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THESIS

Estimation of Sonobuoy Position
Relative to an Aircraft
Using Extended Kalman Filters

by
Nicholas Mason Brownsberger

September 1979

Thesis Advisor:

D. J. Collins

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REPORT DOCUMENTATION PAGE		READ INSTRUCTIONS BEFORE COMPLETING FORM
1. REPORT NUMBER	2. GOVT ACCESSION NO.	3. RECIPIENT'S CATALOG NUMBER
4. TITLE (and Subtitle) 6 Estimation of Sonobuoy Position Relative to an Aircraft Using Extended Kalman Filters .		5. TYPE OF REPORT & PERIOD COVERED 9 Master's and Engineer's Thesis, September 1979
7. AUTHOR(s) 10 Nicholas Mason/Brownsberger		6. PERFORMING ORG. REPORT NUMBER
9. PERFORMING ORGANIZATION NAME AND ADDRESS Naval Postgraduate School Monterey, California 93940		8. CONTRACT OR GRANT NUMBER(s) 12 151
11. CONTROLLING OFFICE NAME AND ADDRESS Naval Postgraduate School Monterey, California 93940		10. PROGRAM ELEMENT, PROJECT, TASK AREA & WORK UNIT NUMBERS
14. MONITORING AGENCY NAME & ADDRESS (if different from Controlling Office) Naval Postgraduate School Monterey, California 93940		12. REPORT DATE 11 September 1979
		13. NUMBER OF PAGES
		15. SECURITY CLASS. (of this report) Unclassified
		15a. DECLASSIFICATION/DOWNGRADING SCHEDULE
16. DISTRIBUTION STATEMENT (of this Report) Approved for public release; distribution unlimited		
17. DISTRIBUTION STATEMENT (of the abstract entered in Block 20, if different from Report)		
18. SUPPLEMENTARY NOTES		
19. KEY WORDS (Continue on reverse side if necessary and identify by block number) SRS (Sonobuoy Reference System) Kalman Filter Sonobuoy		
20. ABSTRACT (Continue on reverse side if necessary and identify by block number) In airborne anti-submarine warfare there is a need to more accurately determine the positions of sonobuoys on the surface of the water. This report develops two algorithms which employ extended Kalman filters to determine estimated position. The bearing from the aircraft to the sonobuoy is the primary measurement. Range information is not available. The first algorithm is a six-state filter which was reduced from the \rightarrow next page		

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Estimation of Sonobuoy Position
Relative to an Aircraft
Using Extended Kalman Filters

by

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Lieutenant, United States Navy
B.S., United States Naval Academy, 1972

Submitted in partial fulfillment of the
requirements for the degrees of

MASTER OF SCIENCE IN AERONAUTICAL ENGINEERING
AND
AERONAUTICAL ENGINEER

from the

NAVAL POSTGRADUATE SCHOOL
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ABSTRACT

In airborne anti-submarine warfare there is a need to more accurately determine the positions of sonobuoys on the surface of the water. This report develops two algorithms which employ extended Kalman filters to determine estimated position. The bearing from the aircraft to the sonobuoy is the primary measurement. Range information is not available. The first algorithm is a six-state filter which was reduced from the 13-state system developed by the Orincon Corporation. Its states include relative position, relative velocity, and inertial misalignments. The second algorithm includes two cascaded Kalman filters. The primary two-state filter estimates sonobuoy position. A secondary filter estimates drift from information obtained from the primary filter. Both algorithms successfully estimated sonobuoy position for simulated aircraft data. The effect of aircraft-to-sonobuoy range, the frequency of measurement, and changes in altitude are also analyzed.

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ACKNOWLEDGEMENT

I would like to thank Dr. Birnbaum and Ms. Peggy Pembroke of NADC for their responsiveness in answering questions and supplying computer information. And to Dr. Collins I would like to express my gratitude for his help and guidance without which this thesis could not have been completed. To my wife I would like to offer my special thanks and love for her patience, understanding, and fortitude during a time which tested both of us.

Nick Brownsberger

I. INTRODUCTION

The P3 Orion is the U.S. Navy's primary long range anti-submarine warfare (ASW) aircraft. It is outfitted with equipment which allows it to search for, locate, and track submarines. The aircraft carries a Univac digital computer, the CP 901, which performs much of the navigational and tactical plotting chores. The primary sensor used by this and many other ASW aircraft is an airdropped listening device known as a sonobuoy. The Orion generally deploys several sonobuoys (4 to 20) in patterns which can cover a thousand square miles of ocean while searching for the submarine. Once contact has been made, these sonobuoys provide information which locates the submarine. The target can then be tracked (or attacked if required) until the Orion's mission is complete.

The sonobuoy is dropped from the aircraft at the geographical location designated by the aircrew. Once in the water the sonobuoy floats and deploys a hydrophone to depths varying from 50 to 500 feet. The information picked up is transmitted back to the aircraft where it is analyzed. This information can consist of the relative intensity of target noise, bearings and sometimes ranges to the target, all of which are used to fix the current position of the submarine. In order to maintain close, accurate tracking and be able to launch an attack, the submarine's position must

be accurate to within several hundred yards. The information received from the sonobuoys has some error inherent in the nature of the measurements made by the hydrophone. Sonobuoy position error also contributes significantly to the submarine tracking inaccuracies.

Historically, the positions of the sonobuoys were determined and updated by "mark-on-top"s. This required the aircraft to home on the transmitting sonobuoy until the buoy was overflown. At that instant the aircraft's position was entered into the on-board computer which slewed the buoy to this updated position. After many of these updates the computer was able to develop a bias which was applied to the sonobuoy positions in the computer effectively allowing them to drift. The method had several disadvantages. Error in the updated position was at least as great as the aircraft altitude at the time of the "mark" which could vary from 300 to 22,000 feet. The accuracy also depended on the consistency of the several pilots who might be making the mark-on-tops during the flight. The updating was not continuous in that it was several minutes between consecutive marks on one buoy at best, and more likely 30 to 60 minutes. Not all buoys were even updated. Furthermore, this method required the aircraft to overfly the submarine many times in order to make the mark-on-tops. This should be avoided.

The purpose of this thesis was to investigate some alternative methods for accurately fixing the position of

sonobuoys. They should allow the aircraft to stand-off from the sonobuoy field and still produce more accurate fixing than the historical method provided. The Naval Air Development Center (NADC) at Warminster, PA. had already partially developed such a system. This thesis was undertaken in support of their work but was conducted independently. Their system, the Sonobuoy Reference System (SRS), was already installed on the aircraft and had the capability of measuring the relative bearing to any transmitting sonobuoy. Additional information available for use included aircraft heading, altitude, and airspeed as well as doppler velocity and drift angle. Also, the aircraft's Inertial Navigational System (INS) provided geographical position although the Schuler cycle and inertial drifts could make this position several nautical miles in error. On the other hand, aircraft-to-sonobuoy range and sea surface drift information were assumed not to be available. An attempt was made to determine sonobuoy position at least relative to the aircraft with a less accurate geographical position as a secondary objective. Kalman filtering techniques were used based primarily on measurements of bearing from the aircraft to the sonobuoy.

II. FILTERS

Kalman filtering is a recursive technique for estimating the state of a system. It was developed in the 1960's by R. E. Kalman and improved upon previous methods by Wiener and others. The Wiener filter is based on frequency domain designs which are statistically optimal but are only applicable to stationary processes. The Kalman filter is based on state-space, time domain formulations and is especially suited to digital computers. From a simplistic, one-dimensional point of view, the Kalman filter recursively averages noisy measurements to provide a more precise estimate of the actual value.

Assume that a system can be linearly modeled with state equations in matrix form as

$$X_k = \Phi_{k-1} X_{k-1} + A_{k-1} U_{k-1} + \Gamma_{k-1} W_{k-1} \quad (1)$$

where X_k represents the states of the system at the k th interval. (Only the discrete case was be considered in this study.) Φ_k is the transition matrix and is used to "propagate" the system from K to $K+1$. U_k represents the control input to the system and W_k represents white, gaussian noise with $\bar{0}$ mean and Q_k variance, written $N(\bar{0}, Q_k)$. Measurements are required to update the system and are

described by the linear matrix equation

$$Z_k = H_k X_k + V_k \quad (2)$$

where Z_k is the measurement. H_k describes the relationship between the states and the measurement and V_k is measurement noise described by $N(0, R_k)$.

The discrete Kalman filter equations are:

Propagate

$$\hat{X}_k(-) = \Phi_{k-1} \hat{X}_{k-1}(+) \quad (3)$$

$$P_k(-) = \Phi_{k-1} P_{k-1}(+) \Phi_{k-1}^T + Q_{k-1} \quad (4)$$

Update

$$G_k = P_k(-) H_k^T \left[H_k P_k(-) H_k^T + R_k \right]^{-1} \quad (5)$$

$$\hat{X}_k(+) = \hat{X}_k(-) + G_k \left[Z_k - H_k \hat{X}_k(-) \right] \quad (6)$$

$$P_k(+) = \left[I - G_k H_k \right] P_k(-) \quad (7)$$

where

\hat{X}_k = estimate of the state X_k

P_k = covariance

G_k = Kalman gain

The estimate prior to the measurement is denoted by $\hat{X}_k(-)$ and can be updated after the measurement to a new estimate denoted $\hat{X}_k(+)$. This notation is shown in figure 1. The covariance matrix provides a statistical measure of the uncertainty of \hat{X} . Consider a 2 x 2 covariance matrix where the error \tilde{X} in the estimate is defined as $\hat{X} - x$:

$$P = \begin{bmatrix} E(\tilde{X}_1^2) & E(\tilde{X}_1, \tilde{X}_2) \\ E(\tilde{X}_1, \tilde{X}_2) & E(\tilde{X}_2^2) \end{bmatrix} = \begin{bmatrix} \sigma_1^2 & \sigma_1 \sigma_2 \\ \sigma_1 \sigma_2 & \sigma_2^2 \end{bmatrix} \quad (6)$$

The diagonal elements represent the mean square errors of the corresponding state variables x_1 and x_2 . The off-diagonal elements are indicators of cross-correlation between the states. The Kalman gain G is an optimal gain chosen so as to minimize the sum of the diagonal terms of the covariance matrix.

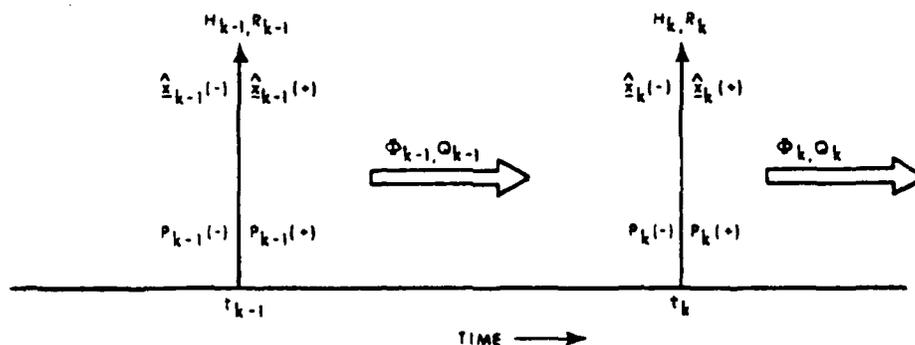


Figure 1. Kalman filter notation

If either the system model or the measurement model is non-linear, then non-linear filters must be used. Let the non-linear system be described by

$$X_k = \phi(X_{k-1}) + W_{k-1} \quad ; \quad W_k \sim N(0, Q_k) \quad (9)$$

$$Z_k = h(X_k) + V_k \quad ; \quad V_k \sim N(0, R_k) \quad (10)$$

where $\phi(\hat{X}_{k-1})$ represents the non-linear equations of the system. They can be linearized by a Taylor series expansion around the latest estimate of the state to obtain $\Phi(\hat{X}_{k-1})$. Likewise, $h(X_k)$ represents the non-linear measurement equations and by a Taylor series expansion yields $H(\hat{X}_k)$.

$$\phi(X_k) = \phi(\hat{X}_k) + \Phi(\hat{X}_k) [X_k - \hat{X}_k] + \dots \quad (11)$$

$$h(X_k) = h(\hat{X}_k) + H(\hat{X}_k) [X_k - \hat{X}_k] + \dots \quad (12)$$

where

$$\Phi(\hat{X}_k) = \frac{\partial}{\partial \mathbf{x}} \left[\phi(X_k) \right]_{\mathbf{x}=\hat{\mathbf{x}}} \quad (13)$$

$$H(\hat{X}_k) = \frac{\partial}{\partial \mathbf{x}} \left[h(X_k) \right]_{\mathbf{x}=\hat{\mathbf{x}}} \quad (14)$$

Second order Taylor series terms are neglected. Now, the extended Kalman filter can be implemented with the following equations:

Propagate

$$\hat{X}_k(-) = \phi(\hat{X}_{k-1}) \quad (15)$$

$$P_k(-) = \Phi(\hat{X}_{k-1}(-)) P_{k-1}(+) \Phi(\hat{X}_{k-1}(-))^T + Q_{k-1} \quad (16)$$

Update

$$G = P_k(-) H(\hat{X}_k(-))^T \left[H(\hat{X}_k(-)) P_k(-) H(\hat{X}_k(-))^T + R_k \right]^{-1} \quad (17)$$

$$\hat{X}_k(+) = \hat{X}_k(-) + G_k \left[Z_k - H(\hat{X}_k(-)) \right] \quad (18)$$

$$P_k(+) = \left[I - G_k H(\hat{X}_k(-)) \right] P_k(-) \quad (19)$$

Higher order filters can be used if the linearization errors are large. The second order Kalman filter employs one more term in the Taylor series expansion by modifying the update equations of the extended Kalman filter to account for this term. The iterated extended Kalman filter uses the same equations as does the extended Kalman filter. However, the calculations are repeated, each time linearizing about the most recent estimate, until there is little further improvement with each new iteration. The iterated extended filter can greatly reduce the errors due to non-linearities, more so than the second order filter.

Kalman filters should be based on correctly modeled systems and accurate noise statistics to ensure proper

performance. This is not always possible either due to ignorance about the system or lack of sufficient statistical information. A filter which is not operating properly may diverge. Apparent divergence describes that situation where the true estimation errors are larger than the filter predicted errors although they are bounded. True divergence is characterized by errors which continue to grow with time and eventually become infinite. These divergence phenomenon are depicted in figure 2.

There are several ways to overcome the divergence problem when the modeling is not completely accurate such as adding fictitious noise. This allows the filter a little more freedom to adjust to whatever modeling inconsistencies may exist, but makes the filter estimate appear more erratic. Another method which helps overcome divergence is finite memory filtering. Since Kalman gains tend to grow smaller and smaller as time passes, they may reach a point

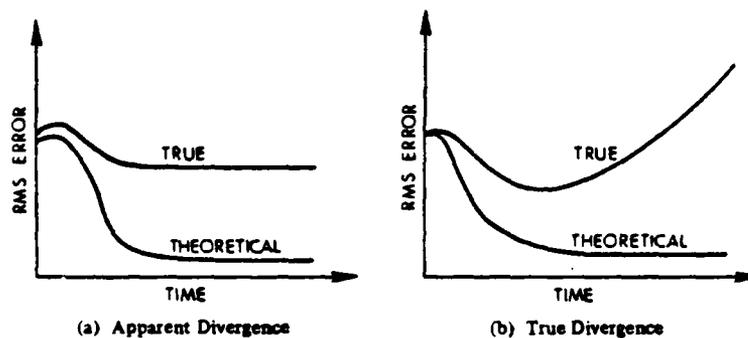


Figure 2. Divergence

where new measurement information has no effect on the estimates. Finite memory filtering effectively eliminates old data which is no longer useful by keeping the gains significant. This is sometimes called a "moving window".

Some simplifying techniques used in Kalman filters include precomputed gains. Although forfeiting the optimal Kalman gains, this has the advantage of reducing computational burden. More importantly, the gains can be controlled to overcome modeling weaknesses if necessary. Another technique which can be used when more than one measurement is provided to a filter is processing them one at a time. This avoids taking the inverse of more than a scalar when computing the updated covariance matrix $\hat{P}_k(+)$. This is possible if the simultaneous measurements are considered to be taken sequentially over a zero time span.

III. MODELS

A. GENERAL APPROACHES TO SYSTEM MODELING

The aircraft-sonobuoy system must be modeled in state-space for use with the Kalman filter. There are at least two approaches to the modeling of this problem depending on the point of view. One intuitive approach is to assume the sonobuoy drifts at a constant velocity and then use aircraft-to-sonobuoy bearing measurements to locate the sonobuoy. The states become sonobuoy position and sonobuoy velocity. Unfortunately, this problem is not observable. The bearing measurements provide only information about position; there is no rate of change information in the bearings themselves. In addition, the aircraft must maintain a track of its geographical position between updates in order to determine the next expected measurement for the Kalman filter. As mentioned before, this aircraft position is subject to non-linear as well as linear navigational drifts which are not taken into account in this model. However, an observable system can be obtained by reducing the number of states to sonobuoy position only and introducing fictitious process noise to account for the drift. This noise effectively allows sonobuoy position to update so as to keep up with the drift. One approach developed by this thesis is a variation of this concept.

Another approach considers only the relative position of the sonobuoy with respect to the aircraft. Relative velocities and relative accelerations must be taken into account and these change radically as the aircraft flies in the tactical situation. Sonobuoy position is not obtained directly. (When only the word "position" is used it will indicate the position relative to an earth fixed coordinate system, such as latitude and longitude. "Relative position" will always mean the location with respect to an aircraft fixed coordinate system.)

There are some other considerations which should be addressed. The sonobuoy drift is generally slow (less than 5 NM/hr most of the time) and constant. It is not unreasonable to assume that the entire sonobuoy field drifts at the same velocity. Another point is that aircraft navigational drift can not be distinguished from sonobuoy drift. In other words, the drift that is perceived by the aircraft is the combination of sonobuoy drift and aircraft navigational drift. If this navigational drift is linear and not excessive it causes few problems. However, non-linear navigational drifts, such as the Schuler cycle, can cause large errors.

B. THE COORDINATE SYSTEM

A right-handed coordinate system was chosen as depicted in figures 3 and 4. This is a departure from the work of Orincon Corporation which is described in the next section. This system allows all angles to be measured positive in the direction they are normally defined, i.e., aircraft heading measured clockwise from north. It also coincides with the usual aerodynamic coordinate system.

