOVED

In this section and the next, attention will be confined solely to the case of scale factor $\gamma=0$. The characteristic function of the crosscorrelator output $q$ is then given in (31) and (32), and the mean and variance of $q$ are listed in (33). Due to the complexity of the exponential term in characteristic function (31), there are no general probability density function or cumulative distribution function results for arbitrary $N$, like those given earlier in (40), (41), and (44) for $\gamma=1$. Here, we can have $N \geq 1$.

SPECIALIZATION TO THE SIGNAL AND NOISE MODEL

For the model presented earlier in (3)-(8), the original parameters in (24) take the form already given in (37), but now with $\gamma=0$. Specifically, characteristic function (31) is

$$f_q(\xi) = \left(1 - i\xi E_1 + \xi^2 E_2\right)^{-N/2} \exp\left[i\xi \left(\frac{G_1^{(0)} + i\xi G_2^{(0)}}{1-i\xi E_1 + \xi^2 E_2}\right)\right],$$

(91)

where

$$E_1 = 2\rho_s(S_u S_v)^{1/2}, \quad E_2 = D_u D_v + D_u S_v + D_v S_u + S_u S_v (1-\rho_s^2),$$

$$G_1^{(0)} = N\mu_u \mu_v, \quad G_2^{(0)} = \frac{1}{2} N[(S_u + D_u)\mu_v^2 + (S_v + D_v)\mu_u^2 - 2\rho_s (S_u S_v)^{1/2}\mu_u \mu_v].$$

(92)

The mean and variance of crosscorrelator output $q$ follow from (38) according to

$$\mu_q = N[\rho_s (S_u S_v)^{1/2} + \mu_u \mu_v],$$

$$\sigma_q^2 = N[D_u D_v + D_u S_v + D_v S_u + (1+\rho_s^2) S_u S_v +$$

$$(S_u + D_u)\mu_v^2 + (S_v + D_v)\mu_u^2 + 2\rho_s (S_u S_v)^{1/2}\mu_u \mu_v].$$

(93)
NORMALIZED CROSSCORRELATOR OUTPUT

As in (49), and for the same reasons, we define a normalized crosscorrelator output, relative to the additive random noise disturbances \( u_d(n) \) and \( v_d(n) \) in (3) and (5), according to

\[
h = \frac{q}{D_u D_v^{1/2}}. \tag{94}
\]

The characteristic function of \( h \) is available from (91) and (92):

\[
f_h(x) = f_q \left( \frac{x}{(D_u D_v)^{1/2}} \right) = \\
\left[ 1 - i2a + i^2(\beta^2 - \alpha^2) \right]^{-N/2} \exp \left[ i \frac{\alpha + ib \beta}{1 - i2a + i^2(\beta^2 - \alpha^2)} \right], \tag{95}
\]

where

\[
\alpha = \rho_s (R_u R_v)^{1/2}, \quad \beta = [(1+R_u)(1+R_v)]^{1/2},
\]

\[
a = N r_u r_v, \quad b = \frac{1}{2} N [(1+R_u)r_v^2 + (1+R_v)r_u^2 - 2\rho_s (R_u R_v)^{1/2} r_u r_v], \tag{96}
\]

and where we have defined

\[
R_u = \frac{S_u}{D_u}, \quad R_v = \frac{S_v}{D_v}, \quad r_u = \frac{\mu_u}{D_u^{1/2}}, \quad r_v = \frac{\mu_v}{D_v^{1/2}}. \tag{97}
\]

The characteristic function in (95) depends on six fundamental parameters, namely \( N, \rho_s, R_u, R_v, r_u, r_v \). The mean and variance of \( h \) follow from (94), (93), and (97):
\[ \mu_h = N[\rho_s (R_u R_v)^{1/2} + r_u r_v] , \]

\[ \sigma_h^2 = N[(1+R_u)(1+R_v)+\rho_s^2 R_u R_v+(1+R_u)r_u^2+(1+R_v)r_v^2+2\rho_s (R_u R_v)^{1/2} r_u r_v] . \] (98)

\( r_u \) and \( r_v \) are referred to as normalized means.

**REDUCTION TO IDENTICAL SIGNAL COMPONENTS**

In order to prepare for numerical evaluation of the operating characteristics of the crosscorrelator with \( \gamma = 0 \), we further specialize the signal and noise model to the case where

\[ R_u = R_v = R, \ \rho_s = 1, \ r_u = r_v = r ; \] (99)

see (56) et seq. This leads to

\[ \alpha = R, \ \beta = 1+R, \ a = b = Nr^2, \]

via (96). The characteristic function in (95) then reduces to

\[ f_h(\xi) = [(1+i\xi)(1-i\xi(1+2R))]^{-N/2} \exp\left[\frac{i\xi Nr^2}{1-i\xi(1+2R)}\right], \] (100)

and the mean and variance in (98) become

\[ \mu_h = N(R+r^2), \ \sigma_h^2 = N[1+2R+2R^2+2(1+2R)r^2] . \] (101)

The characteristic function in (100) has a branch point at \( \xi = i \), and another branch point at \( \xi = -i/(1+2R) \) which overlaps an essential singularity; this complicates some of the analytical development to follow.

There are three fundamental parameters in (100), namely \( N, R, r \). Since normalized mean \( r \) appears only through its square, we can presume \( r \geq 0 \) without loss of generality. Furthermore, if \( r = 0 \), characteristic function (100)
reduces to (57) if \(N-1\) there is replaced by \(N\). Thus the curves for \(r=0\) here can be obtained from the earlier curves for \(\gamma=1\) in figures 1-14 by looking at a value for \(N\) there which is one greater; accordingly we can confine attention to \(r>0\) in this and the next section.

When the random signal components \(u_s(n), v_s(n)\) in model (3) are absent, then \(R=0\), and the exceedance distribution function corresponding to characteristic function (100) becomes the false alarm probability. However, (100) still depends on \(r\), meaning that the false alarm probability must, likewise. Thus, a non-zero mean in (3), i.e. \(r>0\), is not considered a signal attribute here, but rather is a nuisance quantity; it may, in fact, degrade the operating characteristics if not removed.

For notational convenience in the following, we define

\[
\omega = 1 + 2R.
\]  

(102)

The magnitude of the exponential term in characteristic function (100) then can be expressed as

\[
\exp\left[-\frac{\varphi^2 \omega N r^2}{1 + \varphi^2 \omega^2}\right],
\]  

(103)

which is monotonically decreasing for \(\varphi \geq 0\). Coupled with the observation immediately under (39), it is seen that \(|f_h(\varphi)|\) in (100) is monotonically decreasing for all \(\varphi \geq 0\) and any \(N,R,r\). This property allows for a convenient termination procedure in the numerical transformation [3] of characteristic function (100). It should however be observed that (103) does not decrease to zero, but saturates at value \(\exp(-Nr^2/2)\), regardless of how \(\varphi\) increases to infinity; thus the eventual decay of the characteristic function (100) is furnished only by the leading factor.
ASYMPTOTIC BEHAVIOR OF CUMULATIVE AND EXCEEDANCE DISTRIBUTION FUNCTIONS

In appendix E, it is shown that if characteristic function (100) is substituted in (63) and (64), and the contours moved appropriately in the complex \( \xi \)-plane, then the following asymptotic behaviors obtain. The cumulative distribution function

\[
P_h(u) \sim \left[ \frac{N}{2} \right]^{-1} \left( -u \right)^{N/2} \exp \left[ u - \frac{N \sigma^2}{2(1+R)} \right] \times
\]

\[
times \left[ 1 - \frac{N}{2u} \left( 1 + \frac{N(1+2R)}{4(1+R)} + \frac{N \sigma^2}{4(1+R)^2} \right) \right] \quad \text{as } u \to -\infty. \tag{104}
\]

We see again, in similar fashion to (61) and (62), that in order for the correction term in the second line of (104) not to be too significant, we must have \( u < -N^2 \). For reasons elucidated in (61) et seq., (104) is not useful for large \( N \).

As checks on (104), we note that for \( r=0 \) and \( N=2 \), (104) reduces precisely to the upper line of (71); this latter result pertains to \( \gamma=1, N=3 \) and is consistent with the observation already made in the paragraph below (101). In addition, if we let \( r=0 \) and \( N=4 \) in (104), it reduces to the upper line of (75); this latter result holds for \( \gamma=1, N=5 \) and is likewise consistent.

Also given in appendix E are a variety of asymptotic expansions for the exceedance distribution function; the simplest one is

\[
1 - P_h(u) \sim \left[ 2^{1/2} \left( \frac{N}{2} \right)^{N/2} (1+R)^{N/2} \left( \frac{N \sigma^2}{4} \right)^{N-1} \right]^{-1} \left( \frac{u^{1/2} - N^{1/2} \sigma}{1+2R} \right)^2 \quad \text{as } u \to +\infty; \quad r>0. \tag{105}
\]
However, the same reservations as above, regarding \( u \) large relative to \( N^2 \), are again in order.

**DISTRIBUTIONS FOR \( N=1 \)**

If characteristic function (100) with \( N=1 \) is substituted in (63), and if the contour is moved as indicated under (66), there follows the exact result for the cumulative distribution function

\[
P_h(u) = \frac{2}{\pi} \int_0^\infty \frac{dt}{(1+t^2)(1+\omega t^2)} \exp \left[ u(1+t^2) - \frac{r^2(1+t^2)}{1+\omega(1+t^2)} \right] \text{ for } u<0. \quad (106)
\]

(This reduces to (67) for \( r=0 \), as it must.) This integral form possesses all the desirable attributes listed under (67).