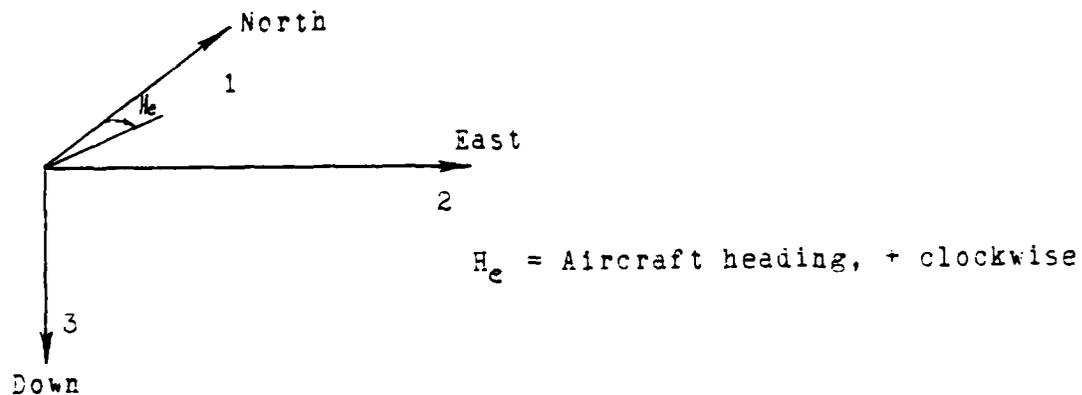


Figure 3. Earth fixed

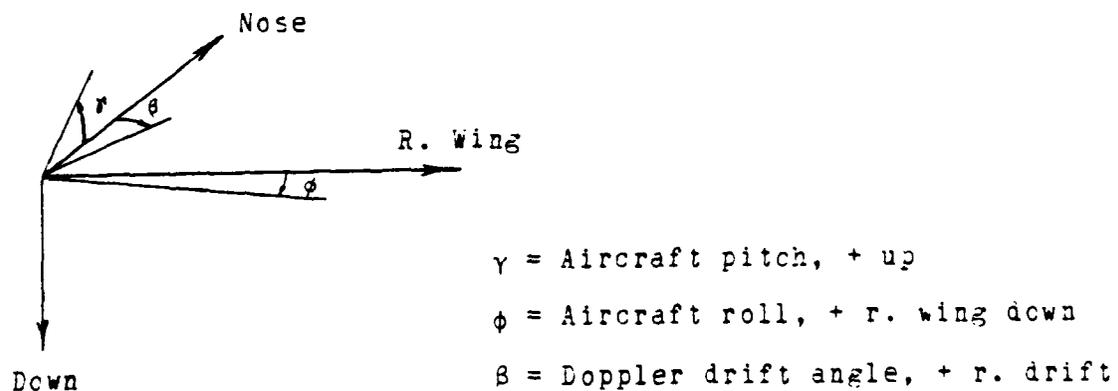


Figure 4. Aircraft fixed

C. THE SIX-STATE SYSTEM

Orincon Corporation of La Jolla, CA., completed a technical report in December of 1978 concerning the Sonobuoy Reference System. They were contracted by NADC to investigate the effect of errors in the aircraft navigational system on the performance of the sonobuoy tracking procedure. As a result of their work they developed a 13-state mathematical model which successfully determined sonobuoy positions in spite of these navigational errors. The complete state vector was

$$X^T = \left[\Delta x_1 \ \Delta x_2 \ \Delta v_1 \ \Delta v_2 \ x_1 \ x_2 \ v_1 \ v_2 \ \psi_1 \ \psi_2 \ \psi_3 \ m_1 \ m_2 \right] \quad (20)$$

where

- Δx = relative position of the sonobuoy from the aircraft;
- Δv = relative velocity of the sonobuoy from the aircraft;
- x = position of the sonobuoy on the ocean surface;
- v = velocity of the sonobuoy on the ocean surface;
- ψ = inertial system's platform azimuth alignment errors
- m = inertial system's gyro drift rates assumed constant.

(Subscripts 1, 2, and 3 represent North, East, and down respectively.) The model was developed using state equations which described sonobuoy motion relative to the aircraft and relative to the ocean's surface. These were derived by

Orincon from

$$\Delta X = X - (X_a^I + E) \quad (21)$$

and other mechanization equations to obtain

$$\begin{aligned} \Delta \dot{x}_1 &= \Delta v_1 \\ \Delta \dot{x}_2 &= \Delta v_2 \\ \Delta \ddot{v}_1 &= -\ddot{R}_1 + \omega^2 (x_1 - x_{a1}^I - \Delta x_1) - \psi_3 \dot{A}_2 + \psi_2 g \\ \Delta \ddot{v}_2 &= -\ddot{R}_2 + \omega^2 (x_2 - x_{a2}^I - \Delta x_2) - \psi_3 \dot{A}_1 - \psi_1 g \end{aligned} \quad (22)$$
$$\begin{aligned} \dot{x}_1 &= v_1 \\ \dot{x}_2 &= v_2 \\ \dot{v}_1 &= \delta \\ \dot{v}_2 &= \delta \end{aligned}$$

where

- x_a^I = aircraft inertial navigation system position
- E = aircraft inertial navigation system error
- R = vector from the center of the earth to the aircraft
- g = acceleration due to gravity (assumed constant)
- ω = Schuler frequency $\sqrt{g/R}$
- A = aircraft accelerometer outputs

The sonobuoys were assumed to drift at a constant velocity and the aircraft was assumed to fly at a constant altitude. The inertial alignment errors were described by

$$\begin{aligned}
\dot{\psi}_1 &= \psi_2 \Omega \sin \lambda - \psi_3 \Omega \cos \lambda + m_1 \\
\dot{\psi}_2 &= -\psi_1 \Omega \sin \lambda + m_2 \\
\dot{\psi}_3 &= \psi_1 \Omega \cos \lambda
\end{aligned}
\tag{23}$$

where

Ω = earth rate

λ = latitude

There were several sources of measurement information. Aircraft to sonobuoy bearing was available from the aircraft interferometer system in the form of $\cos \theta$ where θ is the angle between the direction to the sonobuoy and the base line of the antenna. Doppler velocity and drift angle from the aircraft doppler radar system provided information about Δv_1 and Δv_2 . The only other relevant source of measurement information would have come from GPS (Global Positioning System) which would have provided error free aircraft position.

Orincon performed a numerical observability analysis on this system based on the observability matrix M_k given by

$$M_k = \sum_{i=1}^N \phi_k^T H_k^T R^{-1} H_k \phi_k
\tag{24}$$

If this matrix is positive definite for $N \geq 2$ then the system is considered observable. In addition, an eigenvalue analysis provided information as to the conditioning of the system. The result of their analysis indicated that the

system was not fully observable with measurements from the interferometer and doppler systems alone. Only ten states were observable; sonobuoy position most likely was not observable and ψ_3 , m_1 , m_2 , v_1 and v_2 were weakly observable at best. They also noted that "observability is reduced when the aircraft pursues a straight flight path". By removing sonobuoy positions and gyro drifts from the system they found the observability improved. Further, the system became fully observable if sonobuoy velocity was also removed from the state vector.

The Orincon report presented the results of a Monte-Carlo simulation on the 13-state system. However, they concluded that a reduced state filter made up of

$$X^T = \left[\Delta x_1 \quad \Delta x_2 \quad \Delta v_1 \quad \Delta v_2 \quad \psi_1 \quad \psi_2 \right] \quad (25)$$

would provide a workable solution to the problem although they noted that it had a tendency toward divergence. It was partly the intention of this report to further the investigation of this reduced state filter.

The state equations for the reduced six-state filter can be obtained directly from Equations (22) and (23).

$$\begin{aligned}
 \Delta \dot{x}_1 &= \Delta v_1 \\
 \Delta \dot{x}_2 &= \Delta v_2 \\
 \Delta \dot{v}_1 &= -\ddot{R}_1 + \omega^2(x_1 - x_{a1}^I - \Delta x_1) + \psi_2 g \\
 \Delta \dot{v}_2 &= -\ddot{R}_2 + \omega^2(x_2 - x_{a2}^I - \Delta x_2) - \psi_1 g \\
 \dot{\psi}_1 &= \psi_2 \sin \lambda \\
 \dot{\psi}_2 &= -\psi_1 \sin \lambda
 \end{aligned} \tag{26}$$

where ψ_3 , m_1 , and m_2 are set to zero. In matrix form

$$\dot{X} = \begin{bmatrix} \Delta \dot{x}_1 \\ \Delta \dot{x}_2 \\ \Delta \dot{v}_1 \\ \Delta \dot{v}_2 \\ \dot{\psi}_1 \\ \dot{\psi}_2 \end{bmatrix} = \begin{bmatrix} \emptyset & \emptyset & 1 & \emptyset & \emptyset & \emptyset \\ \emptyset & \emptyset & \emptyset & 1 & \emptyset & \emptyset \\ -\omega^2 & \emptyset & \emptyset & \emptyset & \emptyset & -g \\ \emptyset & -\omega^2 & \emptyset & \emptyset & g & \emptyset \\ \emptyset & \emptyset & \emptyset & \emptyset & \emptyset & -\Omega \sin \lambda \\ \emptyset & \emptyset & \emptyset & \emptyset & \Omega \sin \lambda & \emptyset \end{bmatrix} \begin{bmatrix} \Delta x_1 \\ \Delta x_2 \\ \Delta v_1 \\ \Delta v_2 \\ \psi_1 \\ \psi_2 \end{bmatrix} + \begin{bmatrix} \emptyset \\ \emptyset \\ -\ddot{R}_1 + \omega^2(x_1 - x_{a1}^I) \\ -\ddot{R}_2 + \omega^2(x_2 - x_{a2}^I) \\ \emptyset \\ \emptyset \end{bmatrix}$$

which can be abbreviated

$$\dot{X} = \begin{bmatrix} A \Delta & A \Delta \psi \\ \emptyset & A \psi \end{bmatrix} X + B \tag{27}$$

Expressed in discrete time the state equation is

$$X_{k+1} = \Phi X_k + \Lambda_k + W_k \tag{28}$$

The discrete formulation of the plant matrix can be obtained from

$$\bar{\Phi} = \mathcal{L}^{-1} \left\{ (sI - A)^{-1} \right\} = e^{At} \quad (29)$$

Developing this

$$[sI - A] = \left[\begin{array}{cccc|cc} s & 0 & 1 & 0 & 0 & 0 \\ 0 & s & 0 & 1 & 0 & 0 \\ \omega^2 & 0 & s & 0 & 0 & g \\ 0 & \omega^2 & 0 & s & -g & 0 \\ \hline 0 & 0 & 0 & 0 & s & \Omega \sin \lambda \\ 0 & 0 & 0 & 0 & -\Omega \sin \lambda & s \end{array} \right] \quad (30)$$

from which the determinant of the system is

$$|sI - A| = (s^2 + \omega^2)^2 (s^2 + \Omega^2 \sin^2 \lambda)^2 \quad (31)$$

The inverse of this matrix and its Laplace transform are presented on the following pages as Equations (32) and (33) respectively.

$$[sI - A]^{-1} =$$

$$\begin{array}{c}
 \begin{array}{cccc}
 \frac{s}{s^2 + \omega^2} & 0 & 1 & 0 \\
 0 & \frac{s}{s^2 + \omega^2} & 0 & \frac{1}{s^2 + \omega^2} \\
 \frac{-\omega}{s^2 + \omega^2} & 0 & \frac{s}{s^2 + \omega^2} & 0 \\
 0 & \frac{-\omega^2}{s^2 + \omega^2} & 0 & \frac{s}{s^2 + \omega^2}
 \end{array} \\
 \hline
 \begin{array}{cccc}
 \frac{-g \Omega \sin \lambda}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)} & & & \frac{-g s}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)} \\
 \frac{g s}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)} & & & \frac{-g \sin \lambda}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)} \\
 \frac{-g s \Omega \sin \lambda}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)} & & & \frac{-g s^2}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)} \\
 \frac{g s^2}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)} & & & \frac{-g s \Omega \sin \lambda}{(s^2 + \omega^2)(s^2 + \Omega^2 \sin^2 \lambda)}
 \end{array} \\
 \hline
 \begin{array}{cccc}
 0 & 0 & 0 & 0 \\
 0 & 0 & 0 & 0 \\
 0 & 0 & 0 & 0 \\
 0 & 0 & 0 & 0
 \end{array} \\
 \hline
 \begin{array}{cccc}
 & & & \frac{-\Omega \sin \lambda}{s^2 + \Omega^2 \sin^2 \lambda} \\
 & & & \frac{s}{s^2 + \Omega^2 \sin^2 \lambda} \\
 & & & 0 \\
 & & & 0
 \end{array}
 \end{array}$$

$\cos \omega t$	0	$\frac{1}{\omega} \sin \omega t$	0	$-\frac{g}{\omega} \left[\frac{\Omega \sin \lambda}{\Omega^2 \sin^2 \lambda - \omega^2} \sin \omega t - \sin(\Omega \sin \lambda) t \right]$	$-\frac{g}{\Omega^2 \sin^2 \lambda - \omega^2} \left[\cos \omega t - \cos(\Omega \sin \lambda) t \right]$
0	$\cos \omega t$	0	$\frac{1}{\omega} \sin \omega t$	$-\frac{g}{\Omega^2 \sin^2 \lambda - \omega^2} \left[\frac{\Omega \sin \lambda}{\omega} \sin \omega t - \sin(\Omega \sin \lambda) t \right]$	$-\frac{g}{\Omega^2 \sin^2 \lambda - \omega^2} \left[\cos \omega t - \cos(\Omega \sin \lambda) t \right]$
$-\omega \sin \omega t$	0	$\cos \omega t$	0	$-\frac{g \Omega \sin \lambda}{\Omega^2 \sin^2 \lambda - \omega^2} \left[\cos \omega t - \cos(\Omega \sin \lambda) t \right]$	$\frac{g}{\Omega^2 \sin^2 \lambda - \omega^2} \left[\omega \sin \omega t - \Omega \sin \lambda \sin(\Omega \sin \lambda) t \right]$
0	$-\omega \sin \omega t$	0	$\cos \omega t$	$-\frac{g}{\Omega^2 \sin^2 \lambda - \omega^2} \left[\omega \sin \omega t - \Omega \sin \lambda \sin(\Omega \sin \lambda) t \right]$	$-\frac{g \Omega \sin \lambda}{\Omega^2 \sin^2 \lambda - \omega^2} \left[\cos \omega t - \cos(\Omega \sin \lambda) t \right]$
0	0	0	0	$\cos(\Omega \sin \lambda) t$	$-\sin(\Omega \sin \lambda) t$
0	0	0	0	$\sin(\Omega \sin \lambda) t$	$\cos(\Omega \sin \lambda) t$

The discrete formulation of the continuous matrix B is

$$\Lambda_k = \int_{t_k}^{t_{k+1}} e^{A\lambda} d\lambda B \quad \text{where } \lambda = \Delta t - \tau \quad (34)$$

Hence

$$\Lambda_k = \begin{bmatrix} \left\{ -\frac{\ddot{R}_1}{\omega^2} + (x_1 - x_{a1}^I) \right\} (1 - \cos\omega\Delta t) \\ \left\{ -\frac{\ddot{R}_2}{\omega^2} + (x_2 - x_{a2}^I) \right\} (1 - \cos\omega\Delta t) \\ \left\{ -\frac{\ddot{R}_1}{\omega} + \omega(x_1 - x_{a1}^I) \right\} \sin\omega\Delta t \\ \left\{ -\frac{\ddot{R}_2}{\omega} + \omega(x_2 - x_{a2}^I) \right\} \sin\omega\Delta t \\ 0 \\ 0 \end{bmatrix} \quad (35)$$

The noise term is $W_k \sim N(0, Q)$. The variances are estimated to be .01 NM for relative position, 1.0 (NM/Hr) for relative velocity, and .001 radians for inertial misalignments. As a result, the following constant diagonal matrix is used for Q:

$$Q = \begin{bmatrix} .01 & & & & & \\ & .01 & & & & \\ & & 1.0 & & & \\ & & & 1.0 & & \\ & & & & .001 & \\ & & & & & .001 \end{bmatrix} \quad (36)$$

The same measurement information is available as before

in the larger system, namely interferometer bearings and doppler velocity and drift angle. These introduce non-linearities and must be modeled as in Equation (10).

$$Z_k = h(X_k) + V_k$$

A Taylor series expansion as in Equation (11) about the latest estimate of sonobuoy position provides a method of linearization. As a result

$$H = \left[\begin{array}{cccccc} \frac{\partial z}{\partial \Delta x_1} \Big|_{x=\hat{x}} & \frac{\partial z}{\partial \Delta x_2} \Big|_{x=\hat{x}} & \frac{\partial z}{\partial \Delta v_1} \Big|_{x=\hat{x}} & \frac{\partial z}{\partial \Delta v_2} \Big|_{x=\hat{x}} & \frac{\partial z}{\partial \psi_1} \Big|_{x=\hat{x}} & \frac{\partial z}{\partial \psi_2} \Big|_{x=\hat{x}} \end{array} \right] \quad (37)$$

To determine z and its partial derivatives, recall that

$$z_1 = \cos \theta = \overrightarrow{\Delta X} \cdot \overrightarrow{RB} \quad (38)$$

where $\overrightarrow{\Delta X}$ is the unit vector representation of a line from the aircraft to the sonobuoy and \overrightarrow{RB} is the unit vector describing the base line between the antenna pairs of the interferometer.

$$\begin{aligned} \hat{z}_1 &= \cos \theta \Big|_{x=f} = \frac{\Delta \hat{x}_1}{|\Delta \hat{x}|} RB_1 + \frac{\Delta \hat{x}_2}{|\Delta \hat{x}|} RB_2 + \frac{\Delta \hat{x}_3}{|\Delta \hat{x}|} RB_3 \\ \hat{z}_1 &= \frac{\Delta \hat{x}_1 RB_1 + \Delta \hat{x}_2 RB_2 + h RB_3}{(\Delta \hat{x}_1^2 + \Delta \hat{x}_2^2 + h^2)^{1/2}} \end{aligned} \quad (39)$$

where h is the aircraft altitude.

$$H = \begin{bmatrix} \frac{\partial z_1}{\partial \Delta x_1} \Big|_{x=\hat{x}} & \frac{\partial z_1}{\partial \Delta x_2} \Big|_{x=\hat{x}} & 0 & 0 & 0 & 0 \end{bmatrix} \quad (40)$$

where

$$\frac{\partial}{\partial \Delta x_1} (z_1) \Big|_{x=\hat{x}} = \frac{1}{(\Delta \hat{x}_1^2 + \Delta \hat{x}_2^2 + h^2)^{\frac{1}{2}}} \left[RB_1 - \frac{\Delta \hat{x}_1 (\Delta \hat{x}_1 RB_1 + \Delta \hat{x}_2 RB_2 + h RB_3)}{(\Delta \hat{x}_1^2 + \Delta \hat{x}_2^2 + h^2)} \right] \quad (41)$$

$$\frac{\partial}{\partial \Delta x_2} (z_1) \Big|_{x=\hat{x}} = \frac{1}{(\Delta \hat{x}_1^2 + \Delta \hat{x}_2^2 + h^2)^{\frac{1}{2}}} \left[RB_2 - \frac{\Delta \hat{x}_2 (\Delta \hat{x}_1 RB_1 + \Delta \hat{x}_2 RB_2 + h RB_3)}{(\Delta \hat{x}_1^2 + \Delta \hat{x}_2^2 + h^2)} \right] \quad (42)$$

The doppler velocity, V_d , is a measurement of aircraft speed relative to the ocean surface and doppler drift angle, β , is a measure of the angle between aircraft track and aircraft heading. These measurements can be related to the state vector by assuming the sonobuoys drift at the same velocity as the ocean surface. If this is true then

$$z_2 = V_d = (\Delta v_1^2 + \Delta v_2^2)^{\frac{1}{2}} \quad (43)$$

Expanding in a Taylor series yields

$$z_2 = (\Delta \hat{v}_1^2 + \Delta \hat{v}_2^2)^{\frac{1}{2}} \quad (44)$$

$$H_2 = \begin{bmatrix} 0 & 0 & \frac{\partial z_2}{\partial \Delta v_1} \Big|_{x=\hat{x}} & \frac{\partial z_2}{\partial \Delta v_2} \Big|_{x=\hat{x}} & 0 & 0 \end{bmatrix} \quad (45)$$

where

$$\frac{\partial z_2}{\partial \Delta v_1} \Big|_{x=\hat{x}} = \frac{\Delta \hat{v}_1}{(\Delta \hat{v}_1^2 + \Delta \hat{v}_2^2)^{\frac{1}{2}}} \quad (46)$$

$$\frac{\partial z_2}{\partial \Delta v_2} \Big|_{x=\hat{x}} = \frac{\Delta \hat{v}_2}{(\Delta \hat{v}_1^2 + \Delta \hat{v}_2^2)^{\frac{1}{2}}} \quad (47)$$

Similarly,

$$\begin{aligned} z_3' &= \beta = \text{Track} - \text{Heading}_e \\ &= \tan^{-1}\left(\frac{\Delta v_2}{\Delta v_1}\right) - H_e \end{aligned} \quad (48)$$

or since the aircraft heading, H_e , is deterministic

$$z_3 = H_e + \beta = \tan^{-1}\left(\frac{\Delta v_2}{\Delta v_1}\right) \quad (49)$$

$$H_3 = \begin{bmatrix} 0 & 0 & \left. \frac{\partial z_3}{\partial \Delta v_1} \right|_{x=\hat{x}} & \left. \frac{\partial z_3}{\partial \Delta v_2} \right|_{x=\hat{x}} & 0 & 0 \end{bmatrix} \quad (50)$$

where

$$\left. \frac{\partial z_3}{\partial \Delta v_1} \right|_{x=\hat{x}} = \frac{-\Delta \hat{v}_2}{\Delta \hat{v}_1^2 + \Delta \hat{v}_2^2} \quad (51)$$

$$\left. \frac{\partial z_3}{\partial \Delta v_2} \right|_{x=\hat{x}} = \frac{\Delta \hat{v}_1}{\Delta \hat{v}_1^2 + \Delta \hat{v}_2^2} \quad (52)$$

Finally,

$$Z = \begin{bmatrix} \cos \theta \\ v_d \\ H_e - \beta \end{bmatrix} \quad (53)$$

$$H = \begin{bmatrix} \frac{\partial(\cos \theta)}{\partial x_1} & \frac{\partial(\cos \theta)}{\partial x_2} & 0 & 0 & 0 & 0 \\ 0 & 0 & \frac{\partial(v_d)}{\partial \Delta v_1} & \frac{\partial(v_d)}{\partial \Delta v_2} & 0 & 0 \\ 0 & 0 & \frac{\partial(H_e + \beta)}{\partial \Delta v_1} & \frac{\partial(H_e + \beta)}{\partial \Delta v_2} & 0 & 0 \end{bmatrix} \quad (54)$$

The measurement noise, V_d , is described by $N(0, R)$. It is assumed that the measurement noise R and the plant noise Q are white, gaussian, and uncorrelated. Based on information obtained from NADC, R was assigned the following values:

$$\begin{aligned} R_1 &= .0007 \text{ rad} && (\text{interferometer}) \\ R_2 &= 1.0 \text{ (NM/Hr)} && (\text{doppler velocity}) \\ R_3 &= .01 \text{ deg} && (\text{doppler drift angle}) \end{aligned} \quad (55)$$

It should be pointed out that the doppler measurements are not always available. The doppler system freezes the last value of groundspeed and drift angle anytime the doppler radar is not receiving good information. This happens whenever the aircraft is above 5000' altitude, or can happen when flying above a cloud cover or when the sea surface is too smooth. As a result, the doppler could be inaccurate during a significant portion of the flight.

The six-state system was programmed and tested for several conditions. Flow charts for significant portions of the program and the entire program listing are presented in Appendix A. The main program has two purposes: first, to control the simulation; and second, to read and prepare measurement data for the subroutine FILTER. This subroutine performs all the Kalman filter computations as specified in the above equations. The Carlson square root technique is used to ensure a positive definite covariance matrix. Also, in order to avoid inverting a matrix in the calculations for the Kalman gain, the three measurements are processed one at

a time. The results of the simulation are presented in section IV.B.

D. THE TWO-STATE SYSTEM

An attempt was made earlier on in this research to model sonobuoy motion as

$$x_{k+1} = x_k + U_k \Delta t \quad (56)$$

$$U_{k+1} = U_k$$

with a state vector made up of x_1 , x_2 , u_1 , and u_2 . This system proved to be unobservable although the simple approach was appealing. Since sonobuoy drift rates are generally slow, another attempt was made to model the system without velocity. The equations reduced to

$$x_{k+1} = x_k + w_k \quad (57)$$

With only x_1 and x_2 as states, this system was observable. By describing large statistical values of system noise there would hopefully be enough freedom in the "update" to compensate for any sonobuoy drifts. Unfortunately, this approach was not completely successful; however, it was observed that the general direction of sonobuoy drift was correct. As a result of these investigations a technique of cascading Kalman filters was used to solve the tracking problem. The intent was to use results from one filter as measurement for a second filter. Then, the result of the

second filter could be used as a deterministic control input to better propagate the first filter.

Two Kalman filters are used in this approach. Figure 5 shows how they are related (or cascaded). The first filter models sonobuoy position and is based on the discrete state equation

$$\dot{X}_k = X_k + U_k \Delta t + W_k \quad (58)$$

The state vector consists of the sonobuoy positions, x_1 and x_2 . U_k is a deterministic input accounting for sonobuoy drift and W_k is system noise. The interferometer measurement, $\cos \theta$, is used to update this filter.

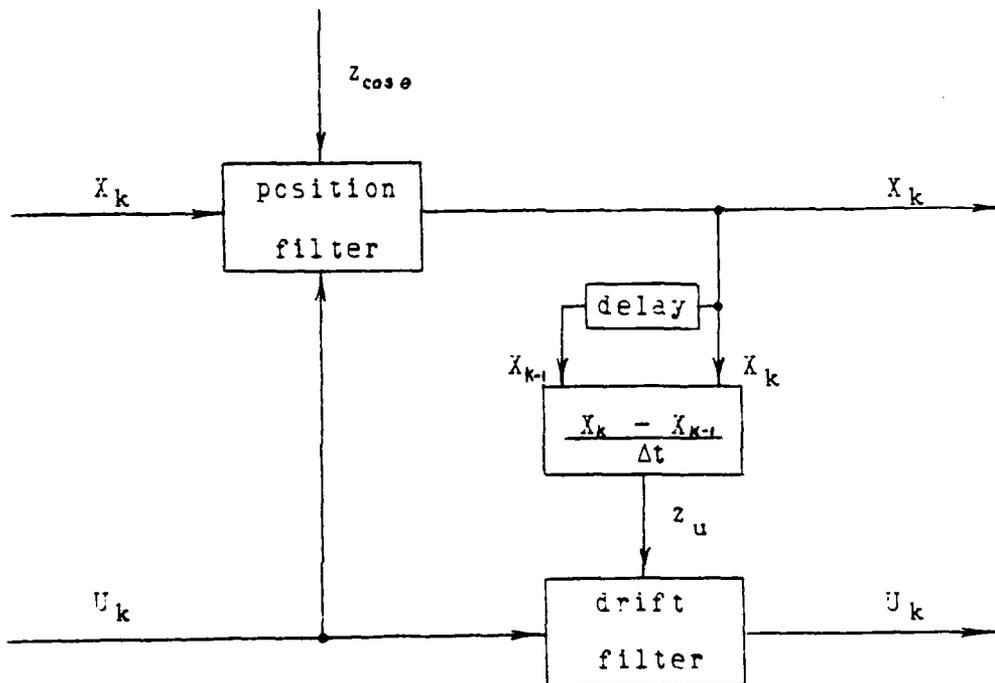


Figure 5. Cascaded Kalman filters.

The second filter models the sonobuoy drift with the state equation

$$U_{k+1} = U_k \quad (59)$$

As the sonobuoy's position changes with each update of the position filter, the velocity required to move from X_k to X_{k+1} in the time interval Δt can be calculated. This value is used as a measurement for the drift filter. All sonobuoys contribute to this filter which produces one overall estimate of drift for the sonobuoy field. This value is then used as an input to the position filter when the next interferometer measurement occurs.

The technique is understandably sensitive to the output of the drift filter. If the estimated drift is too much in error, the estimated positions will be affected. Divergence could result. For this reason, a simplified Kalman filter is used to estimate the drift. It is of the form

$$U_{k+1} = U_k + G_k (Z_k - U_k) \quad (60)$$

The gain is precomputed from $1/(n+1)$ where n is the total number of measurements and is limited to a value at least as great as $\Delta t/(84 \text{ min.})$. This limit is reached after the first 84 minutes of tracking time. From that point on the gain remains constant effectively averaging the drift over one period of the Schuler cycle. It must be understood that

the drift this filter estimates consists not only of sonobuoy drift but aircraft drift as well and is, in fact, the vector sum of these. An estimate which attempts to follow the changing velocities through the Schuler cycle tends to cause the position filters to diverge. This might be the case if larger values of gain are used, averaging only the last few measurements of drift. Therefore, many measurements of drift are averaged which provides an estimate for linear drift only. The lower limit of gain prevents new drift measurements from being ignored and also provides for flexibility in estimating the drift.

Unlike the six-state system, the interferometer measurement is the only measurement required by the position filter. The measurement equation is essentially the same as in the preceding section where z is determined as shown in Equation (39). Since there is only one term in z , the H matrix is a row vector consisting of

$$H = \left[\frac{\partial z}{\partial x_1} \quad \frac{\partial z}{\partial x_2} \right] \quad (61)$$

where the terms are given in Equations (41) and (42). It is necessary to determine the aircraft-to-sonobuoy range for these values from

$$\Delta X = \hat{X} - X_a^I \quad (62)$$

where \hat{X} is the estimated sonobuoy position. X_a^I represents

the aircraft position and is considered deterministic. Undoubtedly, x_a^I is in error due to aircraft navigational drifts and therefore causes errors in the estimated geographical position of the sonobuoy. But, since the measurement of bearing is based on relative positions, the estimate is relatively correct.

An observability analysis on the 2-state position filter was performed again using

$$M_k = \sum_{k=1}^N \Phi_k^T H_k^T R^{-1} H_k \Phi_k \quad (63)$$

Substitution into this formula with $N=2$ yields

$$M = \begin{bmatrix} \left(\frac{\partial z}{\partial x_{11}}\right)^2 & \left(\frac{\partial z}{\partial x_{11}}\right)\left(\frac{\partial z}{\partial x_{21}}\right) \\ \left(\frac{\partial z}{\partial x_{11}}\right)\left(\frac{\partial z}{\partial x_{21}}\right) & \left(\frac{\partial z}{\partial x_{21}}\right)^2 \end{bmatrix} + \begin{bmatrix} \left(\frac{\partial z}{\partial x_{11+1}}\right)^2 & \left(\frac{\partial z}{\partial x_{21+1}}\right)\left(\frac{\partial z}{\partial x_{11+1}}\right) \\ \left(\frac{\partial z}{\partial x_{11+1}}\right)\left(\frac{\partial z}{\partial x_{21+1}}\right) & \left(\frac{\partial z}{\partial x_{21+1}}\right)^2 \end{bmatrix} \quad (64)$$

Expanding,

$$M = \begin{pmatrix} \left(\frac{\partial z}{\partial x_{11+1}}\right)^2 & \left(\frac{\partial z}{\partial x_{21}}\right)^2 \\ \left(\frac{\partial z}{\partial x_{11+1}}\right)\left(\frac{\partial z}{\partial x_{21+1}}\right) & \left(\frac{\partial z}{\partial x_{11}}\right)\left(\frac{\partial z}{\partial x_{21}}\right) \end{pmatrix} - \begin{pmatrix} \left(\frac{\partial z}{\partial x_{11+1}}\right)\left(\frac{\partial z}{\partial x_{21+1}}\right) & \left(\frac{\partial z}{\partial x_{11}}\right)\left(\frac{\partial z}{\partial x_{21}}\right) \\ \left(\frac{\partial z}{\partial x_{11}}\right)\left(\frac{\partial z}{\partial x_{21}}\right) & \left(\frac{\partial z}{\partial x_{11+1}}\right)\left(\frac{\partial z}{\partial x_{21+1}}\right) \end{pmatrix} + \begin{pmatrix} \left(\frac{\partial z}{\partial x_{11}}\right)^2 & \left(\frac{\partial z}{\partial x_{21+1}}\right)^2 \\ \left(\frac{\partial z}{\partial x_{11}}\right)\left(\frac{\partial z}{\partial x_{21+1}}\right) & \left(\frac{\partial z}{\partial x_{11+1}}\right)\left(\frac{\partial z}{\partial x_{21+1}}\right) \end{pmatrix} \quad (65)$$

This matrix is positive definite for all except the case when

$$\left(\frac{\partial z}{\partial x_{1i}}\right) = \left(\frac{\partial z}{\partial x_{1i+1}}\right) \quad \text{and} \quad \left(\frac{\partial z}{\partial x_{2i}}\right) = \left(\frac{\partial z}{\partial x_{2i+1}}\right)$$

This occurs whenever the relative bearing from the aircraft to the sonobuoy is not changing (i.e. when flying directly toward or away from the sonobuoy). This is intuitively correct since two or more bearings must cross in order to determine a position.

The two-state system was programmed and tested to determine the usefulness of this simplified filter. The majority of the program is the same as the one used for the six-state system. The significant differences occur in subroutine FILTER. The Carlson square root technique is used again but only one measurement instead of three is processed. The drift filter is programmed in this subroutine along with the position filter. A flow chart and program listing are presented in Appendix A.

IV. ANALYSIS

A. THE SIMULATION

Actual data from the Orion was not easily obtainable for this research. Consequently, a computer program had to be written to generate the information required by these algorithms. A flow chart in Appendix B describes the program. An aircraft track was created by alternating lines and curves of various lengths and then a determination of noise free measurements was made as the track was flown. (Measurement noise was added later during the simulations.) Sonobuoys were allowed to drift at a constant velocity, and the aircraft's navigational drift was modeled with a constant velocity and Schuler cycle variations as

$$X_a^I = X_a + K_x \Delta t + A_x \sin \omega t \quad (66)$$

where

X_a^I = aircraft inertial position

X_a = aircraft true position

ω = Schuler frequency $\sqrt{g/R}$

The constant drift rate, K_x , was found by NADC to have 0 mean and a standard deviation of 2.5 NM/HR. Likewise, the amplitude of the Schuler cycle, A_x , had 0 mean and a standard deviation of .5 NM. The effect of wind was also available in the program but was never included for analysis.

It was obvious that both algorithms were extremely sensitive to the aircraft flight path. Since there is no "typical" flight path for an Orion during its on-station period, several uniform patterns were selected which would provide meaningful information on each algorithm's performance. These base parameters were chosen:

airspeed	180 kts
altitude	3000 ft
range	15 NM aircraft-to-sonobuoy
frequency	20 sec between measurements on one sonobuoy

A circular pattern was flown around a sonobuoy at a range of about 15 NM. Initial sonobuoy placement was at $(1,1)$ ¹ with the aircraft flying clockwise starting at $(15,0)$. This is shown in Figure 6 which depicts a portion of the aircraft's true track. Sonobuoy drift was west at 1 NM/HR; aircraft navigational drift was south at 1 NM/HR and included a Schuler cycle (amplitude .5 NM) when specified. (This drift would be indicated in Figures 12 thru 30 by the following notation: DR=(a/c 180-1 + Schuler, b 270-1) .) The sonobuoy remained near the center of the pattern allowing the range to remain relatively constant. This was desirable since this basic flight path was used to test the response of the algorithms to variations in altitude and range. Similarly, a square pattern was flown counter-clockwise around a

¹ (x_1, x_2) measured in nautical miles from an arbitrary origin near the aircraft's starting point.

sonobuoy at a range of 15 NM. In this case the aircraft began on-top the initial position of the sonobuoy at (-3,-3) and proceeded to fly the track shown in Figure 7. The sonobuoy was allowed to drift to the east at 2 NM/HR and the aircraft's navigational system to the north at 2.5 NM/HR. This pattern was used eventually to analyze performance at different measuring frequencies. Figure 8 shows the output of the aircraft's navigational system when it has a northerly drift of 2.5 NM/HR while the aircraft is flying the square pattern. Figure 9 shows this same pattern when a Schuler cycle with an amplitude of .5 NM is also present.