If characteristic function (100) with \( N=1 \) is substituted in (64) in an attempt to get the exceedance distribution function, the analysis becomes rather difficult, due to the overlapping essential singularity and branch point of the integrand at \( \xi = -i/\omega \). This problem is treated in detail in appendix F, with the result that the exceedance distribution function can be found via the characteristic function approach in terms of two integrals; see (F-21)-(F-23). However, a better numerical procedure for the exceedance distribution function is the direct result derived in (F-33); this latter integral is the one actually used here to generate the operating characteristics for \( \gamma=0, N=1 \).

**DISTRIBUTIONS FOR \( N=2 \)**

The characteristic function is available from (100):

\[
f_h(\xi) = (1+i\xi)^{-1}(1-i\xi)^{-1} \exp \left( i\xi^2r^2 \right) \equiv f_1(\xi)f_2(\xi), \quad (107)
\]
where \( \omega = 1 + 2R \). The probability density functions corresponding to these two characteristic functions are [5, 6.631 4]

\[
p_1(u) = \exp(u) \, U(-u),
\]

\[
p_2(u) = \frac{1}{\omega} \exp\left(-\frac{u+2r^2}{\omega}\right) I_0\left(\frac{2r}{\omega} (2u)^{1/2}\right) \, U(u),
\]  

(108)

where \( U \) is the unit step function

\[
U(u) = \begin{cases} 
0 & \text{for } u < 0 \\
1 & \text{for } u > 0 
\end{cases}.
\]  

(109)

The probability density function of \( h \) is given by convolution

\[
p_h(u) = \int_{-\infty}^{+\infty} dt \, p_1(t) \, p_2(u-t) \quad \text{for all } u.
\]  

(110)

Substitution of (108) in (110) yields

\[
p_h(u) = \begin{cases} 
\frac{1}{1+\omega} \exp\left(u - \frac{2r^2}{1+\omega}\right) & \text{for } u \leq 0 \\
\frac{1}{1+\omega} \exp\left(u - \frac{2r^2}{1+\omega}\right) Q\left(\frac{2r}{\omega} \left(\frac{2(1+\omega)}{1+\omega}\right)^{1/2}, \left(\frac{2(1+\omega)}{\omega}\right)^{1/2}\right) & \text{for } u > 0
\end{cases}
\]

(111)

where the \( Q \)-function is defined in [7] and the two integrals encountered have been evaluated by use of [5, 6.631 4] and [7, (9)], respectively. The cumulative and exceedance distribution functions of \( h \) readily follow from (111), the latter by means of [7, (42)]:

39
The characteristic function is available from (100):

\[ f_h(r) = (1+i\xi)^{-2} \cdot (1-i\xi\omega)^{-2} \exp\left(\frac{i\xi 4 r^2}{1-i\xi \omega}\right) \equiv f_1(r) \ast f_2(r) , \quad (113) \]

where \( \omega = 1+2r \). The probability density functions corresponding to these two characteristic functions are \([5, 6.631 4]\)

\[ p_1(u) = -u \exp(u) \ U(-u) , \]

\[ p_2(u) = \frac{1}{2\omega r} \exp\left(\frac{u+4r^2}{\omega}\right) u^{1/2} \ I_1 \left(4ru^{1/2}/\omega\right) \ U(u) . \quad (114) \]

The probability density function of \( h \) is given by convolution (110). In preparation for that result, we use the shorthand notation

\[ q_M(u) = Q_M \left(2r \left(\frac{2}{\omega (1+\omega)}\right)^{1/2} \cdot \left(\frac{2u(1+\omega)}{\omega}\right)^{1/2}\right) , \]

\[ \tilde{q}_M(u) = Q_M \left(2r \left(\frac{2}{\omega}\right)^{1/2} \cdot \left(\frac{2u}{\omega}\right)^{1/2}\right) . \quad (115) \]
where the $Q_M$-function is defined in [8]. We also present a new integral result that will be needed in the sequel,

$$
\int_a^b dx \times Q_M(b, ax) = \frac{1}{2a^2} [b^2Q_{M+1}(b, ac) + 2M Q_M(b, ac) - a^2c^2Q_{M-1}(b, ac)], \quad (116)
$$

which can be interpreted as the limit of [8, (31)] as $p \to 0^+$. Substitution of (114) in (110) yields probability density function

$$
p_h(u) = \begin{cases} 
\frac{1}{(1+\omega)^4} \exp \left( u - \frac{4r^2}{1+\omega} \left[ 4r^2 + 2\omega(1+\omega) - (1+\omega)^2u \right] \right) & \text{for } u \leq 0 \\
\frac{1}{(1+\omega)^4} \exp \left( u - \frac{4r^2}{1+\omega} \left[ 4\omega^2 + 2\omega(1+\omega)q_3(u) + 2\omega q_2(u) - (1+\omega)^2uq_1(u) \right] \right) & \text{for } u > 0
\end{cases} \quad (117)
$$

where (115) has been used; the upper line employed [5, 6.31 4], while the lower line used (116) and an integration by parts procedure to be elaborated below in the exceedance distribution function evaluation.

The cumulative distribution function for $u < 0$ follows readily by integration of (117):

$$
P_h(u) = \frac{1}{(1+\omega)^4} \exp \left( u - \frac{4r^2}{1+\omega} \left[ 4r^2 + 2\omega(1+3\omega) - (1+\omega)^2u \right] \right) \quad \text{for } u \leq 0. \quad (118)
$$

For $u > 0$, a modified approach is required. Integration of (110) yields cumulative distribution function

$$
P_h(u) = \int_{-\infty}^{+\infty} dt \ p_2(t) \ p_1(u-t) . \quad (119)
$$

Probability density functions $p_1$ and $p_2$ are available in (114), and cumulative distribution function $P_1$ follows as
\[ P_1(u) = \begin{cases} (1-u)\exp(u) & \text{for } u < 0 \\ 1 & \text{for } u \geq 0 \end{cases} \]  

(120)

Substitution of (114) and (120) in (119) yields, for \( u \geq 0 \),

\[
P_h(u) = \int_0^u dt \frac{1}{2\omega r} \exp\left(-\frac{t+4r^2}{\omega}\right) t^{1/2} I_1\left(4\omega t^{1/2}/\omega\right) +
\]

\[
+ \int_u^{+\infty} dt \frac{1}{2\omega r} \exp\left(-\frac{t+4r^2}{\omega}\right) t^{1/2} I_1\left(4\omega t^{1/2}/\omega\right) (1-u+t) \exp(u-t) = \]

\[= 1-\tilde{q}_2(u) + \frac{1}{2\omega r} \exp\left(\frac{4r^2}{\omega}\right) \int_u^{+\infty} dt t^{1/2} \exp\left(-t \frac{1+\omega}{\omega}\right) I_1\left(4\omega t^{1/2}/\omega\right) (1-u+t), \quad \text{(121)} \]

data via [8, (22)]. We now integrate by parts, letting \( U(t) = 1-u+t \), and the remainder \( dV(t) \). Then using [8, (22)] again, we find

\[ V(t) = -\frac{2\omega r}{(1+\omega)^2} \exp\left(\frac{4r^2}{\omega(1+\omega)}\right) q_2(t) \]  

(122)

Combining these results, (121) and (116) yield

\[ 1 - P_h(u) = \tilde{q}_2(u) - \frac{1}{(1+\omega)^2} \exp\left(u - \frac{4r^2}{1+\omega}\right) q_2(u) - \int_u^{+\infty} dt q_2(t) = \]

\[= \tilde{q}_2(u) - \frac{1}{(1+\omega)^4} \exp\left(u - \frac{4r^2}{1+\omega}\right) \left[4r^2 q_3(u)+(1+\omega)(1+3\omega)q_2(u)-(1+\omega)^2 uq_1(u)\right] \]

for \( u \geq 0 \).  

(123)
The final results for N=4 are given by (115), (118), and (123). This case was used as a numerical check on the computational approach [3] proceeding directly from characteristic function (100) to the exceedance distribution function, with excellent agreement for numerous values of R, r, and u.
SUMMARY OF PARTICULAR CASE CONSIDERED

The situation of interest has already been summarized in (76)-(78); in addition, we have a common value for the normalized means,

\[
\frac{\mu_u}{D_u^{1/2}} = \frac{\mu_v}{D_v^{1/2}} = r ,
\]

and the normalized crosscorrelator output is not (79)-(80), but rather is, for \(\gamma=0\),

\[
h = \frac{1}{\sqrt{(D_u D_v)}} \sum_{n=1}^{N} u_n v_n .
\]

The characteristic function is given by (100):

\[
f_h(\xi) = \left[ (1+i\xi)(1-i\xi) \right]^{-N/2} \exp \left( \frac{i\xi N r^2}{1-i\xi \omega} \right) ,
\]

where \(\omega = 1+2R\). When the signal is absent, then \(R=0\); however the false alarm probability corresponding to this characteristic function still depends on \(r\). Thus since the sample mean has not been removed, the operating characteristics will also depend on \(r\). Since results for \(r=0\) can be found from an earlier section, we only consider \(r>0\) here.

OPERATING CHARACTERISTICS FOR \(\gamma=0\)

A sample program for evaluation of the cumulative and exceedance distribution functions corresponding to characteristic function (126), and thereby the detection probability vs. false alarm probability operating characteristics of the crosscorrelator without sample mean removal, is given in appendix G. In figures 15-35 are presented the operating characteristics for values of \(\gamma\). See page 65 et seq.
\[ N = 1, 2, 3, 4, 8, 16, 32, 64, 128, 256, \] \hspace{1cm} (127)

and for various values of \( r \). The case of \( N=1 \) was accomplished by use of (106) and (F-33); results for \( N=2 \) employed (112); and the remainder for \( N>3 \) employed a numerical procedure [3] proceeding directly from characteristic function (126) to the exceedance distribution function. False alarm probabilities \( P_F \) in the range 1E-10 to .5 and detection probabilities \( P_D \) covering 1E-10 to .999 are presented. The abscissa and ordinate on these plots employ a normal probability transformation, as explained earlier in (84)-(86).