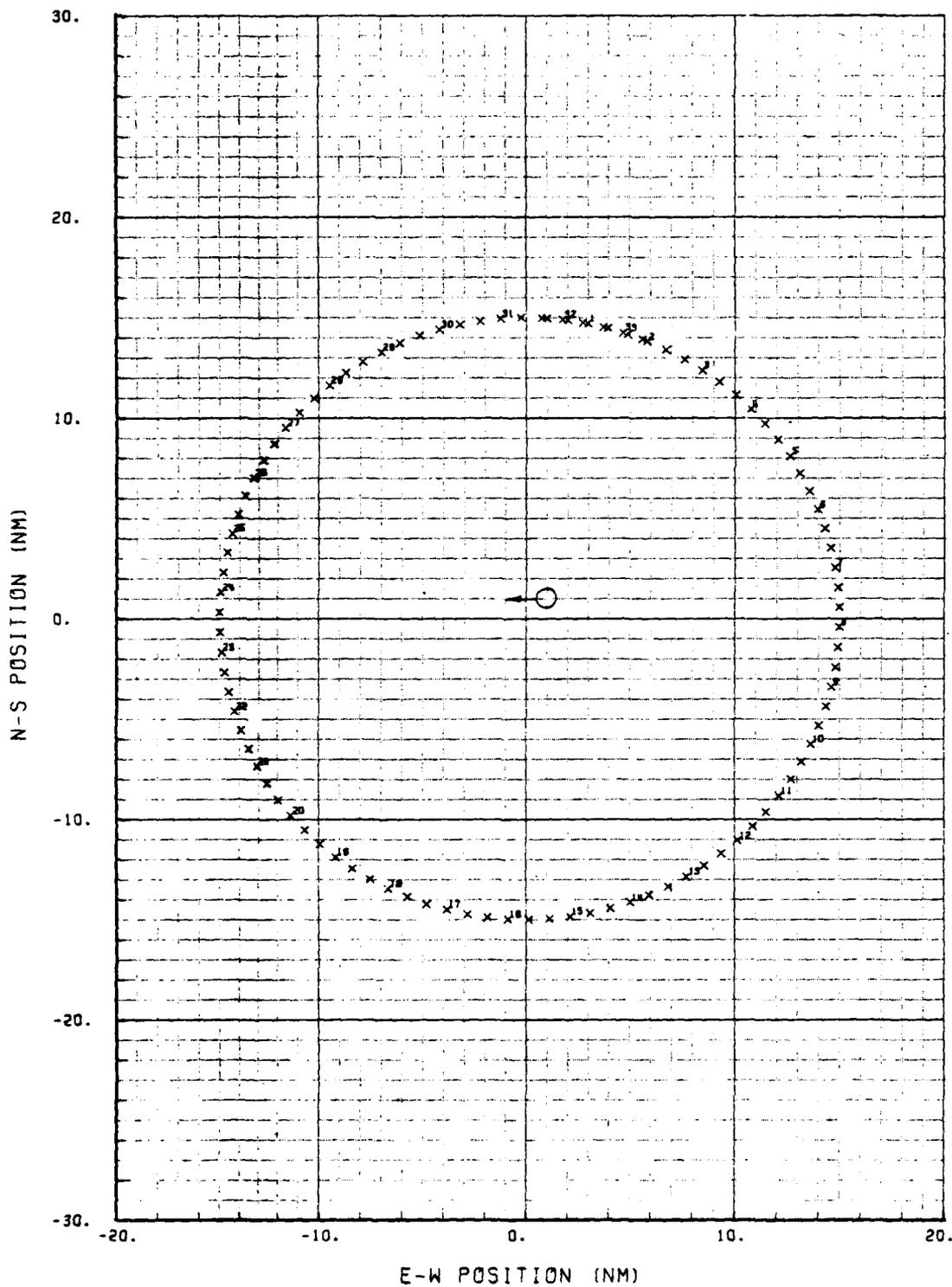


Figure 6. Aircraft track for the 15 NM circular pattern with initial sonobuoy location and direction of drift indicated.

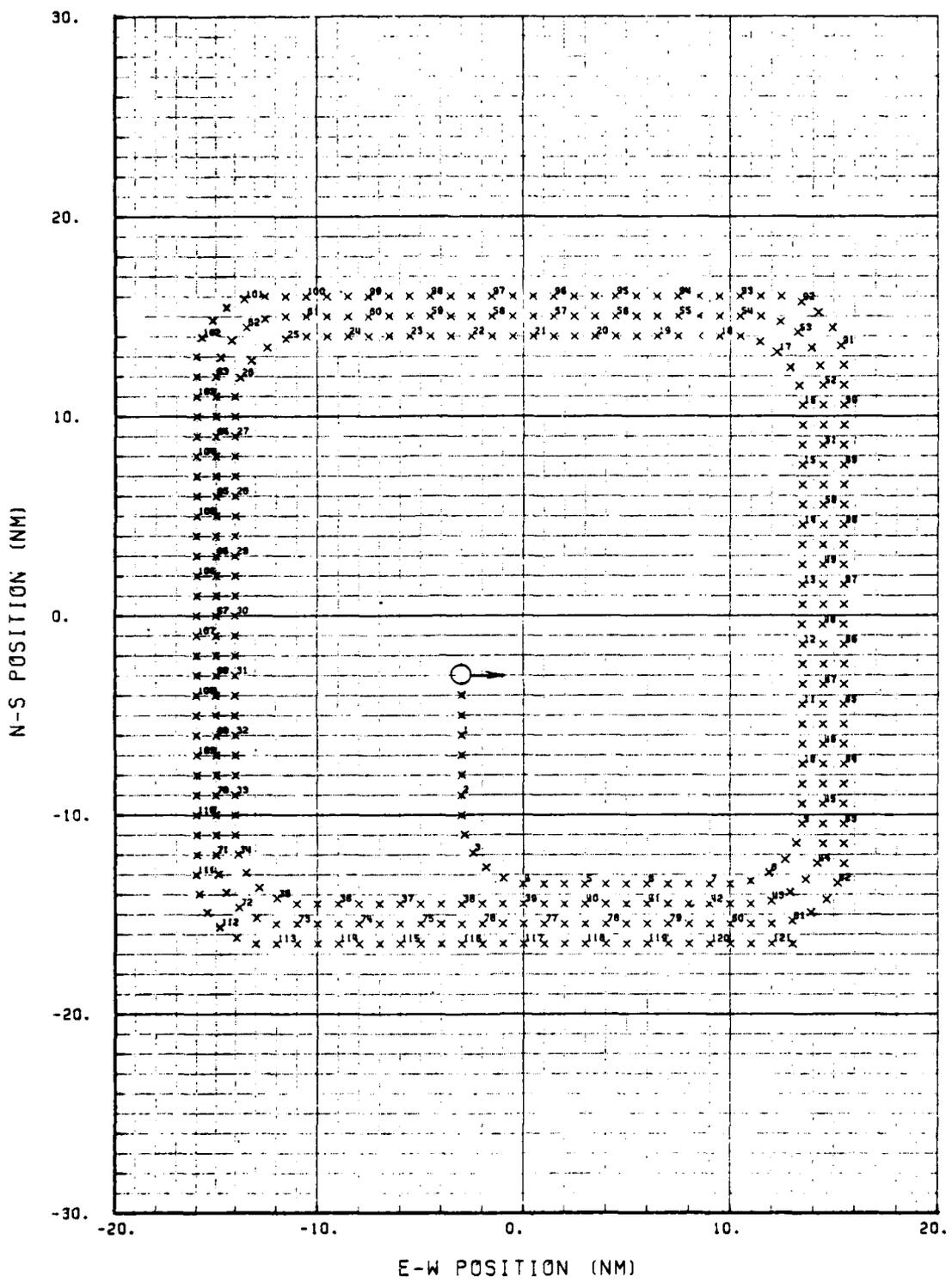


Figure 7. Aircraft track for the 15 NM square pattern with initial sonobuoy location and direction of drift indicated.

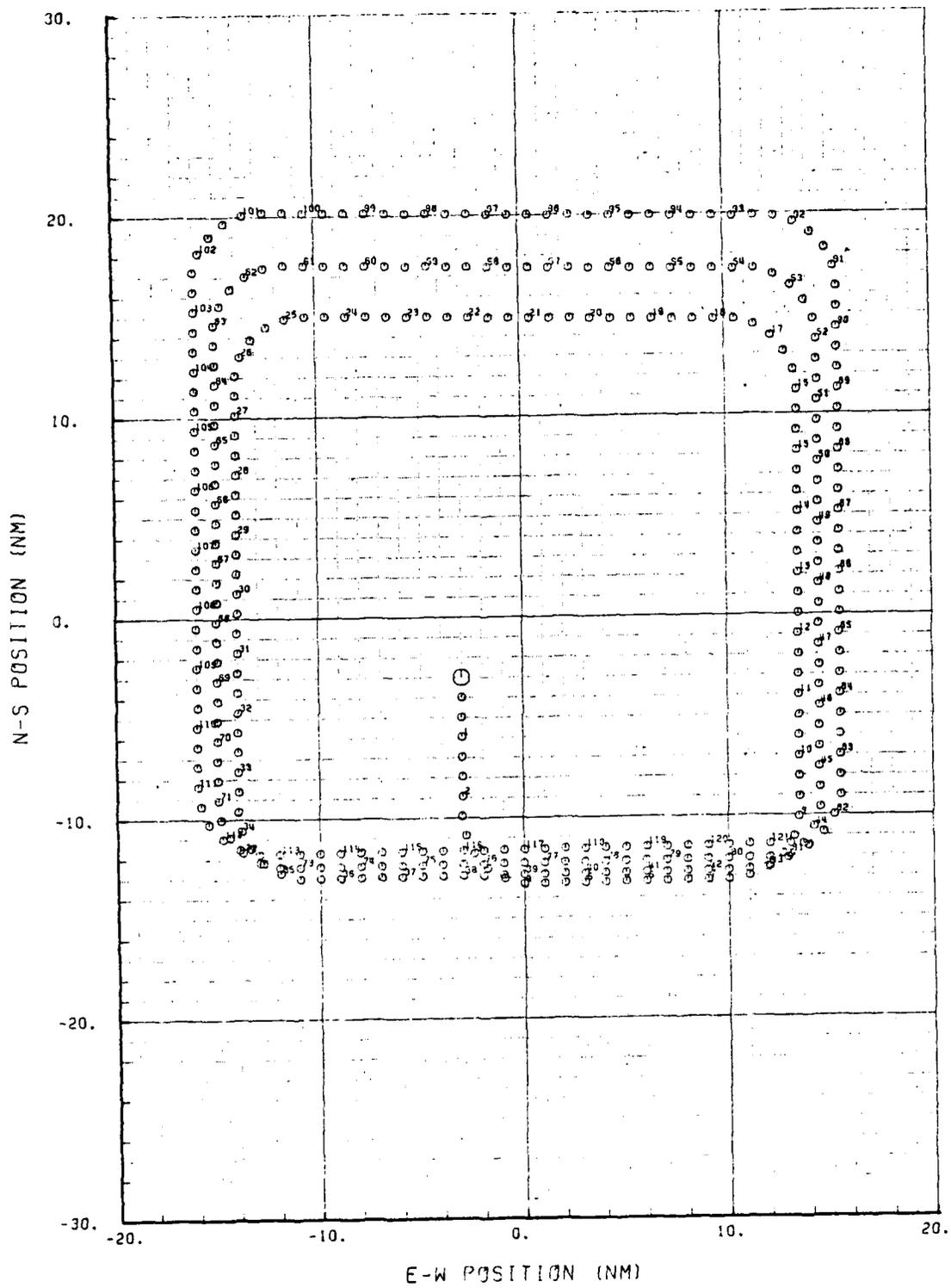


Figure 8. Aircraft's navigational output for the 15 NM square; navigational drift is 2.5 NM/Hr to the north.

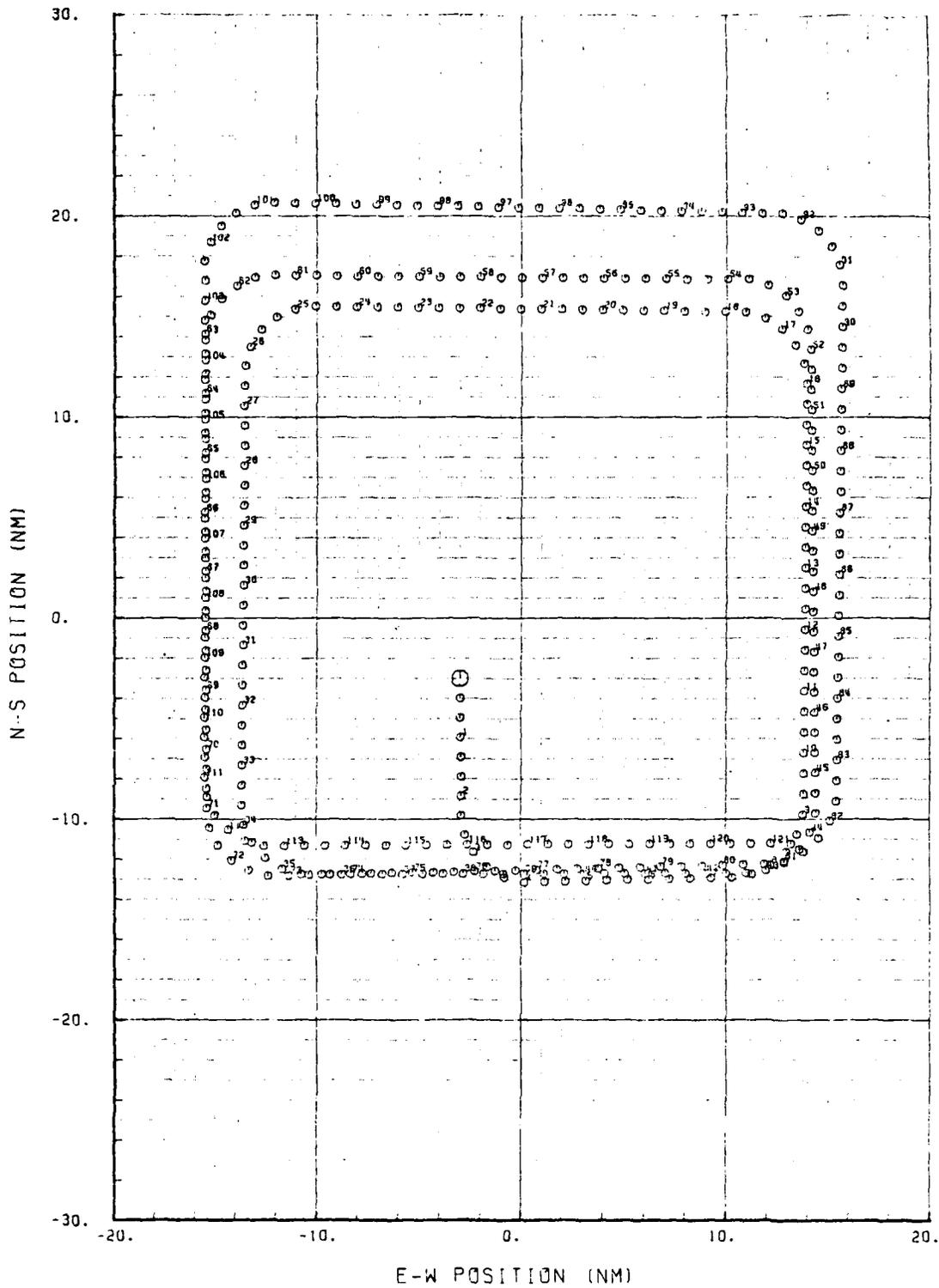


Figure 9. Aircraft's navigational output for the 15 NM square; navigational drift is 2.5 NM/Hr north with a Schuler cycle.

These patterns were used to generate the data which included sonobuoy number, time of measurement and measurement, aircraft latitude and longitude, altitude, heading, pitch angle, roll angle, N-S acceleration, E-W acceleration, and the antenna used for the measurement. Also determined, but used only for error analysis, were sonobuoy position and true aircraft position. This information was computed for the entire simulation period and stored in a file to be used when needed.

Initial estimates of the state and the square root of the covariance matrices were obtained as follows:

Six-state

$$X(\varnothing) = \begin{bmatrix} x_1(\varnothing) - x_{a_1}(\varnothing) \\ x_2(\varnothing) - x_{a_2}(\varnothing) + .5 \\ -V_a \cosh \bar{u} \\ -V_a \sinh \bar{u} \\ \varnothing \\ \varnothing \end{bmatrix} \quad S_D(\varnothing) = \begin{bmatrix} 5. \\ 5. \\ 5. \\ 5. \\ .0025 \\ .0025 \end{bmatrix} \quad (67)$$

Two-state

$$X(\varnothing) = \begin{bmatrix} x_1(\varnothing) \\ x_2(\varnothing) + .5 \end{bmatrix} \quad S_D(\varnothing) = \begin{bmatrix} \text{Alt} \\ \text{Alt} \end{bmatrix} \quad (68)$$

where $S_D(\varnothing)$ indicates the diagonal elements of $S(\varnothing)$.

For the circular pattern

$$x_1(0) = 1.$$

$$x_2(0) = 1.$$

$$x_{a_1}(0) = 15.$$

$$x_{a_2}(0) = 0.$$

and for the square pattern

$$x_1(0) = -3.$$

$$x_2(0) = -3.$$

$$x_{a_1}(0) = -3.$$

$$x_{a_2}(0) = -3.$$

The initial values of the covariance were chosen so as to describe the errors in the state associated with dropping a sonobuoy from an aircraft. The two-state system operated better with a lower initial variance than did the six-state system.

Simulations were run for a nominal period of two hours. Each unit of data was sequentially read into the simulation program from the storage file. Measurement noise was added with the following normal distributions:

ccsθ N(0,.0007) in radians

doppler velocity N(0,1.0) in NM/ER

doppler drift angle N(0,.01) in degrees

These values agree with the ones chosen for R as described previously. A random number generator using an initial seed was used to create the noise from the proper distributions. After the estimate was made, errors were measured and manipulated in subroutine RESULT. This cycle was repeated

until the simulation was complete. (Refer again to Appendix A for a flow diagram of the simulation process.)

Information was gathered and is presented primarily in two ways. First, for both the square and circular patterns the results of the first run of the simulation are plotted. (Figure 13 is an example.) The top portion of the results shows the estimated positions of the sonobuoy on the aircraft's navigational plot. It should be noted that these are not true positions since inertial errors may be causing this plot to drift. The bottom portion shows the North-South and East-West errors in the estimated positions. These errors are the differences between the true and the estimated relative positions of the sonobuoy. They are depicted in Figure 10 and were computed from

$$\Delta \tilde{X}_k = \hat{\Delta X}_k - \Delta X_k \quad (69)$$

$$\Delta \tilde{X}_k = (\hat{X} - X_a^I)_k - (X - X_a)_k \quad (70)$$

where $\Delta \tilde{X}_k$ is the relative error at time k. The true aircraft

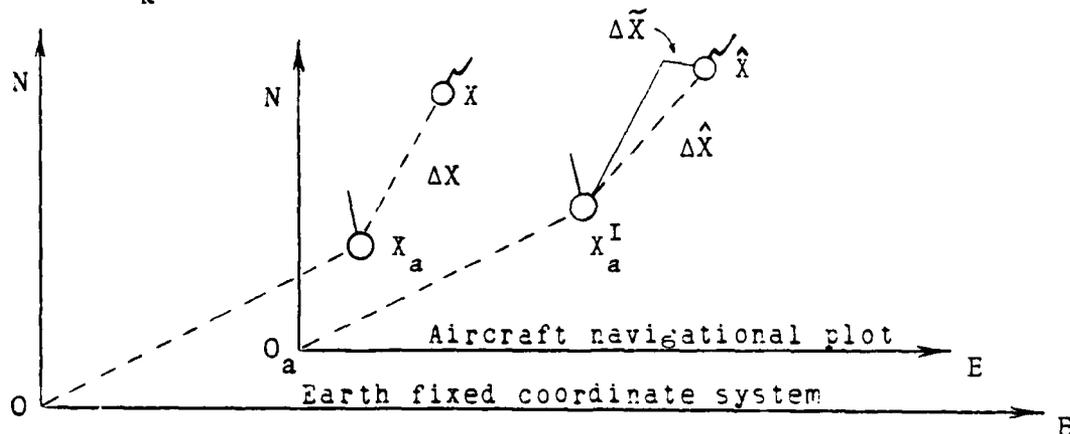


Figure 10. The effect of the aircraft navigational plot

position is denoted by X_a and the true geographical position of the sonobuoy is denoted by X . These values are known from the simulated data. \hat{X} is the filter's estimate of position and X_a^I is a deterministic input from the aircraft's navigational system. The errors indicated in these plots are different from those perceived in the upper plot whenever the aircraft's navigational picture is drifting. The positive and negative values of the square root of the covariance are also shown on these plots as solid lines. Specifically,

$$\begin{aligned} \text{N-S error} &= \pm \sqrt{P_{11}} \\ \text{E-W error} &= \pm \sqrt{P_{22}} \end{aligned}$$

(Only position errors are analyzed in this report.)

The second way in which the information was gathered consisted of a shortened Monte-Carlo simulation. For each scenario under study, 20 two-hour simulations were run, each with new values of measurement noise provided by the random number generator from their respective distributions. The objective was to compare the RMS errors predicted by the covariance matrix in the filters to the actual RMS errors observed in the simulations. Three RMS statistics were collected and plotted versus time as follows:

1. The square root of the covariance computed by the Kalman filter is represented by a solid line on the plots. It was computed from

$$\sigma_{f_k} = \sqrt{\frac{1}{n} \sum P_{11} + \frac{1}{n} \sum P_{22}}, \quad n=20, \quad k = 1, 2, 3 \dots \quad (71)$$

This standard deviation is the filter's estimate of its accuracy.

2. The mean and standard deviation of the error in the estimated sonobuoy position relative to the aircraft was determined by

$$\mu_k = \frac{1}{n} \sum_{i=1}^n \Delta \tilde{X}_{ki}, \quad n = 20, k = 1, 2, 3 \dots \quad (72)$$

$$\sigma_{x_k}^2 = \frac{1}{n} \sum_{i=1}^n \Delta \tilde{X}_{ki}^2 - \mu_k^2, \quad n = 20, k = 1, 2, 3 \dots \quad (73)$$

This standard deviation is represented by an 'X' on the plot and is a measure of the variability in the filter's estimated position.

3. The RMS value of the actual relative error was also determined. It was computed as

$$\sigma_{a_k}^2 = \frac{1}{n} \sum_{i=1}^n \Delta \tilde{X}_{ki}^2, \quad n = 20, k = 1, 2, 3 \dots \quad (74)$$

and is represented by an 'O' on the plot. This value is a measure of the error which occurred between the actual relative position of the sonobuoy and the estimated one. (See Figure 11.)

These values are plotted over the two hour simulation period ($k = 1, 2, 3 \dots$) and provide a measure of the accuracy obtained by each of the algorithms.

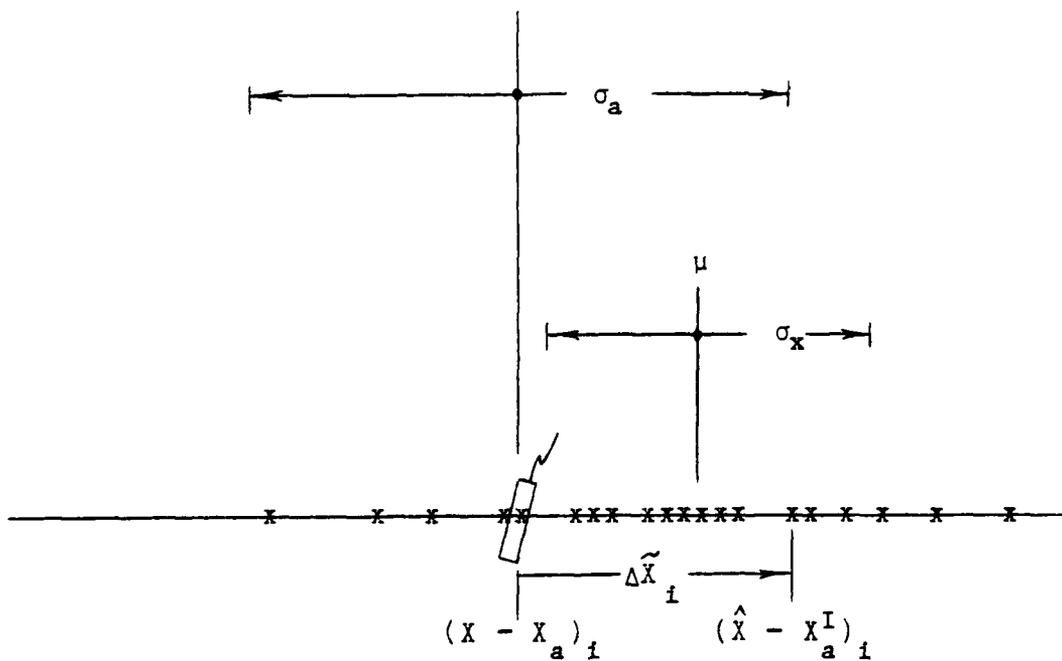


Figure 11. RMS statistics at time = k.

B. THE RESULTS

Orincon Corporation performed much of their analysis using the small circular pattern shown in Figure 12. The aircraft sped around this track at 388 knots taking a measurement every 20 seconds. Altitude was constant at 3000 feet. The initial position of the sonobuoy was at (2,2) and it drifted on a heading of 045 degrees at 5.5 NM/HR while the aircraft's navigational system drifted in the opposite direction at 5.2 NM/HR. A Schuler cycle was also superimposed on this drift.

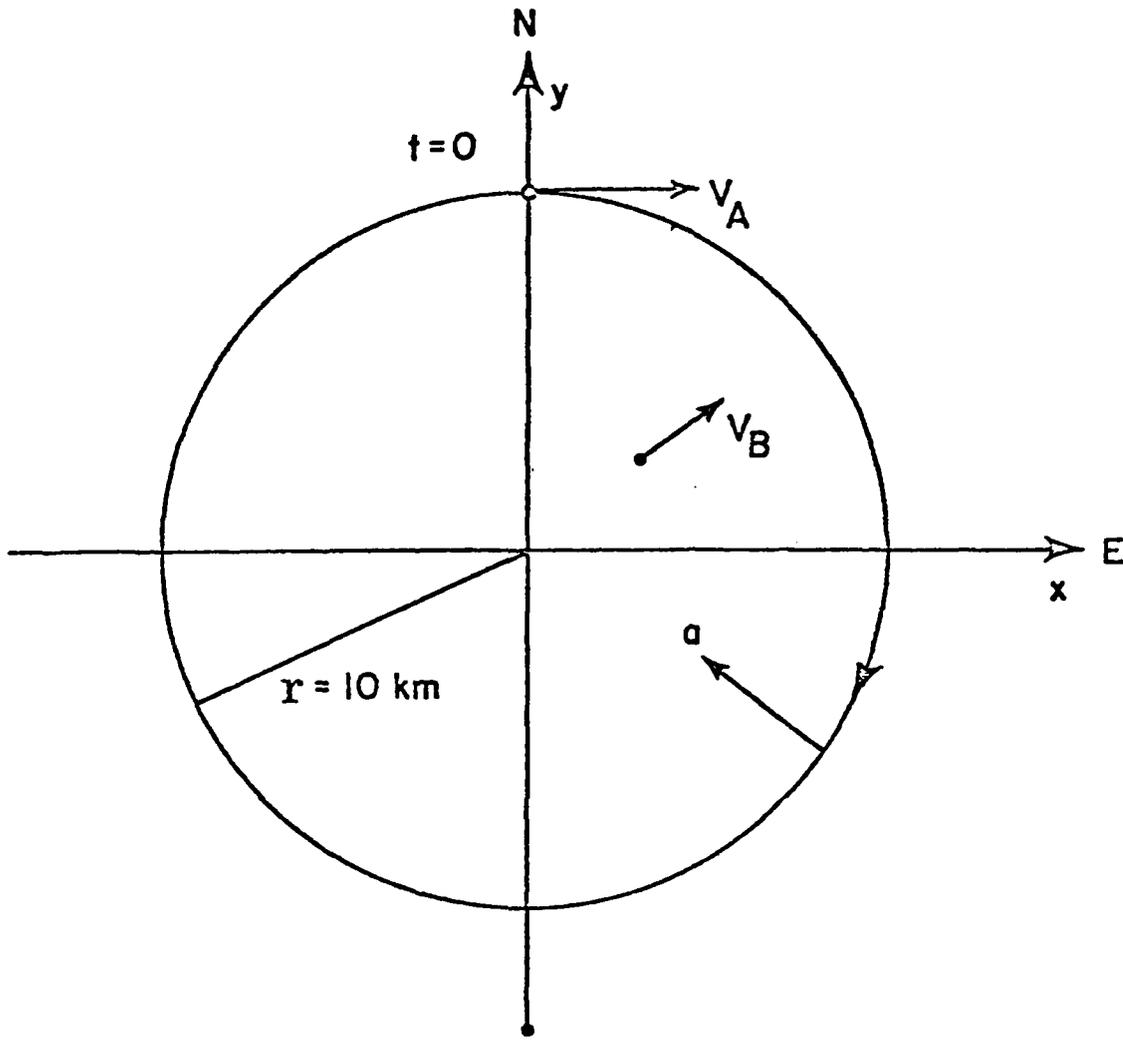
The initial estimates of sonobuoy position and variances used by the algorithms were different for this particular pattern in order to coincide with the Orincon simulation as much as possible. They were

Two-state

$$X(\delta) = \begin{bmatrix} \delta \\ \delta \end{bmatrix} \quad S_D(\delta) = \begin{bmatrix} 2. \\ 2. \end{bmatrix} \quad (74)$$

Six-state

$$X(\delta) = \begin{bmatrix} -5.4 \\ \delta \\ -V_a \cosh H \\ -V_a \sinh H \\ \delta \\ \delta \end{bmatrix} \quad S_D(\delta) = \begin{bmatrix} 2. \\ 2. \\ 5. \\ 5. \\ .0025 \\ .0025 \end{bmatrix} \quad (75)$$



CIRCLE

$$a = -4 \text{ M/S}^2 = -0.4 \text{ g}$$

$$v_A = 200 \text{ M/S}$$

$$v_B = 2.8 \text{ M/S}$$

Figure 12. Orincon's simulation flight path.

Initially, both the six-state and the two-state algorithms were tested using this flight path and comparisons were made with the results of Orincon's 13-state system. It was discovered that the six-state system's results did not change as a consequence of whether or not a Schuler cycle existed in the navigational system. Therefore, the six-state system was run only with the Schuler cycle active. The two-state system was run with and without a Schuler cycle.

The results of the first run of the simulation are shown in Figures 13, 14, and 15. The sawtooth shape of the covariance (solid lines) is the result of the circular pattern and is caused in two ways. Primarily, since the sonobuoy remains closer to one side of the circle, there is a minimum value of the variance each time the aircraft passes on this side. A second reason for a change in variance is the aircraft's location in the pattern. However, these modulations are not as apparent since the range variation dominates. RMS errors were obtained by repeating these runs 20 times and are shown in Figures 16 thru 18. The fluctuations generated by the flight path are apparent, more so in the six-state system than in the two-state system. Also, steady state values of the error gradually increase. This is true because the average range to the sonobuoy increases as the buoy moves farther away

from the center of the pattern. (It will be demonstrated later that the errors are range dependent.) The covariance of the six-state system increases from 400 yards at 25 minutes to 800 yards at the end of the simulation. The two-state system increases slowly from 250 yards until it is hampered by the Schuler cycle. It is worthy of note that without the navigational errors stemming from the Schuler cycle the RMS errors of the two-state system are considerably lower than those of the six-state system.

Table 1 compares the RMS errors of these two systems to the results obtained by Orincon for their 13-state system. (The values for the two-state and six-state systems are taken from the covariance at a point 35 minutes into the simulation.) The first line considers measurements of bearing, doppler velocity, and drift angle. Navigational errors are present including a Schuler cycle. As expected there is an increase in relative position error because of the reduction from 13 to 6 states. The next line considers the effect of the Global Positioning System (GPS), with very accurate aircraft positioning information, on the relative

	13-state	6-state	2-state
Schuler SRS+dv+DA	394	450	
No Schuler SRS+DV+DA+GPS		450	200

Table 1. Comparison of RMS results in yards.

position errors of the sonobuoy. In the 13-state system Orincon observed that the errors in the states pertaining to the aircraft inertials improved markedly; relative position errors most likely remained the same. This is exactly what was found in the analysis of the six-state system: that the relative position errors were unaffected by navigational errors. However, the two-state system is very much affected by navigational errors. Relative position errors were significantly less for the two-state system when accurate aircraft positions were known.

To obtain a more realistic analysis of their performance, each algorithm was tested using the circular and square flight patterns of Figures 6 and 7. The results of the first simulation run for the circular path at a range of 15 NM are shown in Figures 19 thru 21. In this case, the wavy nature of the covariance is due to the location of the aircraft in the pattern since the range to the sonobuoy remains relatively constant. For instance, as the aircraft passes directly north or south of the sonobuoy the ability to correct E-W errors is greatest. Therefore, the E-W variance reaches a minimum value at this time. The N-S variance is affected in a similar manner. The two-state filter is not as sensitive to this as the six-state filter. Also, the effect of the Schuler cycle on the two-state filter can be seen in Figure 20.

The RMS errors are shown in Figures 22 thru 24. The steady state RMS value of error predicted by the filter is

425 yards for the two-state and 750 yards for the six-state system. The actual RMS error in the relative position of the sonobuoy is 250 to 1200 yards for the two-state and 400 yards for the six-state system. Without Schuler cycle error the two-state system drops to a steady 250 yards, better than the six-state system. It is interesting to note that the measured errors are significantly less than those predicted by the filter. Both measured values of RMS error, σ_x and σ_a , are in close agreement; that is, the deviation of the filter's estimated position about its mean is generally the same as the deviation of the estimated position about the actual relative position of the sonobuoy. The closer these values are to one another, the more confidence can be placed in the analysis. The only exception to this is Figure 23 which shows the reaction of the two-state filter to the Schuler cycle. One complete cycle with a period of 84 minutes is obvious. The Kalman filter does not recognize this cycle since the modeling equations do not account for it. The dip in the center of the two peaks is once again the result of flight path geometry. The aircraft is in such a position relative to the sonobuoy that the measurements provide enough information to correct for Schuler cycle errors. However, it is not in this position long enough to influence the errors anymore than it does.

The same analysis was performed using the square pattern of Figure 7. This scenario has more drift than the other and in a different direction providing the algorithms

with another motion to track. The outcome was generally the same as can be seen in Figures 25 thru 30. There is no measurable difference in the results from the two different flight paths. Flying a straight path does not adversely affect the results as was concluded by the Oricon Corporation for their 13-state system. The wavy nature of the covariance is again a result of the flight path. It is not as smooth as before because a square pattern was flown instead of a circular one. Another look at errors caused in the two-state system by the Schuler cycle can be seen in Figure 30. The algorithm does make some corrections to these errors but they are not as effective as before. In this case the time the aircraft was in a position to make the corrections did not coincide with the time the peak errors occurred. There appears to be no way to predict the optimum time and place for the aircraft to be without knowing when and how the Schuler cycle is occurring.

The distance the aircraft is from the sonobuoy is directly related to the accuracy the algorithms can achieve. Simulations were performed at ranges of 5, 15, 30, and 45 nautical miles using a circular flight path with the sonobuoy in the center. (The RMS plots can be found in Appendix C.) The steady state errors were observed and are plotted in Figure 31. Solid lines represent the RMS values of the covariance and dashed lines represent the actual errors. In all cases the errors increase with distance which is intuitively satisfying. Mathematically, the

covariance increases because the H matrix becomes smaller with an increase in range. Consequently, the updated value of the covariance, $P(+)$, is larger from Equation (19). For the six-state system there is an increase in actual error from 240 yards at 5 NM to 750 yards at 45 NM. And for the two-state system, actual errors increase from 130 yards at 5 NM to 500 yards at 45 NM. (Note that for the two-state system, Schuler cycle errors, indicated by circles, peak approximately 1000 yards above these plotted values; the lower values, indicated by x's, were used since they were more reliable for comparison between ranges.) There is a slight decrease in the slopes of all the error curves as distance increases. However, this decrease is small and the curves might well have been interpreted as linear within the limits of the analysis.