Values of signal-to-noise ratio \( R \) are taken as \( R=2^n \), where \( n \) assumes values appropriate for each plot in order to cover the full range of probabilities of interest. Values of normalized mean \( r \) in (124) have been taken as \( r=1 \) and 2, with the exception of figure 19 where one example for \( r=4 \) was added.

Without exception, increasing \( r \) from zero degrades the operating characteristics of the crosscorrelator. For example, figures 17, 18, 19 give a succession of operating characteristics for \( r = 1, 2, 4 \) respectively, and for common values of signal-to-noise ratio \( R \). (In order to determine the operating characteristics for \( r=0 \) here, we can look at the earlier results in figures 1-14, but for a value of \( N \) which is one greater there.) Thus, not removing the sample mean from the crosscorrelator output requires a larger threshold setting for a specified false alarm probability and thereby lowers the detection probability and degrades performance.

GAUSSIAN APPROXIMATION

If the crosscorrelator output is Gaussian, for both signal absent as well as present, the earlier derivations in (83)-(86) pertain. Now reference to (101) yields statistics

\[ m_0 = Nr^2, \quad m_1 = N(R+r^2), \]

\[ \sigma_0^2 = N(1+2r^2), \quad \sigma_1^2 = N[1+2R+2R^2+2(1+2R)r^2], \] \hspace{1cm} (128)
since setting signal-to-noise ratio $R=0$ corresponds to hypothesis 0, signal absent. Substitution in (86) yields the normal probability approximation

$$y = \frac{N^{1/2} R^{1/2} (1+2r^2)^{1/2}}{[1+2R+2R^2+2(1+2R)r^2]^{1/2}}. \quad (129)$$

These straight lines are superposed as dashed lines in figures 32-35 for $N=128$ and 256. The Gaussian approximation is moderately good for large $N$ such as 256, and in fact crosses the exact curves (solid) at a point which is labeled with the corresponding value of $n$ in signal-to-noise ratio $R=2^n$.

An obvious shortcoming of the Gaussian approximation (129) is apparent:

$$\lim_{R \to \infty} y = \left(\frac{N}{2}\right)^{1/2} \text{ for any } x, r. \quad (130)$$

Reference to (85) then yields the interpretation

$$\lim_{R \to \infty} P_D = \Phi\left(\left(\frac{N}{2}\right)^{1/2}\right) < 1 \text{ for any } P_F, r. \quad (131)$$

That is, as input signal-to-noise ratio $R$ tends to infinity, the approximate detection probability saturates at a value less than 1, regardless of the false alarm probability and normalized mean $r$. Thus the Gaussian approximation must certainly be inaccurate for large $R$; the exact discrepancy for probabilities of practical interest is displayed in figures 32-35.
SUMMARY

A closed form expression for the characteristic function of the output of a crosscorrelator, with or without sample mean removal, has been derived in (23)-(24) for general values of: the number of terms added to yield the correlator output, the means and variances in each of the two input channels, the degree of correlation between the two channels, and the scale factor employed in the sample mean removal. A program for the evaluation of the cumulative and exceedance distribution functions of this general case has also been presented. These results can furnish the basis of a study of the error probabilities of a correlator required to decide between alternative hypotheses on the input statistics [1]; this problem will in fact be the subject of a future technical report by this author.

The general results were first specialized to a signal and noise model, and then to the two distinct cases of sample mean removal ($Y=1$) and no sample mean removal ($Y=0$). Plots of the operating characteristics for numerous values of $N$ and signal-to-noise ratio $R$ were then displayed for a wide range of detection probability vs false alarm probability. Some new analytic results for cumulative and exceedance distribution functions, especially for small $N$, were derived and used as checks on the general numerical procedure. Comparisons with a Gaussian approximation indicated quantitatively when that simplification is valid. Asymptotic results derived were useful for small $N$, but not for large $N$ except in the region of probabilities too small to be of practical importance.
Figure 1. Operating Characteristics for Cross-Correlator with Sample Mean Removal, $N = 2$
Figure 2. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 3
Figure 3. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 4
Figure 4. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 6
Figure 5. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 8
Figure 6. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 12
Figure 7. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 16
Figure 8. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 24
Figure 9. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 32
Figure 10. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 48
Figure 11. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 64
Figure 12. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 96
Figure 13. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 128
Figure 14. Operating Characteristics for Cross-Correlator with Sample Mean Removal, N = 256
Figure 15. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=1, r=1
Figure 16. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=1, r=2
Figure 17. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=2, r=1
Figure 18. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=2, r=2
Figure 19. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=2, r=4
Figure 20. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=3, r=1
Figure 21. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=3, r=2
Figure 22. Operating Characteristics for Cross-Correlator without Sample Mean Removal; N=4, r=1
Figure 23. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=4, r=2
Figure 24. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=8, r=1
Figure 25. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=8, r=2
Figure 26. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=16, r=1
Figure 27. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=16, r=2
Figure 28. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=32, r=1
Figure 29. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=32, r=2
Figure 30. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=64, r=1
Figure 31. Operating Characteristics for Cross-Correlator without Sample Mean Removal, $N=64$, $r=2$
Figure 32. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=128, r=1
Figure 33. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=128, r=2
Figure 34. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=256, r=1
Figure 35. Operating Characteristics for Cross-Correlator without Sample Mean Removal, N=256, r=2
APPENDIX A. CORRELATOR OUTPUT INDEPENDENCE OF MEANS

If we let scale factors $a=1$ but $\beta \neq 1$ in (11), we still get $\gamma = 1$ from (13). This means that correlator output $q$ in (12) is independent of $\mu_u$, $\mu_v$, $\beta$.

To see this directly, let

$$u_n = u_n + y_n, \quad v_n = \mu_v + z_n,$$  \hspace{1cm} (A-1)

where means

$$\bar{y}_n - \bar{z}_n = 0.$$  \hspace{1cm} (A-2)

Then (11) yields, with $a=1$,

$$\tilde{u}_n = y_n - \frac{1}{N} \sum_{m=1}^{N} y_m,$$  \hspace{1cm} (A-3)

which is obviously independent of the actual value of mean $\mu_u$. Also, (11) yields

$$\tilde{v}_n = z_n - \frac{\beta}{N} \sum_{m=1}^{N} z_m + \mu_v(1-\beta),$$  \hspace{1cm} (A-4)

which still depends on $\mu_v$ and $\beta$. When (A-4) is substituted in (12), we get correlator output

$$q = \sum_{n=1}^{N} \tilde{u}_n \left[ z_n - \frac{\beta}{N} \sum_{m=1}^{N} z_m + \mu_v(1-\beta) \right].$$  \hspace{1cm} (A-5)

But since

$$\sum_{n=1}^{N} \tilde{u}_n = 0$$  \hspace{1cm} (A-6)
from (A-3), (A-5) reduces to

\[ q = \sum_{n=1}^{N} u_n z_n = \sum_{n=1}^{N} \left( y_n - \frac{1}{N} \sum_{m=1}^{N} y_m \right) z_n = \]

\[ = \sum_{n=1}^{N} y_n z_n - \frac{1}{N} \sum_{m=1}^{N} y_m \sum_{n=1}^{N} z_n \quad (A-7) \]

in terms of the ac components defined in (A-1) and (A-2). Correlator output (A-7) is obviously independent of means \( \mu_u, \mu_v \) and scale factor \( \beta \) in (11).
APPENDIX B. A USEFUL INTEGRAL OF EXPONENTIALS OF MATRIX FORMS

For symmetric \( K \times K \) matrix \( M \), with \( \det M > 0 \), the following \( K \)-fold integral is well known (see for example, [9, section 8-3]):

\[
\int dX \exp\left[-\frac{1}{2}X^T MX + N^T X\right] = \left(\frac{(2\pi)^K}{\det M}\right)^{1/2} \exp\left[\frac{1}{2} N^T M^{-1} N\right]. \tag{B-1}
\]

Here \( X \) and \( N \) are \( K \times 1 \) column matrices. We wish to extend this result to the case of double integral

\[
I = (2\pi)^{-K} \int \int dU \, dV \exp\left[-\frac{1}{2} U^T AU - \frac{1}{2} V^T BV + U^T CV + D^T U + E^T V\right], \tag{B-2}
\]

where \( A \) and \( B \) are symmetric without loss of generality, and the integral converges; here, matrices \( A, B, C \) are \( K \times K \) while \( U, V, D, E \) are \( K \times 1 \). Notice that if we had the apparently more-general term

\[
U^T C_1 V + V^T C_2 U = U^T (C_1 + C_2^T) V, \tag{B-3}
\]

we would simply let \( C = C_1 + C_2^T \), and thereby immediately have form (B-2).