The frequency at which measurements are made also affects the accuracy of the estimate. Each algorithm was tested with measurement intervals of 4, 10, 20, and 30 seconds on one sonobuoy. The outcomes of the steady state RMS errors are plotted in Figure 32. (Once again the two-state system's errors do not show the effect of the Schuler cycle.) The actual errors in the six-state system increase steadily as the measurement interval increases. However, the covariance decreases rapidly at first to a minimum value somewhere around 10 seconds and then increases with increasing interval. An abbreviated run with a two second measurement interval confirmed these results.

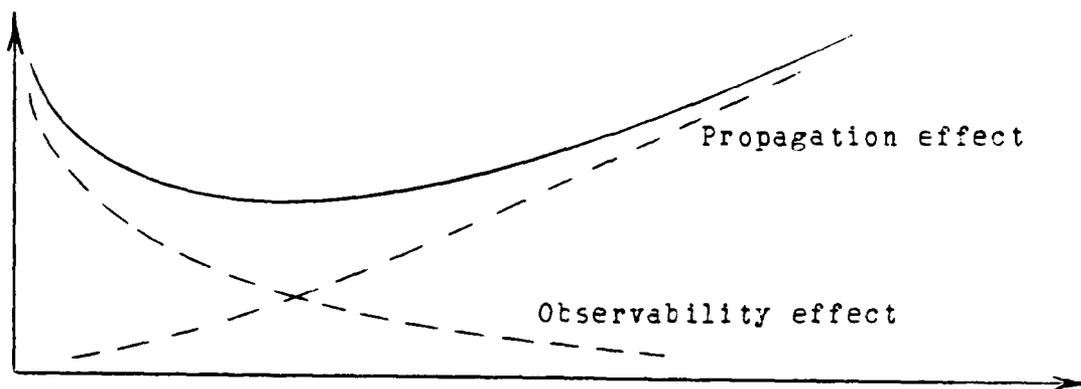


Figure 33. Factors which influence the covariance.

Figure 33 helps to explain this outcome. There are two conditions which affect the covariance. The most obvious is the increase in covariance due to an increase in the propagation interval. This occurs because the plant matrix, which is a function of time, affects the covariance in Equation (16). The second condition is a consequence of the model's observability. If the relative bearing between the aircraft and the sonobuoy does not change then the state is not observable. In other words, two successive bearings must intersect to determine an estimated position. By allowing the interval between measurements to become too small, the aircraft is unable to make a significant change relative to the sonobuoy. The covariance increases as the conditions approach those which make the problem unobservable. There appears to be an optimum frequency with which to make measurements on one buoy, namely 10 seconds for this range of 15 NM and speed of 180 knots.

The two-state system exhibited similar characteristics in Figure 32. In this case the covariance never does increase in the range of intervals studied. The actual errors remain constant. Since the plant matrix for this system is equal to the identity matrix and the Q matrix is prescribed to be constant, the propagation interval, Δt , does not effect the covariance. However, the effects of the observability conditions do cause the covariance to increase when Δt becomes too small. It is believed that as Δt becomes smaller the actual errors would begin to increase also.

The statistical plots for the range tests found in Appendix B show that the Schuler cycle causes larger peaks in the actual error as the range from the sonobuoy increases. However, the peaks are decreasing with time. Further simulation revealed that the peak errors decrease to a steady value about 1000 yards greater than non-Schuler cycle errors. This coincides with the amplitude of the Schuler cycle as it was programmed for this particular flight path. Therefore, in the steady state a Schuler cycle may cause additional estimation errors approximately equal to its amplitude.

Altitude was also tested for any effect it might have on these algorithms. Using the circular pattern at a range of 15 NM, altitudes of 300, 3000, 10,000, and 20,000 feet were tested. Neither the six-state nor the two-state system showed any effect for these changes in altitude. It is

believed that at very small ranges (less than two nautical miles) altitude might hamper good position estimation. However, for the majority of the time altitude is of no concern.

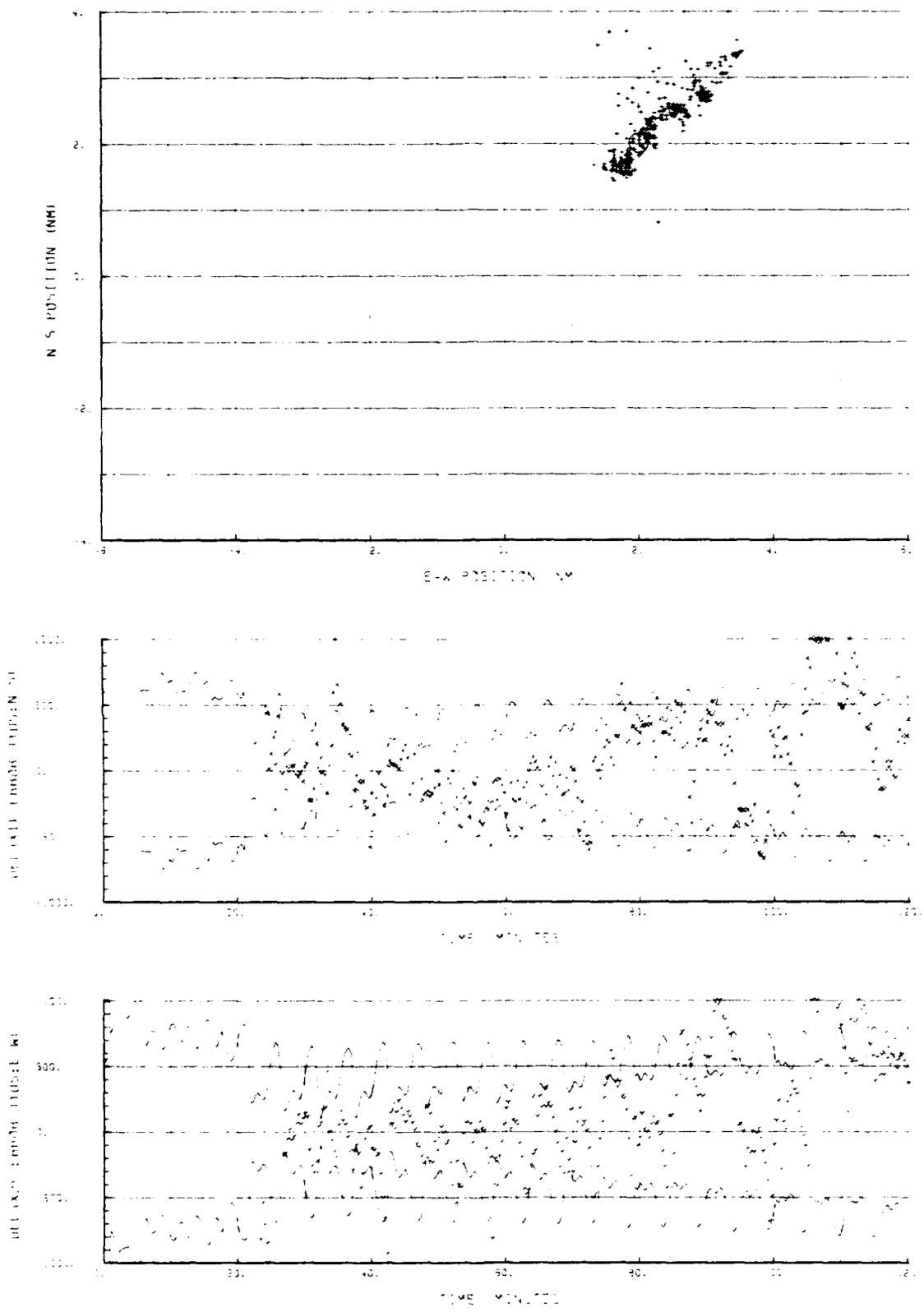


Figure 13. Est. position and relative errors for the six-state system using Orincon's pattern. DR=(a/c 225-5.2 + Schuler, b 045-5.5), $\Delta t=20s$, Alt=3000'.

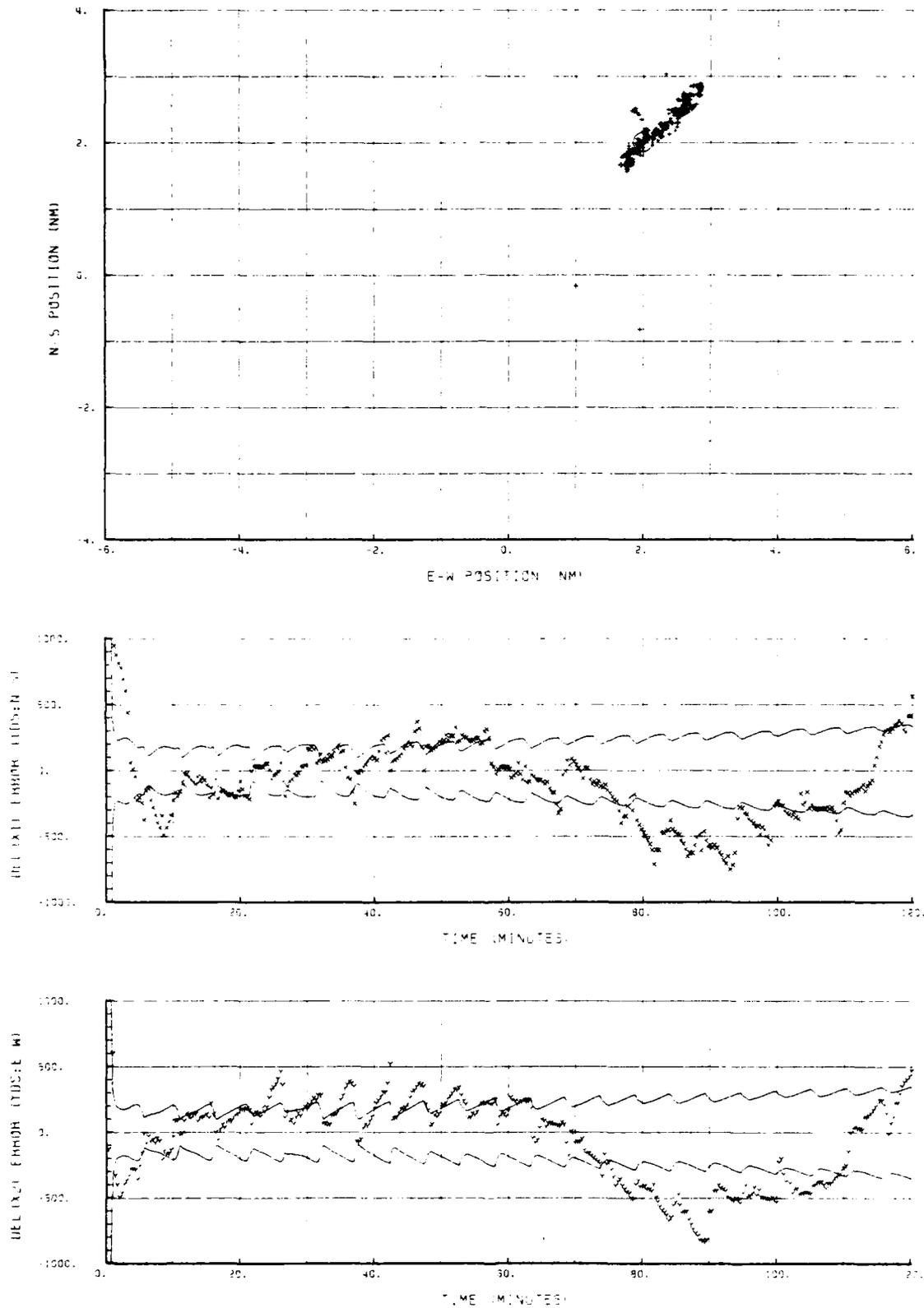


Figure 14. Est. position and relative errors for the two-state system using Orincon's pattern. $DR=(a/c\ 225-5.2 + \text{Schuler}, b\ 045-5.5)$, $\Delta t=20s$, $Alt=3000'$.

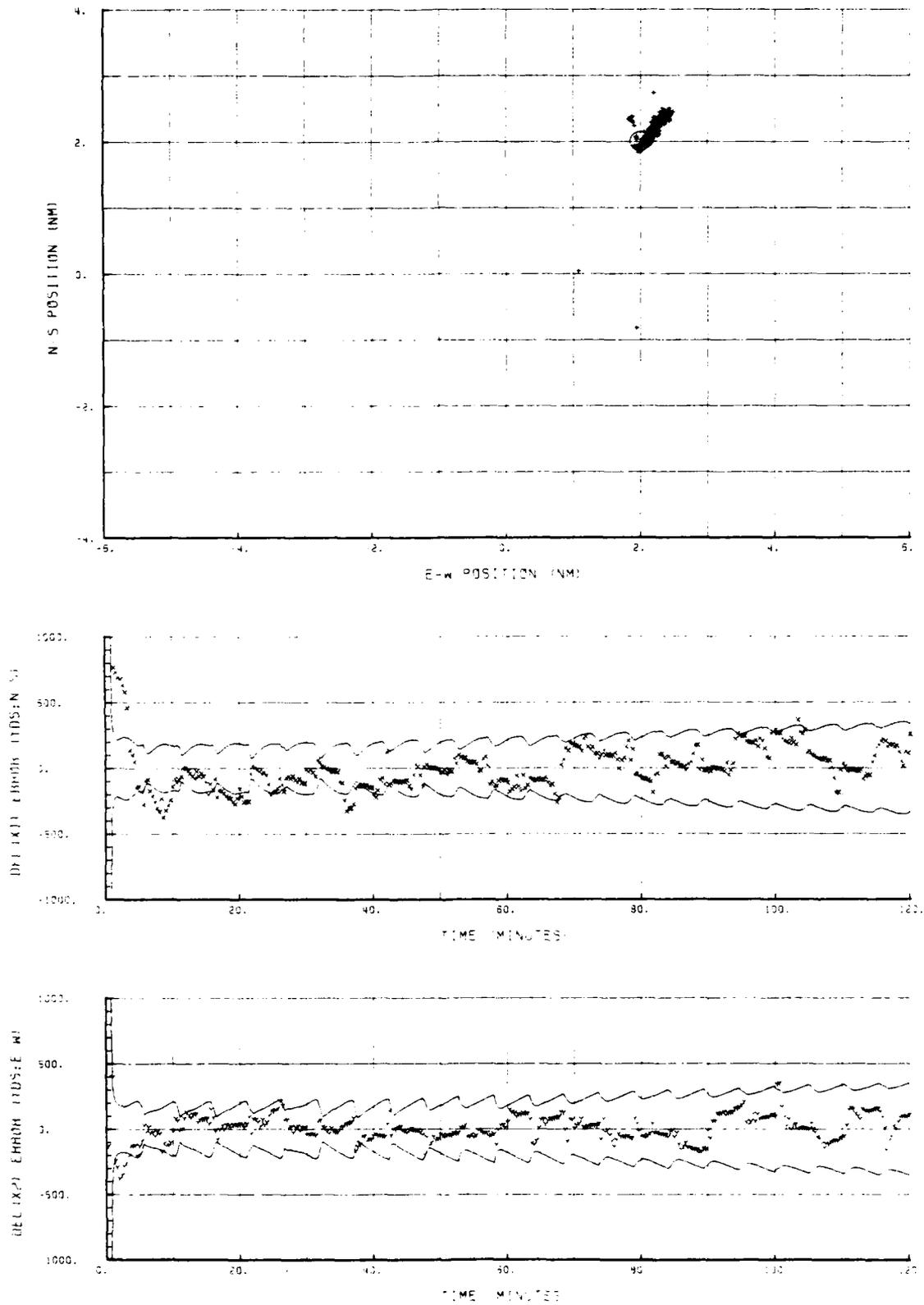


Figure 15. Est. position and relative errors for the two-state system using Orincon's pattern. DR=(a/c 225-5.2 + No Sch., b 045-5.5), $\Delta t=20s$, Alt=3000'.

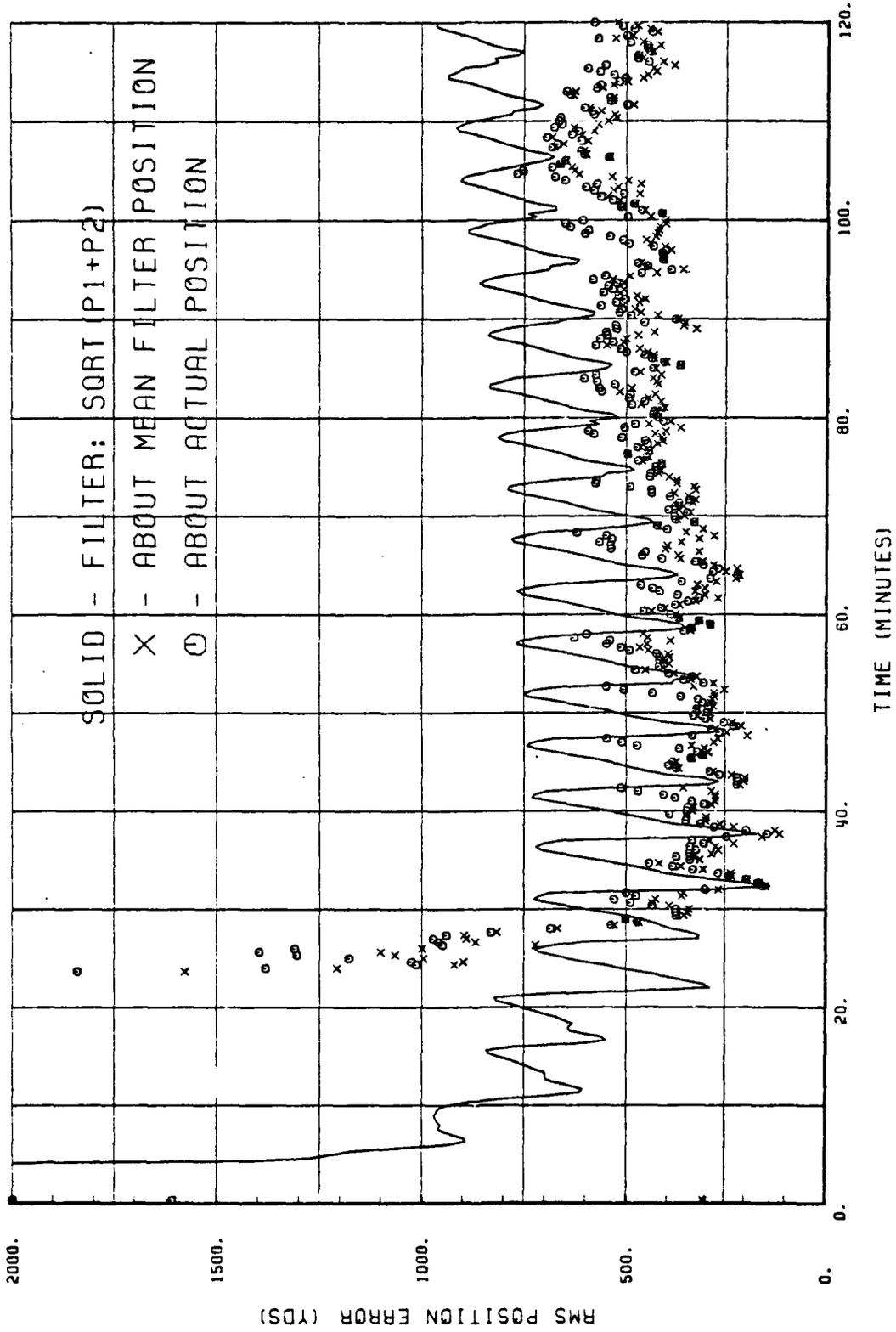


Figure 16. RMS position errors for the six-state system while flying the circular pattern developed by Orincon Corp. DR=(a/c 225-5.2 + Schuler, b 045-5.5), At=20s, Alt=3000'.