To accomplish the evaluation of \( I \) in (B-2), identify \( M = B \), \( N^T = U^T C + E^T \), \( X = V \) in (B-1), and thereby evaluate the \( V \)-integral in (B-2), with result

\[
\int dV \ldots = \left(\frac{(2\pi)^K}{\det B}\right)^{1/2} \exp\left[\frac{1}{2}(U^T C + E^T)B^{-1}(C^T U + E)\right]. \tag{B-4}
\]

Substituting (B-4) in (B-2), regrouping, and using the symmetry of \( B \) (and therefore \( B^{-1} \)), there follows

\[
I = \frac{(2\pi)^{-K/2}}{(\det B)^{1/2}} \int dU \exp\left[-\frac{1}{2} U^T (A-CB^{-1}C^T)U + (D+E^T B^{-1}C^T)U + \frac{1}{2} E^T B^{-1}E\right]. \tag{B-5}
\]
Now reemploying (B-1) with identifications \( M = A - CB^{-1}C^T \), \( N = D + CB^{-1}E \), \( X = U \), we get a closed form result for (B-5):

\[
I = \left[ \det(AB - CB^{-1}C^T B) \right]^{-1/2} \exp \left[ \frac{1}{2} E B^{-1} E + \frac{1}{2} (D + CB^{-1} E)^T (A - CB^{-1} C^T)^{-1} (D + CB^{-1} E) \right].
\]  
\text{(B-6)}

This is the desired general result for integral (B-2).

As an aside, there is probably a more symmetric closed form result than (B-6), since if we represent (B-2) by \( I(A, B, C, D, E) \), we quickly see, by interchange of dummy variables \( U \) and \( V \), that

\[
I(A, B, C, D, E) = I(B, A, C^T, E, D).
\]  
\text{(B-7)}

However, we have not discovered the symmetric form of (B-6). The present form follows as a result of the sequential integration of (B-2), first on \( V \), then on \( U \).
APPENDIX C. PROGRAM FOR CUMULATIVE AND EXCEEDANCE DISTRIBUTION FUNCTIONS VIA CHARACTERISTIC FUNCTION (23)-(24).

The numerical procedure employed in appendices C, D, and G here is heavily based on [3]. The choices of L, Δ, b in lines 90 to 110 to control truncation error and aliasing are also made according to the method of [3]. The parameters in (24) are evaluated once in lines 210-260 so as to minimize computation time. The FFT subroutine used in lines 1030 et seq is listed in [3, pp. B-11 - B-12], and employs a zero-subscribed array. A sample plot of the cumulative and exceedance distribution functions follows the program.

10 ! CROSS-CORRELATOR WITH SAMPLE MEAN REMOVAL; NUSC TR 7045
20 N=5 ! Number of terms summed to yield output
30 Gamma=.5 ! Scale factor in sample mean removal
40 Mu=.4 ! U channel mean
50 Mu=.3 ! V channel mean
60 Su=.7 ! U channel standard deviation
70 Sv=.9 ! V channel standard deviation
80 Rho=.6 ! Correlation coefficient
90 L=60 ! Limit on integral of char. function
100 Delta=.12 ! Sampling increment on char. function
110 Bs=.25*(2*PI/Delta) ! Shift b, as fraction of alias interval
120 Mf=2^8 ! Size of FFT
130 PRINTER IS 0
140 PRINT "L =";L,"Delta =";Delta,"b =";Bs,"Mf =";Mf
150 REDIM X(0:Mf-1),Y(0:Mf-1)
160 DIM XX(0:1023),YY(0:1023)
170 Su2=Su*Su ! Calculate parameters
180 Su2=Su*Su
190 T1=1-Gamma
200 T2=T1*T1
210 E1=2*Rho*Su*Su
220 E2=Su2*Su2*(1-Rho*Rho)
230 F1=E1*T1
240 F2=E2*T2
250 G1=H*Mu*Mu+T1
260 G2=.5*h*(Su2*Mu+Mu+Su2*Mu+Mu-E1*Mu*Mu)*T2
270 N1=.5*(N-1)
280 Muq=G1+(N-Gamma)*Rho*Su+Su ! Mean of random variable q
290 Muv=Muq+Bs ! Mean of shifted variable y
300 X(0)=0
310 Y(0)=.5*Delta+Muv
320 FOR Ns=1 TO INT(L/Delta)
330 Xi=Delta*Ns ! Argument xi of char. function
340 X2=Xi*Xs
350 T1=-Xi*F1 ! Calculation of
360 T2=1+X2*F2 ! characteristic function
370 CALL Div(-X2*G2,Xi=G1,T2,T1,A,B) ! function
380 CALL Log(1+X2*E2,-Xi*E1,C,D) ! fy(x)
390 CALL Log(T2,Ti,E,F)
400 CALL Exp(A-N1+C-.5*E,B-N1*D-.5*F+Xi*Bs,Fy,Fy)

C-1
TR 7045

410 Ms=Ns MOD Mf
420 \(X(Ms)=X(Ms)+Fyr/Ns\)
430 \(Y(Ms)=Y(Ms)+Fyr/Ns\)
440 NEXT Ms
450 CALL FFT10=Mf,\(X(*)\),\(Y(*)\))
460 PLOTTER IS "GRAPHICS"
470 GRAPHICS
480 SCALE 0,Mf,-14,0
490 LINE TYPE 3
500 GRID Mf/3,1
510 PEND
520 LINE TYPE 1
530 \(B=Bs*Mf*Delta/(2*PI)\)
540 MOVE B,0
550 DRAW B,-14
560 PEND
570 FOR \(Ks=0\) TO Mf-1
580 \(T=Y(Ks)/PI-Ks/Mf\)
590 \(X(Ks)=-5-T\)
600 \(Y(Ks)=Pr-.5+T\)
610 IF \(Pr>1E-12\) THEN Y=LGT(Pr)
620 IF \(Pr<-1E-12\) THEN \(Y=-24-LGT(-Pr)\)
630 IF \(ABS(Pr)>1E-12\) THEN \(Y=-12\)
640 PLOT Ks,Y
650 NEXT Ks
660 PEND
670 PRINT Y(0);Y(1);Y(Mf-2);Y(Mf-1)
680 FOR \(Ks=0\) TO Mf-1
690 \(Pr=X(Ks)\)
700 IF \(Pr>1E-12\) THEN Y=LGT(Pr)
710 IF \(Pr<-1E-12\) THEN \(Y=-24-LGT(-Pr)\)
720 IF \(ABS(Pr)>1E-12\) THEN \(Y=-12\)
730 PLOT Ks,Y
740 NEXT Ks
750 PEND
760 PAUSE
770 DUMP GRAPHICS
780 PRINT LIN(5)
790 PRINTER IS 16
800 END
810 !
820 SUB Div(X1,Y1,X2,Y2,A,B)
830 \(T=X2*X2+Y2*Y2\)
840 \(A=(X1*X2+Y1*Y2)/T\)
850 \(B=(Y1*X2-X1*Y2)/T\)
860 SUBEND
870 !
880 SUB Log(X,Y,A,B)
890 \(A=.5*LOG(X*X+Y*Y)\)
900 IF \(XX0\) THEN 930
910 \(B=.5*PI*SGN(Y)\)
920 GOTO 950
930 \(B=ATN(Y/X)\)
940 IF \(XX0\) THEN \(B=B+PI*(1-2*(YY0))\)
950 SUBEND
960 !
970  SUB Exp(X,Y,A,B)
980  T=EXP(X)
990  A=T*COS(Y)
1000  B=T*SIN(Y)
1010  SUBEND
1020  !
1030  SUB FFT10Z(N,X(*),Y(*))  ! N <= 2^10 = 1024, N=2^INTEGER  Ø SUBSCRIPT
Figure C-1. Cumulative and Exceedance Distribution Functions
APPENDIX D. PROGRAM FOR EVALUATION OF OPERATING CHARACTERISTICS FOR $\gamma=1$

The comments in appendix C are relevant here also. The characteristic function used as the starting point is given by (81). Sampling increment $\Delta$ employed on the characteristic function can be coarse for small signal-to-noise ratio $R$, but must be finer for larger $R$. The quantity $\Delta_0$, $\Delta_0$, in line 30 is that used for $R=0$; all other $\Delta$ values are sub-multiples, as indicated by lines 110-130. A table follows.

<table>
<thead>
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<th>N</th>
<th>4</th>
<th>6</th>
<th>8</th>
<th>12</th>
<th>16</th>
<th>24</th>
<th>32</th>
<th>48</th>
<th>64</th>
<th>96</th>
<th>128</th>
<th>256</th>
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</thead>
<tbody>
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<td>.10</td>
<td>.09</td>
<td>.08</td>
<td>.06</td>
<td>.05</td>
<td>.05</td>
<td>.05</td>
<td>.04</td>
<td>.03</td>
<td>.03</td>
<td>.02</td>
<td></td>
</tr>
</tbody>
</table>

Table D-1. Values of $\Delta_0$ for $\gamma=1$

Let $\Delta_1$ denote the sampling increment employed for a particular value of signal-to-noise ratio $R_1>0$. The cumulative and exceedance distribution function values are available at spacing $s = 2\pi/(M_f\Delta)$ in general, where $M_f$ is the FFT size. If, for example, $\Delta_1 = \Delta_0/2$, then $s_1 = 2s_0$, meaning that probability values occur twice as coarsely for $R_1$ as for $R=0$. Then in order to plot detection probability $P_D$ vs false alarm probability $P_F$ without interpolating points, it is necessary to skip every other $P_F$ point available, and only plot those $P_F$, $P_D$ pairs corresponding to the same threshold. More generally, if $\Delta_1 = \Delta_0/K$, where $K$ is an integer, then $s_1 = Ks_0$, and we plot only every $K$-th point of the available $P_F$ values. Here we have chosen $K$ to be a power of 2, for the purpose of ease of plotting.