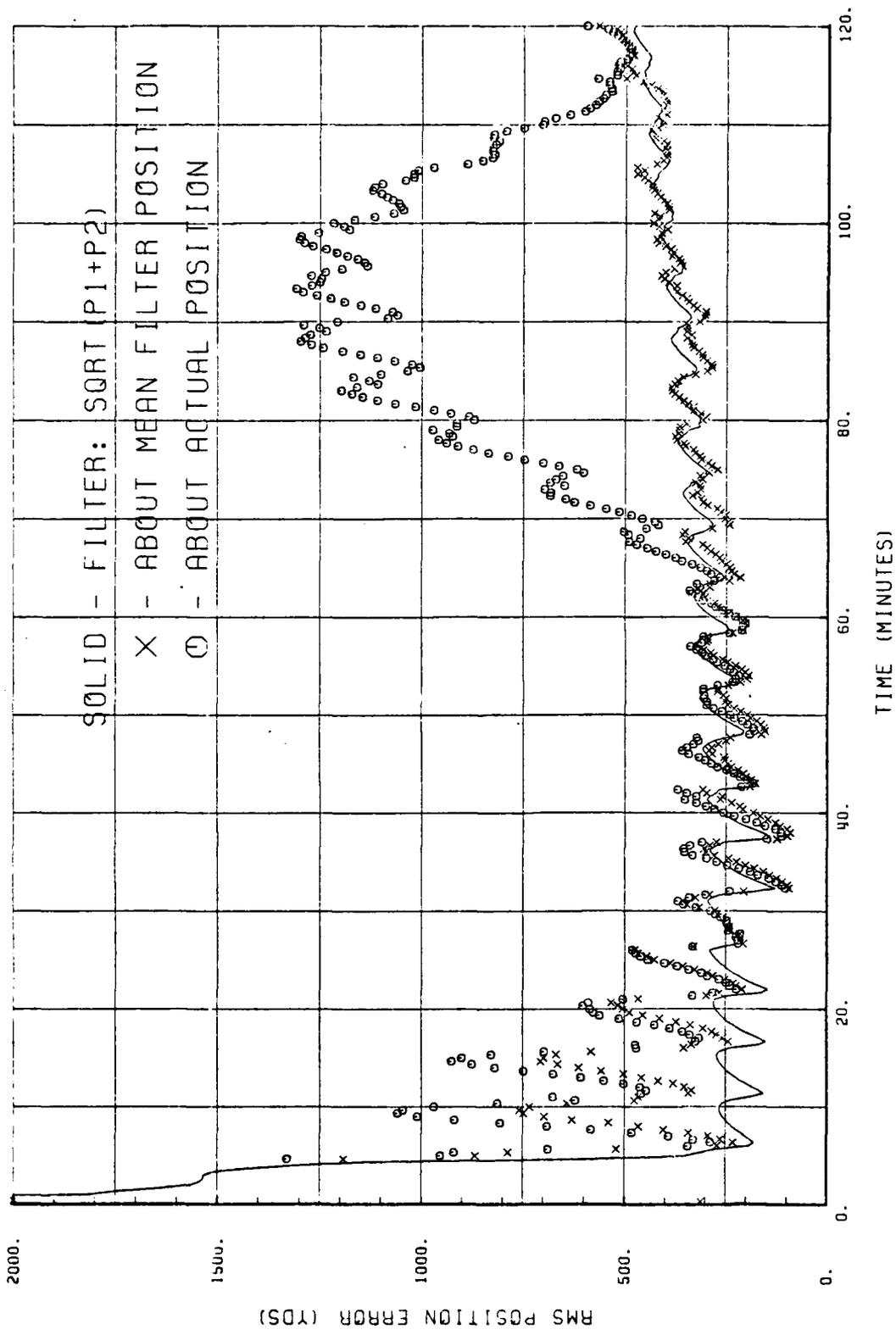


Figure 17. RMS position errors for the two-state system while flying the circular pattern developed by Orincon Corp. DR=(a/c 225-5.2 + Schuler, b 045-5.5), At=20s, Alt=3000'.

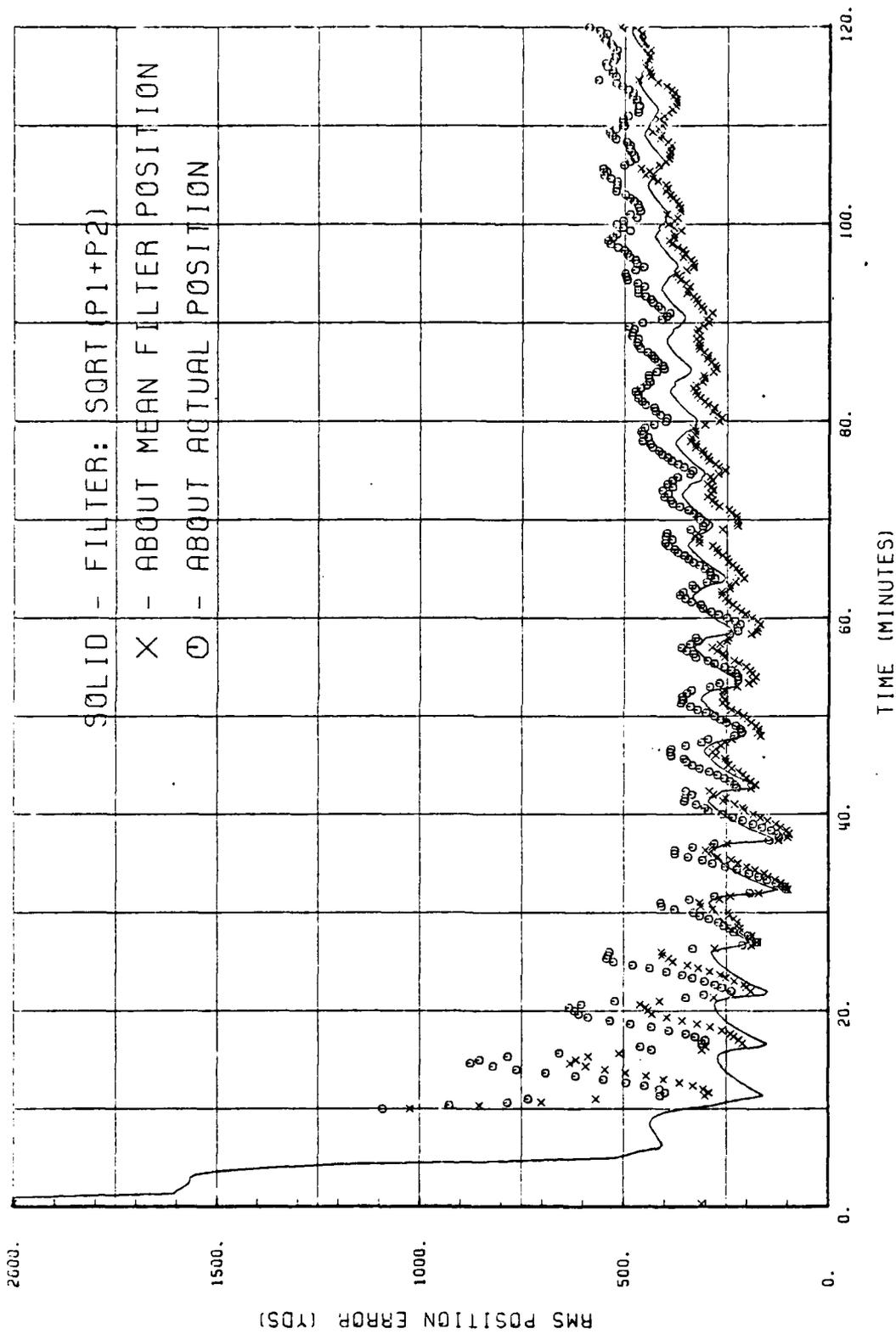


Figure 18. RMS position errors for thr two-state system while flying the circular pattern developed by Orincon Corp. DR=(a/c 225-5.2 No Sch., b 045-5.5), At=20s, Alt=3000'.

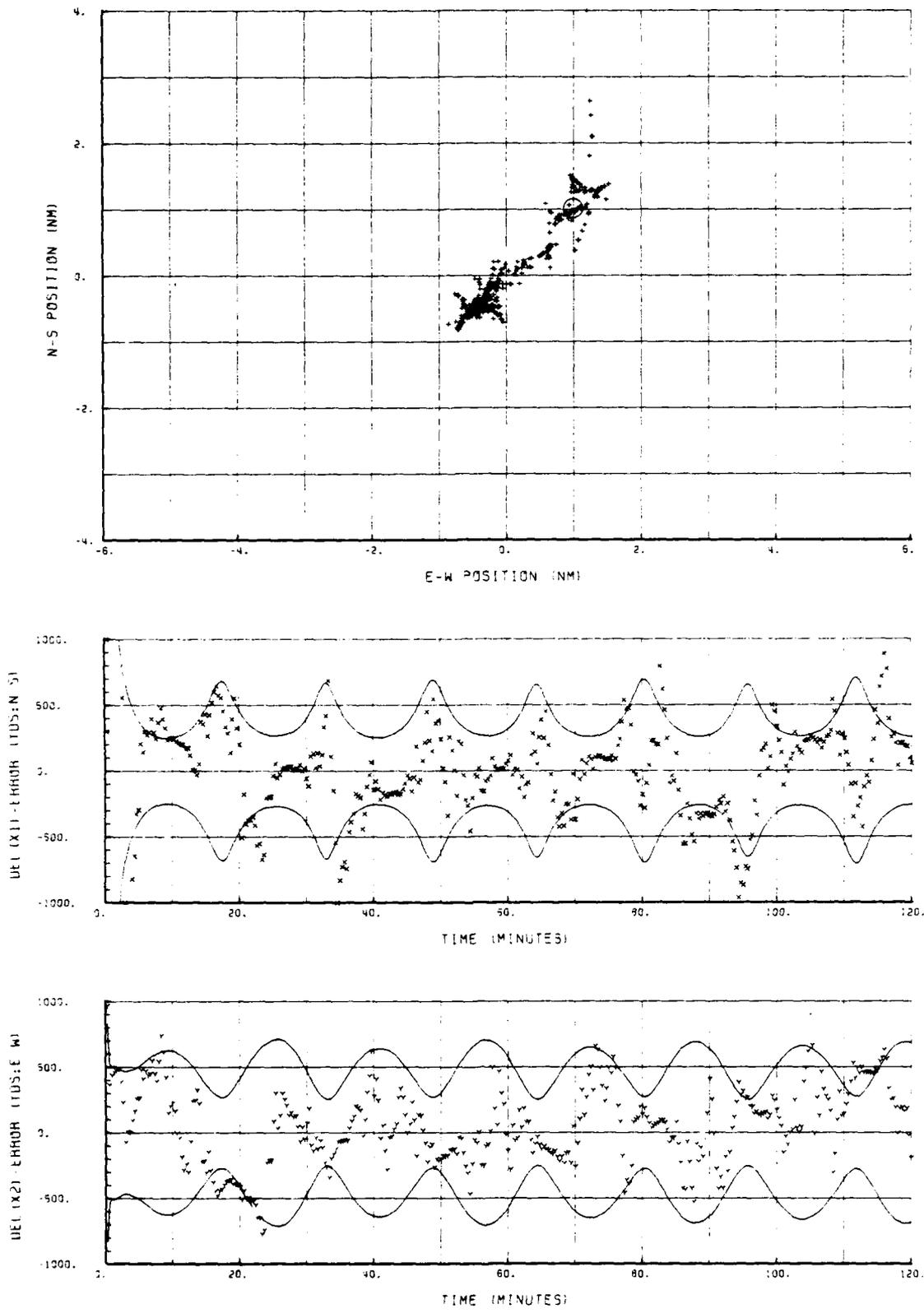


Figure 19. Est. position and relative errors for the six-state system using circular pattern at 15NM. $DR=(a/c\ 180-1 + \text{Schuler}, b\ 270-1)$, $\Delta t=20s$, $Alt=3000'$

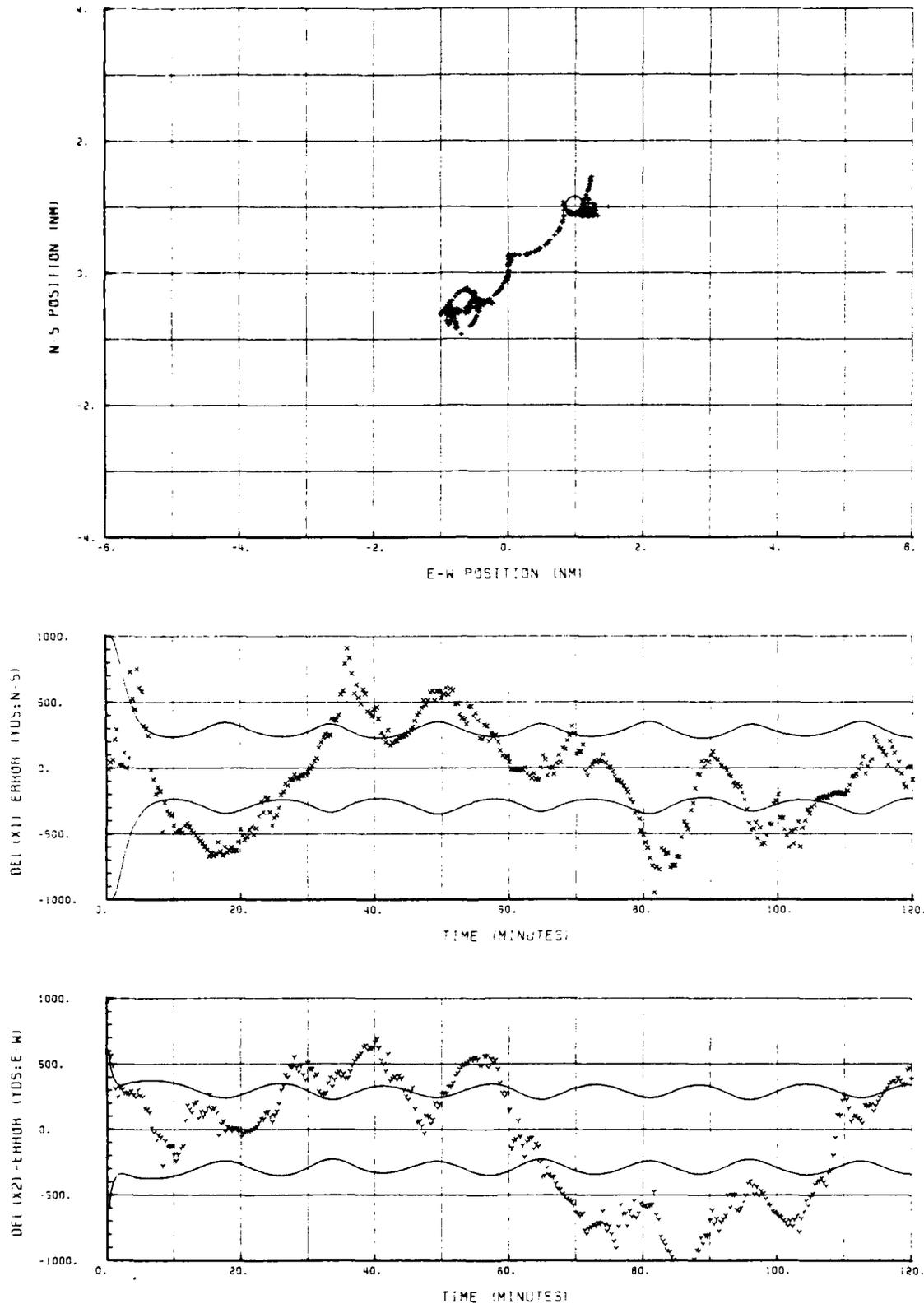


Figure 20. Est. position and relative errors for the two-state system using circular pattern at 15NM. DR=(a/c 180-1 + Schuler, b 270-1), $\Delta t=20s$, Alt=3000'