We choose bias (shift) $b$ in line 50 in order to give a random variable $y$ which is virtually always positive for $R=0$; see [3]. We then keep $b$ fixed as $R$ increases, which makes the probability of $y>0$ even greater. This feature of choosing the same $b$ for all $R>0$ enables an easy comparison of $P_D$ and $P_F$, since common threshold values are then conveniently realized.
It was observed under (39) that the characteristic functions (39) or (81) have monotonically decreasing magnitudes for all $\xi \geq 0$. This makes the choice of $L$, the truncation value on the characteristic function integral in (63) or (64), rather simple; all we need to do is monitor $|f_h(\xi)|$ of (81) until it decreases below a tolerance, here taken as $1E^{-12}$. There is no trial-and-error procedure required as in [3] to guarantee negligible truncation error.

Subroutines Exp and Log have already been listed in appendix C, and so are not relisted here.

```plaintext
10 ! GAMMA = 1 SAMPLE MEAN REMOVAL
20 Nc=8 ! N, Number of terms summed
30 Delta0=.08 ! Initial delta
40 Bs=PI/Delta0 ! Bias b
50 Mf=2^10 ! Size of FFT
60 OUTPUT 0;"GAMMA = 1"; "N ="; Nc
70 OUTPUT 0;
80 DATA -2,-1,0,.5,1,1.5,2,2.5,3,3.5,4,5
90 READ Ns(*) ! SNR R=2^n
100 OUTPUT 0;Ns(*);
110 DATA 1,2,2,2,4,4,4,8,8,16,16,32,64
120 READ Idelta(*)
130 MAT Delta=(Delta0)/Idelta
140 OUTPUT 0;Delta(*);
150 DATA 1E-10,1E-9,1E-8,1E-7,1E-6,1E-5,1E-4,.001,.01,.1,1.5,9,99,999
160 READ Sc(*)
170 DIM Ns(1:12),Idelta(0:12),Delta(0:12),Sc(1:14)
180 DIM X(0:8191),Y(0:8191)
190 FOR I=1 TO 14
200 Sc(I)=FINvphi(Sc(I))
210 NEXT I
220 S=Sc(1)
230 B=Sc(14)
240 Scale=(B-S)/(0-S)
250 X1=30
260 X2=170
270 Y1=35
280 Y2=Y1+(X2-X1)*Scale
290 PLOTTER IS "9872A"
300 LIMIT X1,X2,Y1,Y2
310 OUTPUT 705;"VS3"
320 SCALE S,0,S,B
330 FOR I=1 TO 14
340 MOVE S,Sc(I)
350 DRAW 0,Sc(I)
360 NEXT I
370 FOR I=1 TO 11
380 MOVE Sc(I),S
390 DRAW Sc(I),B
400 NEXT I
410 MOVE S,S
420 DRAN 0,0
430 PEHUP
D-2
```
440   M1=Mf-1
450   M2=Mf/2
460   FOR In=0 TO 12
470   IF In>0 THEN 500
480   Rc=0
490   GOTO 510
500   Rc=2*Ms(In)
510   OUTPUT 0;"R =";Rc, " Delta =";Delta(In)
520   ASSIGN #1 TO "ABSCIS" ! Temporary storage for
530   Delta=Delta(In) ! false alarm probability
540   R2=Rc+2
550   R21=R2+1
560   N12=(Nc-1)/2
570   Mux=(Nc-1)*Rc
580   Muy=Mux+Bs
590   REDIM X(0:M1),Y(0:M1)
600   MAT X=2ER
610   MAT Y=2ER
620   X(0)=0
630   Y(0)=-0.5*Delta+Muy
640   Ls=0
650   Ls=Ls+1
660   Xs=Delta*Ls ! Argument x_i of char. fn.
670   CALL Log(1+Xs*Xs+R21,-Xs+R2,Ai,Bi) ! Calculation
680   CALL Exp(1+N12+R2,-R2,Ai,Bi,Fyri,Fyi) ! of
690   Ms=Ls MOD Mf
700   Ar=Fyri/Ls
710   Ai=Fyri/Ls
720   X(Ms)=X(Ms)+Ar
730   Y(Ms)=Y(Ms)+Ai
740   Magsq=Ar+Ar+Ai+Ai
750   IF Magsq>1E-24 THEN 650
760   OUTPUT 0;"Mag =";SOR(Magsq)
770   CALL FF=132(Mf,X(*)+Y(*))
780   FOR Ms=0 TO M1
790   T=Y(Ms)/PI-Ms/Mf
800   X(Ms)=5-T ! Cumulative distribution function
810   Y(Ms)=5+T ! Exceedance distribution function
820 NEXT Ms
830   OUTPUT 0;Y(0);Y(1);Y(M1-1);Y(M1)
840   PLOTTER IS "GRAPHICS"
850   GRAPHICS
860   SCALE 0,Mf,-14,0
870   LINE TYPE 3
880   GRID Mf/8,1
890   PENUMP
900   LINE TYPE 1
910   FOR Ms=0 TO M1
920   Pr=Y(Ms)
930   IF Pr>1E-12 THEN Y=LGT(Pr)
940   IF Pr<1E-12 THEN Y=-24-LGT(-Pr)
950   IF AABS(Pr)<1E-12 THEN Y=-12
960   PLOT Ms,Y
970 NEXT Ms
980 PENUMP
FOR Ms=0 TO M1
1000 Pr=X(Ms)
1010 IF Pr>1E-12 THEN Y=LGT(Pr)
1020 IF Pr<-1E-12 THEN Y=-24-LGT(-Pr)
1030 IF ABS(Pr)<1E-12 THEN Y=-12
1040 PLOT Ms,Y
1050 NEXT Ms
1060 PENUP
1070 DUMP GRAPHICS
1080 OUTPUT 0;""
1090 IF In>0 THEN 1200
1100 FOR Ms=M2 TO M1
1110 IF Y(Ms)<=0 THEN 1130
1120 NEXT Ms
1130 M3=Ms-1
1140 REDIM X(M2:M3)
1150 FOR Ms=M2 TO M3
1160 X(Ms)=FNInvphi(Y(Ms))
1170 NEXT Ms
1180 PRINT #1;X(*)
1190 GOTO 1380
1200 REDIM X(M2:M3)
1210 READ #1;X(*)
1220 Id=Idel
1230 J2=M2/Id
1240 J3=INT(M3/Id)+1
1250 FOR J=J2 TO J3
1260 Y(J)=FNInvphi(Y(J))
1270 NEXT J
1280 PLOTTER IS "$9872A"
1290 LIMIT X1,X2,Y1,Y2
1300 OUTPUT 705;"VS3"
1310 SCALE S,0,S,B
1320 FOR J=J2 TO J3
1330 T=J*I
1340 IF T>M3 THEN 1370
1350 PLOT X(T),Y(J)
1360 NEXT J
1370 PENUP
1380 NEXT In
1390 END
1400 !
1410 SUB Exp(X,Y,A,B)
1420 !
1430 SUB Log(X,Y,A,B)
1440 !
1450 !
1460 !
1470 !
1480 !
1490 !
1500 !
1510 !
1520 !
1530 !
1540 !
1550 !
1560 DEF FNInvphi(X)
1570 !
1580 !
1590 !
1600 !
1610 !
1620 !
1630 !
1640 !
1650 !
1660 !
1670 !
1680 !
1690 !
1700 SUB Fft13z(N,X(*),Y(*))
1710 !
APPENDIX E. ASYMPTOTIC EXPANSIONS FOR DISTRIBUTIONS WHEN r>0

The characteristic function of interest is given by (100) and (102):

\[ f_n(\xi) = (1+i\xi)^{-v} (1-i\xi \omega)^{-v} \exp\left[\frac{i\xi n}{1-i\xi \omega}\right], \]  

(E-1)

where for notational convenience in this appendix, we let

\[ \nu = \frac{N}{2}, \quad \eta = Nr^2, \quad \omega = 1+2R. \]  

(E-2)

The cumulative distribution function is obtained by substitution of (E-1) in (63):

\[ P_h(u) = \frac{-1}{12\pi} \int_{C_+} d\xi \xi^{-1} (1+i\xi)^{-\nu} (1-i\xi \omega)^{-\nu} \exp\left[\frac{i\xi n}{1-i\xi \omega} - iu\xi\right]. \]  

(E-3)

The \( \nu \)-th powers are principal value, being positive real where \( C_+ \) crosses the positive imaginary axis.

Now let \( z = 1+i\xi \) in (E-3), yielding

\[ P_h(u) = \frac{-1}{12\pi} \int_{C_1} dz (z-1)^{-1} z^{-\nu} (1+z-\omega z)^{-\nu} \exp\left[\frac{(z-1)\eta}{1+z-\omega z} - u(z-1)\right]. \]  

(E-4)

The contours \( C_+ \) and \( C_1 \) in (E-3) and (E-4) are depicted as dashed lines in figure E-1. The pole at \( \xi = 0 \) is moved to \( z=1 \); the remaining singularities are branch points (\( \nu \) non-integer); the \( \nu \)-th powers are positive real where \( C_1 \) crosses the positive real axis. For \( u<0 \), an equivalent contour to \( C_1 \) is that indicated by \( C_2 \) in figure E-1, since the exponential in (E-4) furnishes rapid decay in the left-half \( z \)-plane. We write (E-4) in the form

\[ P_h(u) = \frac{\exp(u)}{12\pi} \int_{C_2} dz z^{-\nu} \exp(-uz) g_1(z), \]  

(E-5)

where

E-1
In order to get the asymptotic development of (E-5), we expand $g_1(z)$ in a power series in $z$,

$$g_1(z) = g_1(0) + g_1'(0)z + ... ,$$

(E-7)

where

$$g_1(0) = (1+\omega)^{-\nu} \exp\left(\frac{-\nu}{1+\omega}\right), \quad \frac{g_1'(0)}{g_1(0)} = 1 + \frac{\nu \omega}{1+\omega} + \frac{n}{(1+\omega)^2}.$$ 

(E-8)

Appeal to [10, p. 96, (4)] then yields (for all $\nu$)

$$p_h(u) \sim \exp\left(u \left[\frac{(-u)^{\nu-1}}{\Gamma(\nu)} \frac{g_1(0)}{g_1'(0)} + \frac{(-u)^{\nu-2}}{\Gamma(\nu-1)} \frac{g_1''(0)}{g_1'(0)}\right]\right) =$$

$$\Gamma(\nu)^{-1} (1+\omega)^{-\nu} (-u)^{\nu-1} \exp\left[u - \frac{n}{1+\omega}\right]\left[1 - \frac{\nu-1}{u} \left(1 + \frac{\nu \omega}{1+\omega} + \frac{n}{(1+\omega)^2}\right)\right]$$

as $u \to -\infty$. 

(E-9)

Substitution of (E-2) in (E-9) then yields result (104).
When (E-1) is substituted into (64) instead, we obtain the exceedance distribution function in the form

$$1 - P_h(u) = \frac{1}{i2\pi} \int_{C_-} d\xi \xi^{-1} (1+i\xi)^{-\nu} (1-i\xi\omega)^{-\nu} \exp\left[\frac{i\xi_j}{1-i\xi\omega} - iu\xi\right]. \quad (E-10)$$

Now let $z = \frac{1}{\omega} - i\xi$, to get

$$1 - P_h(u) = \frac{-1}{i2\pi} \int_{C_3} dz \left(z - \frac{1}{\omega}\right)^{-1} (1 + \frac{1}{\omega} - z)^{-\nu} (\omega z)^{-\nu} \exp\left[\frac{(1-\omega z)\xi}{\omega z} + u(z - \frac{1}{\omega})\right]. \quad (E-11)$$

The contours $C_-$ and $C_3$ in (E-10) and (E-11) are depicted as dashed lines in figure E-2. The pole at $\xi=0$ is moved to $z = 1/\omega$; the remaining singularities are branch points.

![Figure E-2. Contours of Integration for Exceedance Distribution Function](image)

For $u>0$, an equivalent contour to $C_3$ is that indicated by $C_4$ in figure E-2, since the exponential in (E-11) furnishes rapid decay in the left-half $z$-plane. We write (E-11) in the form
1 - P_h(u) = \omega (1 + \omega)^{-\nu} \exp\left(-\frac{u + \eta}{\omega}\right) \frac{1}{12\pi} \int dz \ z^{-\nu} \exp\left[uz + \frac{n}{\omega^2 z}\right] g_2(z), \quad \text{(E-12)}

where

\[ g_2(z) = (1 - \omega z)^{-1} \left(1 - \frac{\omega z}{1 + \omega}\right)^{-\nu}. \quad \text{(E-13)} \]

As above, we expand \( g_2(z) \) in a power series in \( z \),

\[ g_2(z) = 1 + \omega(1 + \frac{\nu}{1 + \omega}) z + ... \quad \text{(E-14)} \]

and substitute it in (E-12); employment of [10, p. 105, (2)] now yields (for all \( \nu \))

\[ 1 - P_h(u) \approx \frac{u}{1 + \omega} \exp(-\frac{u + \eta}{\omega}) \left[ 1 + \omega(1 + \frac{\nu}{1 + \omega}) z + ... \right] \quad \text{as } u \to +\infty, \quad \text{(E-15)} \]

where \( I_z(z) \) is the modified Bessel function of the first kind. This is the general result for the exceedance distribution function; the various parameters given in (E-2) relate it back to the problem of interest in the main text.

As a check on this result, we let \( r \to 0 \); then \( n \to 0 \), and (E-15) reduces, via [6, 9.6.7], to

\[ 1 - P_h(u) \approx \frac{u}{\omega} \exp(-\frac{u + \eta}{\omega}) \left[ 1 + \omega(1 + \frac{\nu}{1 + \omega}) z + ... \right] \quad \text{as } u \to +\infty; \ r=0. \quad \text{(E-16)} \]

Employing the identifications in (E-2), there follows from (E-16)
\begin{equation}
1 - P_h(u) \sim \frac{1+2R}{2(1+R)r(N/2)} \left(\frac{u}{2(1+R)}\right)^{\frac{N-2}{2}} \exp\left(-\frac{u}{1+2R}\right) * \\
* \left[1 + \frac{1+2R}{1+R} \frac{(N-2)(N+4+4R)}{8u}\right] \text{ as } u \to +\infty; \ r=0 . \tag{E-17}
\end{equation}

In order to compare this result with that for \( \gamma=1 \), sample mean removal, we must replace \( N \) here by \( N-1 \); see the paragraph under (101). When this is done, (E-17) reverts precisely to (61) and (62).

Returning to the general result for the exceedance distribution function in (E-15), if we keep \( n>0 \) and use [6, 9.7.1] for large arguments of \( I_u(z) \), there follows the simpler (less accurate) result

\begin{equation}
1 - P_h(u) \sim \left[2\pi^{1/2}(1+\omega)^{\frac{\nu}{2}} - \frac{1}{4}\right]^{-1} \frac{u^{1/2}}{\omega^{1/2}} \\
* u^2 - \frac{3}{4} \exp\left[-\frac{1}{\omega}(u^{1/2} - n^{1/2})^2\right] \text{ as } u \to +\infty. \tag{E-18}
\end{equation}

When (E-2) is substituted in (E-18), the result quoted in (105) follows.

As a special case of (E-18), if \( \nu=1 \) (i.e. \( N=2 \)), then

\begin{equation}
1 - P_h(u) \sim \frac{\omega^{3/2}}{2\pi^{1/2}(1+\omega)(n\omega)^{1/4}} \exp\left[-\frac{1}{\omega}(u^{1/2} - n^{1/2})^2\right] \text{ as } u \to +\infty; \ \nu=1. \tag{E-19}
\end{equation}
APPENDIX F. EXCEEDANCE DISTRIBUTION FUNCTION FOR $\gamma=0$, $N=1$, $r>0$

CHARACTERISTIC FUNCTION APPROACH

When characteristic function (100) with $N=1$ is substituted in (64), the expression for the exceedance distribution function becomes

$$1 - P_h(u) = \frac{1}{2\pi} \int_C \frac{d\xi}{(1+i\xi)^{1/2} (1-i\xi\omega)^{1/2}} \exp \left[ \frac{i\xi^2}{1-\xi\omega} - iu\xi \right], \quad (F-1)$$

where $\omega = 1+2R$ as in (102). The square roots are taken as $+1$ at $\xi=0$. For $u>0$, the contour $C_-$ can be modified to that indicated in figure F-1, where the contributions of the large circular arcs in the lower-half $\xi$-plane tend to zero. The small circle of radius $\rho$ centered at branch point $\xi = -i/\omega$ must have $0 < \rho < 1/\omega$, since the latter is the distance to the pole at the origin.

![Figure F-1. Equivalent Contours for (F-1)](image)

It is easy to show that the two vertical contributions in figure F-1 are equal. Under the change of variable

$$i\xi = \frac{1+t^2}{\omega}, \quad (F-2)$$
the sum of the two vertical contributions to exceedance distribution function (F-1) becomes

\[ V(\rho) = \frac{2\omega^{1/2}}{\pi} \exp\left(-\frac{u^2 + r^2}{\omega}\right) \int_{\xi \rho}^{+\infty} dt \left(1+t^2\right)^{-1} \]

\[ \ast \left(1+\omega t^2\right)^{-1/2} \exp\left(-\frac{u^2 t^2}{\omega} - \frac{r^2}{\omega t^2}\right). \]

(F-3)

This integral remains convergent even as \( \rho \to 0^+ \). Furthermore, the integrand decays rapidly, has no cusps, and involves only elementary functions; also the integral is a sum of positive quantities and retains significance even for large \( \rho \).

On the small circular contour \( C_\rho \) in figure F-1, let

\[ i\Phi = \frac{1}{\omega} - \rho \exp(i\phi), \]

(F-4)

to obtain, for the circular contribution to the exceedance distribution function, the quantity

\[ C(\rho) = \frac{\omega}{2\pi(1+\omega)^{1/2}} \exp\left(-\frac{u^2 + r^2}{\omega}\right) \rho^{1/2} \int_{-\pi}^{\pi} d\phi \left(1-\omega E\right)^{-1} \]

\[ \ast \left(1 - \frac{\omega}{1+\omega} E\right)^{-1/2} \exp\left(i\frac{\phi}{2} + \frac{\lambda E}{\rho} \right) \exp(\rho E), \]

(F-5)

where we define in this appendix

\[ \lambda = \frac{r^2}{\omega} = \left(\frac{r}{1+2R}\right)^2, \ E = \exp(i\phi). \]

(F-6)
The exceedance distribution function is given exactly by the sum of (F-3) and (F-5), for any \( 0 < \rho < 1/\omega \). It would be advantageous numerically to let \( \rho \to 0^+ \) in these two equations; however, the limit of (F-5) is not obvious and can easily be done incorrectly.

AN ERRONEOUS APPROACH FOR \( C(\rho) \)

It is tempting to let \( \rho \to 0^+ \) in those locations in (F-5) where it will "do no damage", obtaining for the integral with scale factor \( \rho^{1/2} \) the quantity

\[
I_\rho = \rho^{1/2} \int_{-\pi}^{\pi} d\phi \exp\left[ i \frac{\phi}{2} + \frac{\lambda}{\rho} \exp(-i\phi) \right].
\]

(The fallacy of doing this for a residue calculation with an essential singularity is demonstrated in the next subsection.) Furthermore, the limit of (F-7) as \( \rho \to 0^+ \) can in fact be determined in closed form, as follows. Observe that the integrand of (F-7) has a saddle point in the complex \( \phi \)-plane at

\[
\phi_s = -iL, \quad L = \ln\left(\frac{2\lambda}{\rho}\right);
\]

this is in fact the only saddle point in the \((-\pi, \pi)\) strip in the \( \phi \)-plane. Now let \( z = \phi - \phi_s \), getting for (F-7)

\[
I_\rho = (2\lambda)^{1/2} \int_{-\pi+iL}^{\pi+iL} dz \exp\left[ i \frac{z}{2} + \frac{1}{2} \exp(-iz) \right].
\]

The radius \( \rho \) now appears only in the limit \( L \) of the integral, and the integrand has a saddle point at \( z=0 \).
The straight line contour for (F-9) can be deformed into contour C, depicted in figure F-2, which goes through the saddle point at z=0. Now if

\[ \rho \to 0^+, \quad L \to +\infty, \]  
and (F-9) yields

\[ I_0 = (2\lambda)^{1/2} \int_{-\pi+i\infty}^{\pi+i\infty} dz \exp \left[ i \frac{z}{2} + \frac{1}{2} \exp(-iz) \right], \quad \text{ (F-10)} \]

where the contour is the limit of C in figure F-2 as L \to +\infty; that is, the integral is between the two valleys at \pm \pi+i\infty and is connected by the saddle point at z=0.

The steepest descent curves out of the saddle point of the integrand of (F-10) are given explicitly by

\[ y = -\ln\left(\frac{\sin x}{x}\right) \quad \text{for} \quad -\pi < x < \pi. \quad \text{ (F-11)} \]

Thus if we let
\[ z = x + iy = x - i \ln\left(\frac{\sin x}{x}\right), \]

\[ \frac{dz}{dx} = 1 - i \left(\frac{\cos x}{\sin x} - \frac{1}{x}\right), \] (F-12)

on the steepest descent curves in (F-10), there follows

\[
I_0 = (2\lambda)^{1/2} \int_{-\pi}^\pi dx \left(1 - i \frac{\cos x}{\sin x} + i \frac{1}{x}\right) \exp \left[i \frac{x}{2} + \frac{1}{2} \ln\left(\frac{\sin x}{x}\right) + \frac{1}{2} \frac{x}{\sin x} \exp(-ix)\right] =
\]

\[
= (2\lambda)^{1/2} \int_{-\pi}^\pi dx \left(1 - i \frac{\cos x}{\sin x} + i \frac{1}{x}\right) \left(\frac{\sin x}{x}\right)^{1/2} \exp\left[\frac{x \cos x}{2 \sin x}\right] =
\]

\[
= (2\lambda)^{1/2} 2 \int_0^\pi dx \left(\frac{\sin x}{x}\right)^{1/2} \exp\left(\frac{x \cos x}{2 \sin x}\right) = (2\lambda)^{1/2} 2(2\pi)^{1/2} = 4(\lambda \pi)^{1/2} \quad \text{(F-13)}
\]

(The integral value of \((2\pi)^{1/2}\) in (F-13) was deduced by numerical integration.)

Recalling the definition of \(I_\rho\) in (F-7), we then have the dubious result for the limit of (F-5):

\[
C(0) \approx \left(\frac{2}{\pi}\right)^{1/2} \frac{r}{(1+R)^{1/2}} \exp\left(-\frac{u+r^2}{1+2R}\right), \quad \text{(F-14)}
\]

where we employed (102) and (F-6). Actual numerical evaluation of (F-14), combined with \(V(0)\) from (F-3), gives incorrect results for the exceedance distribution function (F-1); thus the replacement of \(\rho\) with 0 in (F-5) is invalid. The explanation for this pitfall is the essential singularity of (F-1) at \(\zeta = -i/\omega\); a simpler illustration follows.
RESIDUE OF ESSENTIAL SINGULARITY

The function

\[ \exp\left(\frac{1}{z}\right) = 1 + \frac{1}{z} + \frac{1}{2!} z^2 + \ldots \]  (F-15)

has an essential singularity at \( z=0 \), with residue 1, as exemplified by this Laurent expansion. Now consider the function

\[ f(z) = \exp\left(\frac{1}{z}\right) g(z) , \]  (F-16)

where \( g(z) \) is analytic at \( z=0 \). Then

\[ f(z) = \left(1 + \frac{1}{z} + \frac{1}{2!} z^2 + \ldots \right) \left(g(0) + g^{(1)}(0) z + \frac{1}{2!} g^{(2)}(0) z^2 + \ldots \right) . \]  (F-17)

The coefficient of \( 1/z \) in (F-17) is the residue of \( f(z) \) at \( z=0 \); namely

\[ \text{Res} = g(0) + \frac{1}{2!} g^{(1)}(0) + \frac{1}{3!} \frac{1}{2!} g^{(2)}(0) + \ldots = \sum_{n=0}^{\infty} \frac{g^{(n)}(0)}{n!(n+1)!} . \]  (F-18)

Thus the residue of \( f(z) \) at \( z=0 \) depends on the behavior of \( g(z) \) in a neighborhood of \( z=0 \), and not just the value \( g(0) \).

A couple of examples yield the following:

\[ g(z) = -(1-az)^{-1} , \quad \text{Res} = \frac{\exp(a)-1}{a} ; \]

\[ g(z) = \exp(a^2 z) , \quad \text{Res} = \frac{I_1(2a)}{a} . \]  (F-19)
CORRECT APPROACH FOR C(p)

Reconsider the integral in (F-5) plus the scale factor \( p^{1/2} \); making the substitution \( z = e^{+iL} \), where \( L \) is given in (F-8), there follows for this quantity

\[
(2\lambda)^{1/2} \int_{-\pi+iL}^{\pi+iL} dz \left( 1 - 2\omega e^{iz} \right)^{-1} \left( 1 - \frac{2\omega}{1+i\omega} e^{iz} \right)^{-1/2} \star \left[ i \frac{z}{2} + \frac{1}{2} e^{-iz} + 2\lambda e^{iz} \right].
\]

The uppermost singularity of the integrand in the \( z \)-plane (within the \(-\pi, \pi\) strip) is a pole at \( z_p = i \ln(2\omega) \); however, the straight line contour in (F-20) remains above this pole because \( p < 1/\omega \); see (F-8). Furthermore, the total integrand of (F-20) has a saddle point on the imaginary axis of the \( z \)-plane above the pole location \( z_p \), because the integrand is infinite at the pole and at \( z = 0 + i\omega \). Thus the straight line contour in (F-20) can be modified so as to pass through the saddle point, and yet remain above \( z_p \). Finally, letting \( p \to 0^+ \), then \( L \to +\infty \), and (F-20) combined with (F-5) yields the exact result for the circular component

\[
C(0) = \frac{r}{\pi 2^{1/2}} \frac{1}{(1+i\omega)^{1/2}} \exp \left( \frac{-u+\gamma^2}{\omega} \right) \int_{-\pi+i\omega}^{\pi+i\omega} dz \left( 1 - 2\omega e^{iz} \right)^{-1} \star \left( 1 - \frac{2\omega}{1+i\omega} e^{iz} \right)^{-1/2} \exp \left[ i \frac{z}{2} + \frac{1}{2} e^{-iz} + 2\lambda e^{iz} \right],
\]

where the two valleys of the integrand at \( \pm \pi + i\omega \) are joined with a contour through the saddle point lying above the pole at \( z_p = i \ln(2\omega) \). Here \( \omega = 1+2\gamma \).
The other component of the exceedance distribution function, corresponding to (F-21), is given by (F-3) at \( \rho = 0^+ \):

\[
V(0) = \frac{2}{\pi} \frac{1}{2} \exp \left( - \frac{u^2}{\omega} \right) \int_0^{\infty} dt (1+t^2)^{-1} \exp \left( - \frac{r^2}{\omega t^2} \right).
\]  (F-22)

Thus for \( u \geq 0 \), (F-1) and figure F-1 yield exceedance distribution function

\[
1 - P_h(u) = C(0) + V(0) = (F-21) + (F-22).
\]  (F-23)

Computationally, (F-21) is not too attractive, because of the complex integrand and/or the need to determine the steepest descent paths to \( \pm \pi + i \infty \) numerically. Accordingly, an alternative direct procedure for determining the exceedance distribution function of random variable \( h \) is now presented.

**DIRECT EVALUATION OF EXCEEDANCE DISTRIBUTION FUNCTION**

For \( \gamma = 0, N = 1 \), (12) and (3) yield the crosscorrelator output for the signal and noise model as

\[
q = u_1 v_1 = [u_u + u_s(1) + u_d(1)] [u_v + v_s(1) + v_d(1)].
\]  (F-24)

The normalized crosscorrelator output is then, from (49) and (97),

\[
h = \frac{q}{(D_u D_v)^{1/2}} = \left( r_u + u_s^t + u_d^t \right) \left( r_v + v_s^t + v_d^t \right) = xy,
\]  (F-25)

where \( x \) and \( y \) are joint Gaussian random variables with statistics
\[
\bar{x} = r_u, \quad \bar{y} = r_v,
\]
\[
\sigma_x^2 = 1 + R_u, \quad \sigma_y^2 = 1 + R_v, \quad \frac{(x-\bar{x})(y-\bar{y})}{\rho_s(R_uR_v)^{1/2}}. \quad (F-26)
\]

We now make the same assumptions as in (99); see also (56) et seq. Then (F-26) specializes to
\[
\bar{x} = \bar{y} = r, \quad \sigma_x^2 = \sigma_y^2 = 1 + R, \quad \rho_{xy} = \frac{R}{1+R}. \quad (F-27)
\]

The joint probability density function of \(x, y\) is then given by
\[
p_2(x,y) = \left[2\pi\sigma^2(1-\rho_{xy}^2)^{1/2}\right]^{-1}\exp\left[-\frac{(x-r)^2 + (y-r)^2 - 2\rho_{xy}(x-r)(y-r)}{2\sigma^2(1-\rho_{xy}^2)}\right]. \quad (F-28)
\]

We now have cumulative distribution function
\[
P_h(u) = \iint_{R_2+R_4} dx \, dy \, p_2(x,y) \quad \text{for} \ u \leq 0, \quad (F-29)
\]

and exceedance distribution function
\[
1 - P_h(u) = \iint_{R_1+R_3} dx \, dy \, p_2(x,y) \quad \text{for} \ u \geq 0, \quad (F-30)
\]

where regions \(R_1, R_2, R_3, R_4\) are indicated in figure F-3.
If we now rotate axes according to
\[ s = \frac{x+y}{2}, \quad t = \frac{x-y}{2} \]  \hspace{1cm} (F-31)
and employ (84) and (F-27), there follows (after scale changes of the
variables)
\[ P_h(u) = 2 \int_{-\infty}^{\infty} dv \phi\left(v - r\left(\frac{2}{\omega}\right)^{1/2}\right) \Phi\left(-\left(\omega v^2 - 2u\right)^{1/2}\right) \quad \text{for} \quad u \leq 0 , \]  \hspace{1cm} (F-32)
and
\[ 1 - P_h(u) = 2 \int_{0}^{\infty} dv \phi(v) \left[ \Phi\left(\frac{r - \sqrt{u+v^2/2}}{\sqrt{\omega/2}}\right) + \Phi\left(\frac{r - \sqrt{u+v^2/2}}{\sqrt{\omega/2}}\right) \right] \quad \text{for} \quad u \geq 0 . \]  \hspace{1cm} (F-33)
Here \( \omega = 1+2R \). These real integrals are very useful for the evaluation of the
distributions of \( h \) when \( \gamma=0, N=1 \). In fact, (F-33) is preferred over
(F-21)-(F-23); but (106) is preferred over (F-32) since \( \Phi \) need not be
evaluated in (106). This is in fact the procedure utilized here to obtain
numerical results for this case of \( \gamma=0, N=1 \).
APPENDIX G. PROGRAM FOR EVALUATION OF OPERATING CHARACTERISTICS FOR $\gamma = 0$

The comments in appendices C and D are relevant here also. The characteristic function used as the starting point is given by (100). It was observed under (103) that $|f_h(\xi)|$ for (100) is monotonically decreasing for all $\xi > 0$; thus the choice of truncation value $L$ is simplified; see appendix D comments. A table for sampling increment $\Delta_0$ (when $R=0$) follows.

<table>
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<tr>
<th>$N$</th>
<th>$\Delta_0$ for $r=1$</th>
<th>$\Delta_0$ for $r=2$</th>
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<td>.05</td>
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<tr>
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<td>8</td>
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<td>.02</td>
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<td>.03</td>
<td>.02</td>
</tr>
<tr>
<td>64</td>
<td>.025</td>
<td>.015</td>
</tr>
<tr>
<td>128</td>
<td>.020</td>
<td>.010</td>
</tr>
<tr>
<td>256</td>
<td>.012</td>
<td>.007</td>
</tr>
</tbody>
</table>

Table G-1. Values of $\Delta_0$ for $\gamma = 0$

G-1
TR 7045
230 S=Sc(1)
240 B=Sc(14)
250 Scale=(B-S)/(0-S)
260 X1=30
270 X2=170
280 Y1=35
290 Y2=Y1+(X2-X1)*Scale
300 PLOTTER IS "9872A"
310 LIMIT X1,X2,Y1,Y2
320 OUTPUT 705;"VS3"
330 SCALE S,0,.S,B
340 FOR I=1 TO 14
350 MOVE S,Sc(I)
360 DRAW 0,Sc(I)
370 NEXT I
380 FOR I=1 TO 11
390 MOVE Sc(I),S
400 DRAW Sc(I),B
410 NEXT I
420 MOVE S,S
430 DRAW 0,0
440 PENUP
450 M1=Mf-1
460 N2=Nc/2
470 Rsn=Nc*Rs+Rs
480 FOR In=0 TO 12
490 IF In>0 THEN 520
500 Rc=0
510 GOTO 530
520 Rc=2*Ns(In)
530 OUTPUT 6:"R=";Rc," Delta =";Delta(In)
540 ASSIGN #1 TO "ABSCIS" ! Temporary storage
550 Delta=Delta(In) ! for false alarm probability
560 R2=Rc+2
570 R21=R2+1
580 Mux=Nc*Rc+Rsn ! Mean of random variable h
590 Muy=Mux+Bs ! Mean of shifted variable y
600 REDIM X(0:M1),Y(0:M1)
610 MAT X=2ER
620 MAT Y=2ER
630 X(0)=0
640 Y(0)=.5#Delta+Muy
650 Ls=0
660 Ls=Ls+1
670 Xi=Delta+Ls ! Argument xi of char. fn.
680 Ei=Xi*R21 ! Calculation
690 CALL Log(1+Xi+Ei,-Xi*R2,Ai,Bi)
700 CALL Div(0,Xi+Rsn,1,-Ei,Ci,Di)
710 CALL Exp(Ci-N2+Ai,Di+Xi*Bs-N2+Bi,Fyr,Fyi)
720 Ms=Ls MOD Mf
730 Ar=Fyr/Ls
740 Hi=Fyi/Ls
750 X(Ms)=X(Ms)+Ar
760 Y(Ms)=Y(Ms)+Hi
770 Magsq=Ar*Ar+Ai*Ai
780 IF Magsq>1E-24 THEN 660
790 OUTPUT 0:"Xi =";Xi;" Mag =";SQR(Magsq)
800 CALL Fft13z(Ms,X(*),Y(*))
G-2
FOR Ms=0 TO M1
820 T=Y(Ms)/PI-Ms/Mf
830 X(Ms)=.5-T       ! Cumulative distribution function
840 Y(Ms)=.5+T       ! Exceedance distribution function
850 NEXT Ms
860 OUTPUT 0;Y(0);Y(1);Y(M1-1);Y(M1)
870 PLOTTER IS "GRAPHICS"
880 GRAPHICS
890 SCALE 0,Mf,-14,0
900 LINE TYPE 3
910 GRID Mf/8,1
920 PENUP
930 LINE TYPE 1
940 FOR Ms=0 TO M1
950 Pr=Y(Ms)
960 IF Pr>=1E-12 THEN Y=LGT(Pr)
970 IF Pr<=-1E-12 THEN Y=-24-LGT(-Pr)
980 IF ABS(Pr)<1E-12 THEN Y=-12
990 PLOT Ms,Y
1000 NEXT Ms
1010 PENUP
1020 FOR Ms=0 TO M1
1030 Pr=X(Ms)
1040 IF Pr>=1E-12 THEN Y=LGT(Pr)
1050 IF Pr<=-1E-12 THEN Y=-24-LGT(-Pr)
1060 IF ABS(Pr)<1E-12 THEN Y=-12
1070 PLOT Ms,Y
1080 NEXT Ms
1090 PENUP
1100 DUMP GRAPHICS
1110 OUTPUT 0;"
1120 IF In>0 THEN 1270
1130 FOR Ms=0 TO M1
1140 IF Y(Ms)<.7 THEN 1160
1150 NEXT Ms
1160 M2=Ms-1
1170 FOR Ms=M2 TO M1
1180 IF Y(Ms)<0 THEN 1200
1190 NEXT Ms
1200 M3=Ms-1
1210 REDIM X(M2:M3)
1220 FOR Ms=M2 TO M3
1230 X(Ms)=FNInuphi(Y(Ms))
1240 NEXT Ms
1250 PRINT #1;X(*)       ! Store false alarm probability
1260 GOTO 1460
1270 REDIM X(M2:M3)
1280 READ #1;X(*)       ! Read in false alarm probability
1290 Id=Idelta(In)
1300 J2=INT(M2/Id)
1310 J3=INT(M3/Id)+1
1320 FOR J=J2 TO J3
1330 Y(J)=FNInuphi(Y(J))
1340 NEXT J
TR 7045
1350 PLOTTER IS "0872A"
1360 LIMIT X1,X2,Y1,Y2
1370 OUTPUT 705; "VS3"
1380 SCALE 8,0,8,8
1390 FOR J=J2 TO J3
1400 T=J*I
1410 IF T<M2 THEN 1440
1420 IF T>M3 THEN 1450
1430 PLOT X(T),Y(J)
1440 NEXT J
1450 PENUP
1460 NEXT In
1470 END
1480 !
1490 SUB Div(X1,Y1,X2,Y2,R,B) ! DIV(Z)
1540 !
1550 SUB Exp(X,Y,R,B) ! EXP(Z)
1600 !
1610 SUB Log(X,Y,R,B) ! PRINCIPAL LOG(Z)
1690 !
1700 DEF FNInvphi(X) ! INVPHI(X) via AMS 55, 26.2.23
1830 !
1840 SUB Fftl3z(N,X(*),Y(*)) ! N <= 2^13, N=2^INTEGER, 0 SUBSCRIPT
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