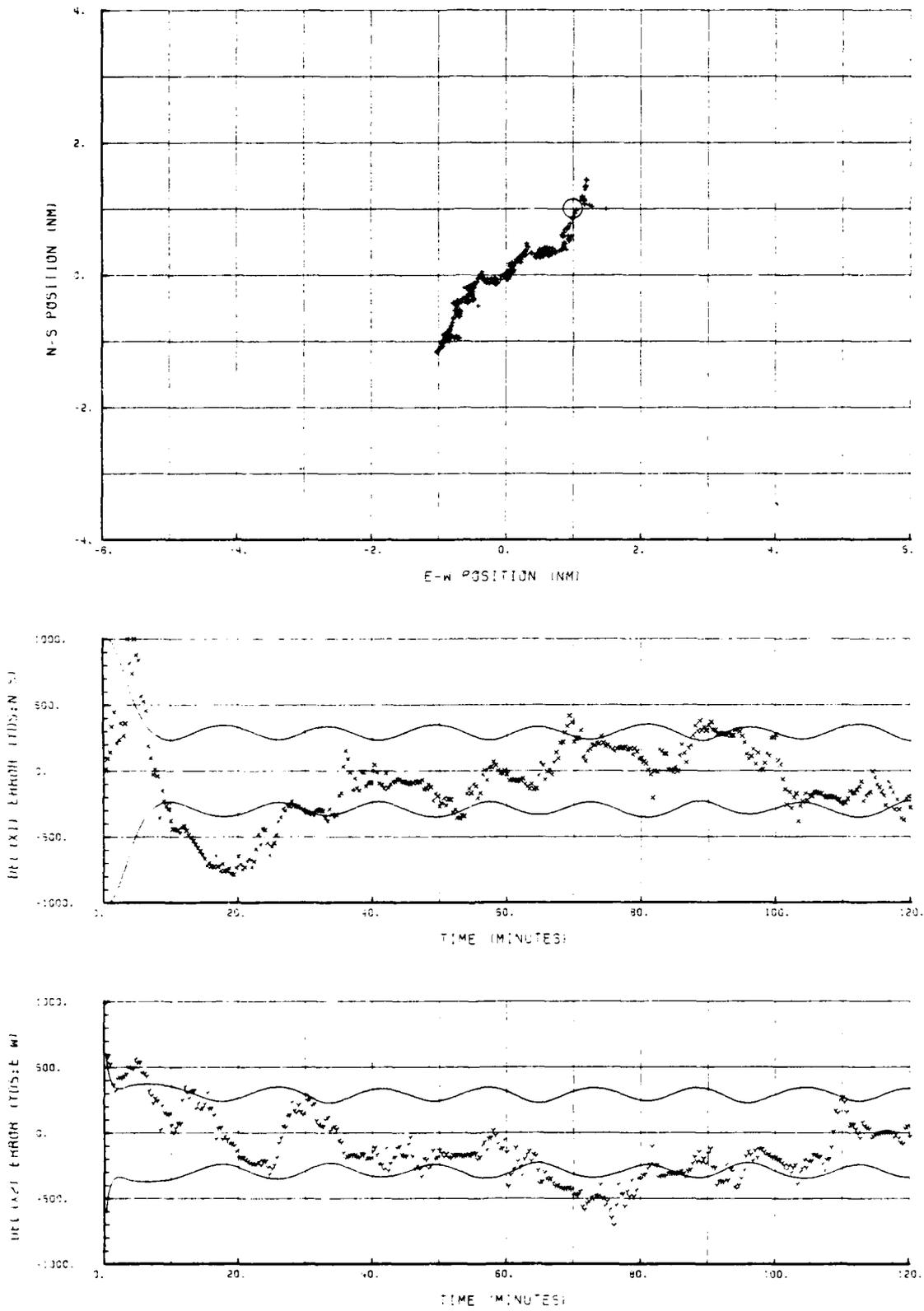


Figure 21. Est. position and relative errors for the two-state system using circular pattern at 15NM. DR=(a/c 180-1 + No Sch., b 270-1), $\Delta t=20s$, Alt=3000'

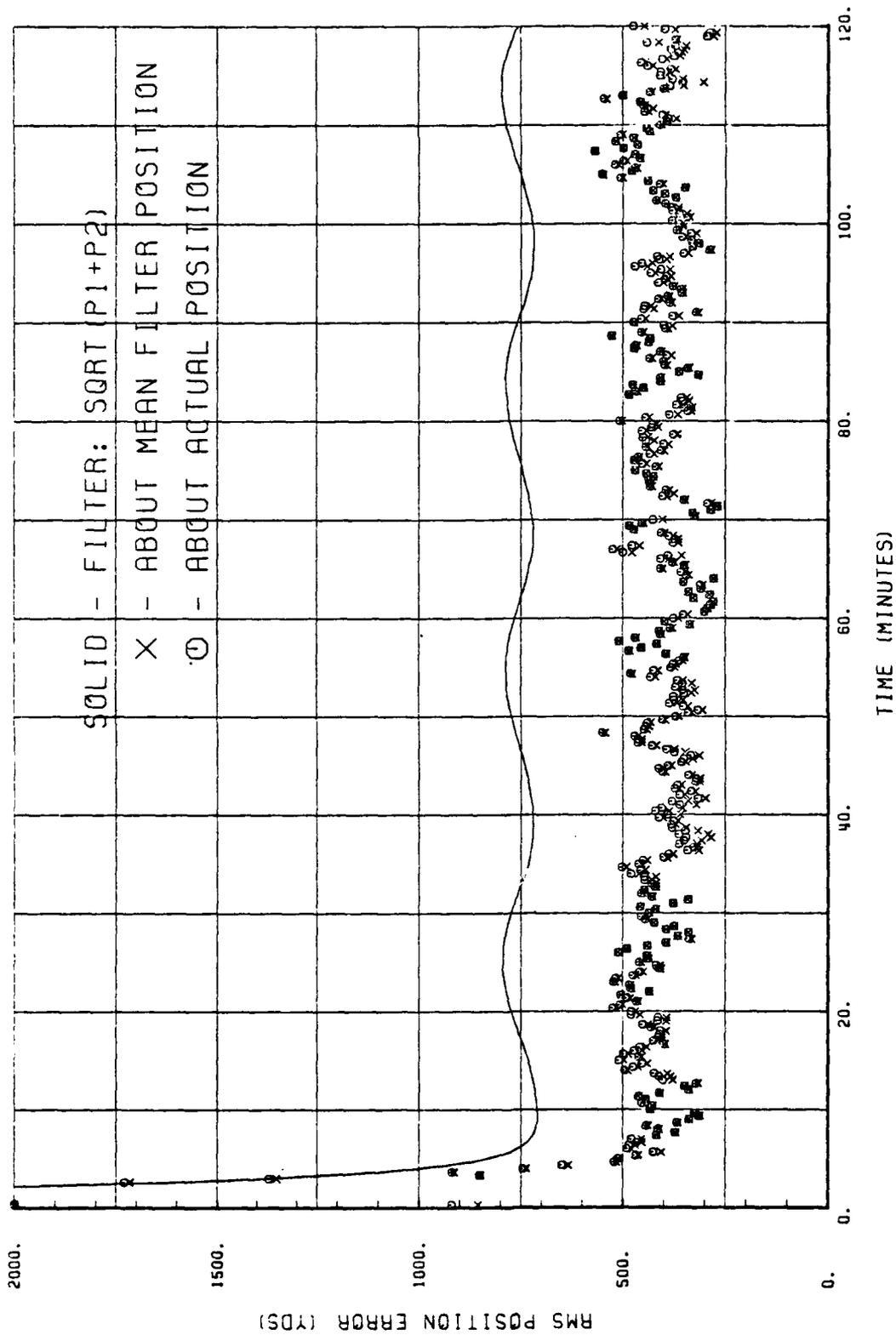


Figure 22. RMS position errors for the six-state system while flying a 15 NM circular pattern around the sonobuoy. DR=a/c 180-1 + Schuler, b 270-1), $\Delta t=20s$, Alt=3000'.

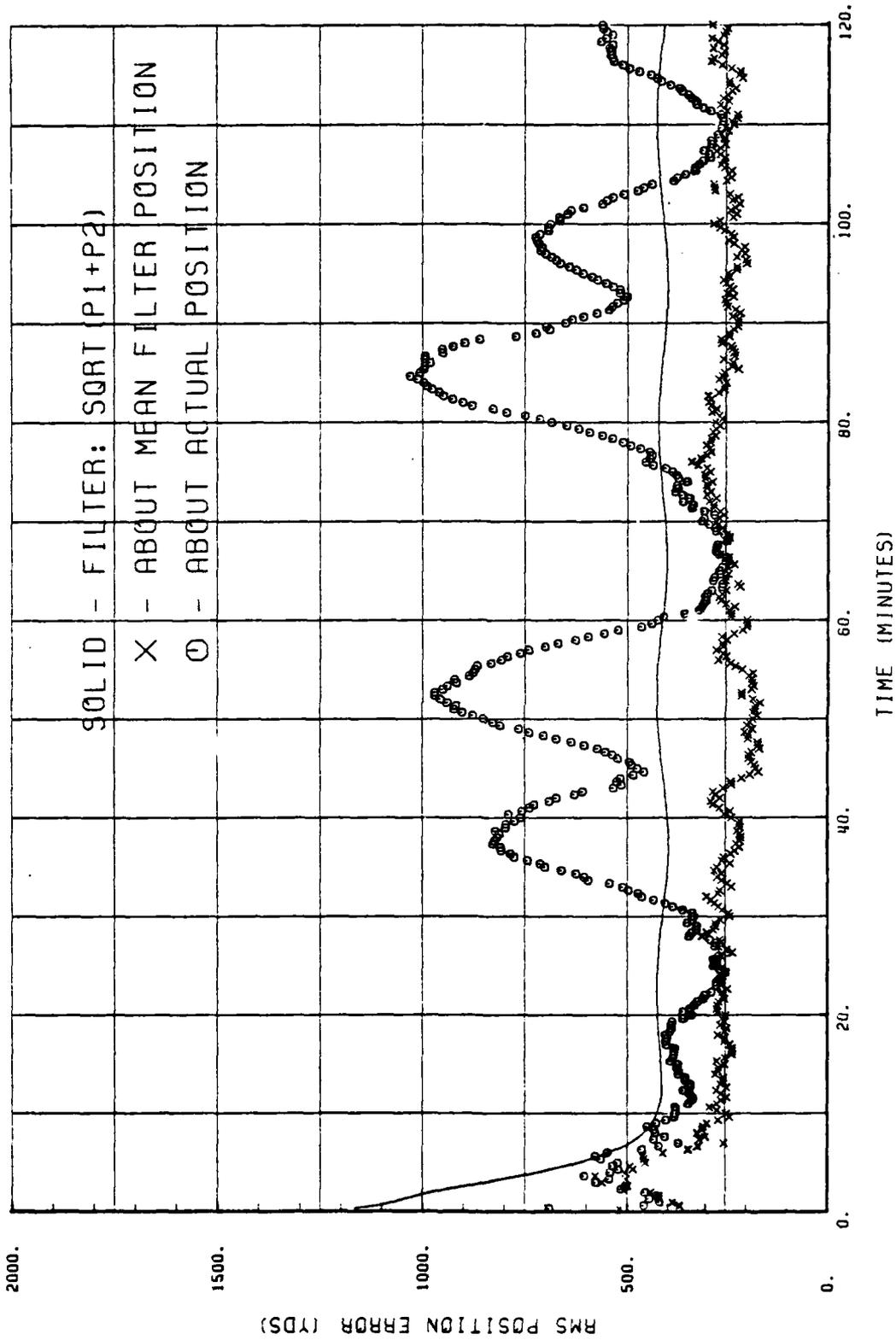


Figure 23. RMS position errors for the two-state system while flying a 15 NM circular pattern around the sonobuoy. DR=(a/c 180-1 + Schuler, b 270-1), $\Delta t=20s$, Alt=3000'.

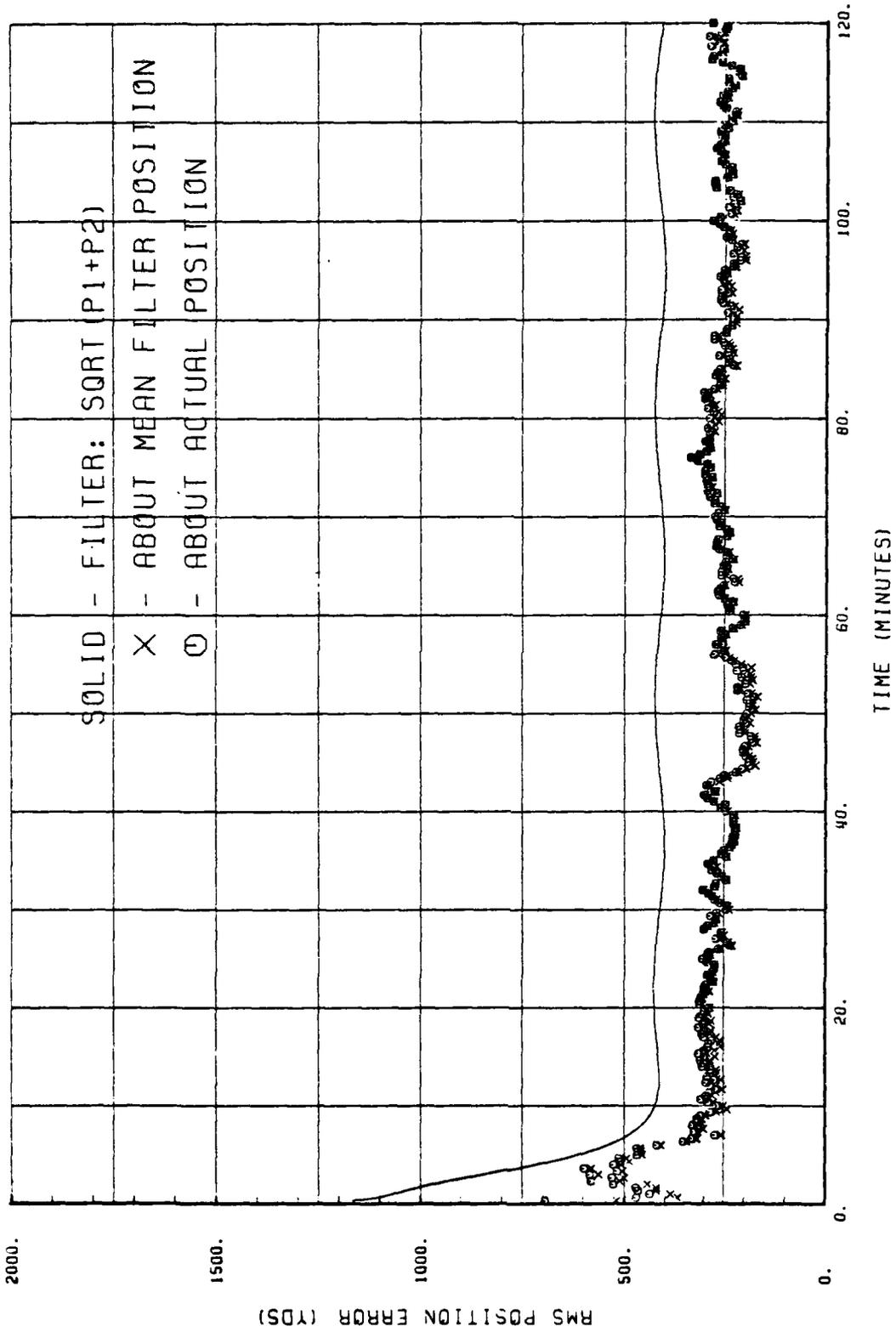


Figure 24. RMS position errors for the two-state system while flying a 15 NM circular pattern around the sonobuoy. DR=(a/c 180-1 + No Sch., b 270-1), t=20s, Alt=3000'.

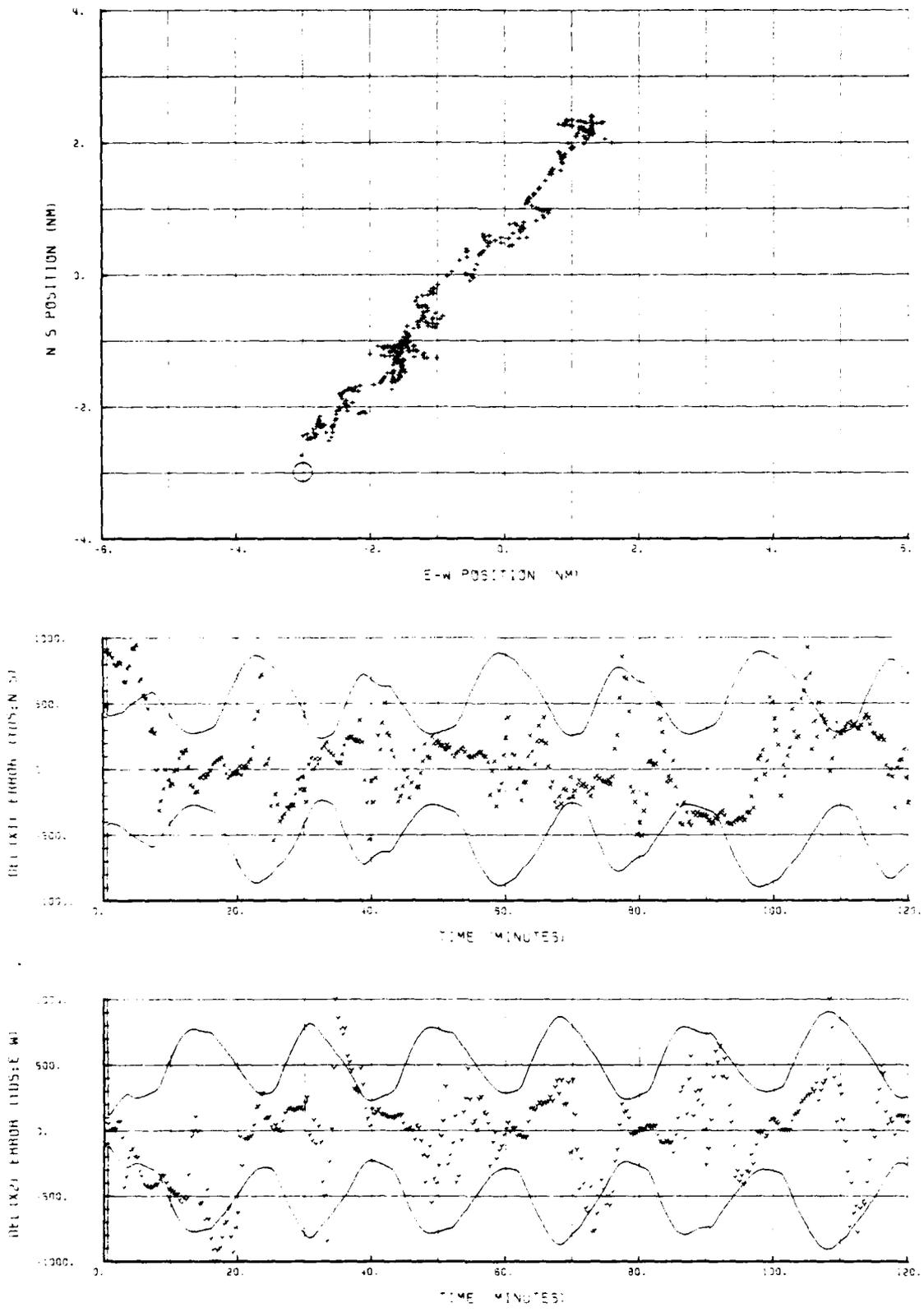


Figure 25. Est. position and relative errors for the six-state system using square pattern at 15NM. DR=(a/c 000-2.5 + Schuler, b 090-2), $\Delta t=20s$, Alt=3000'

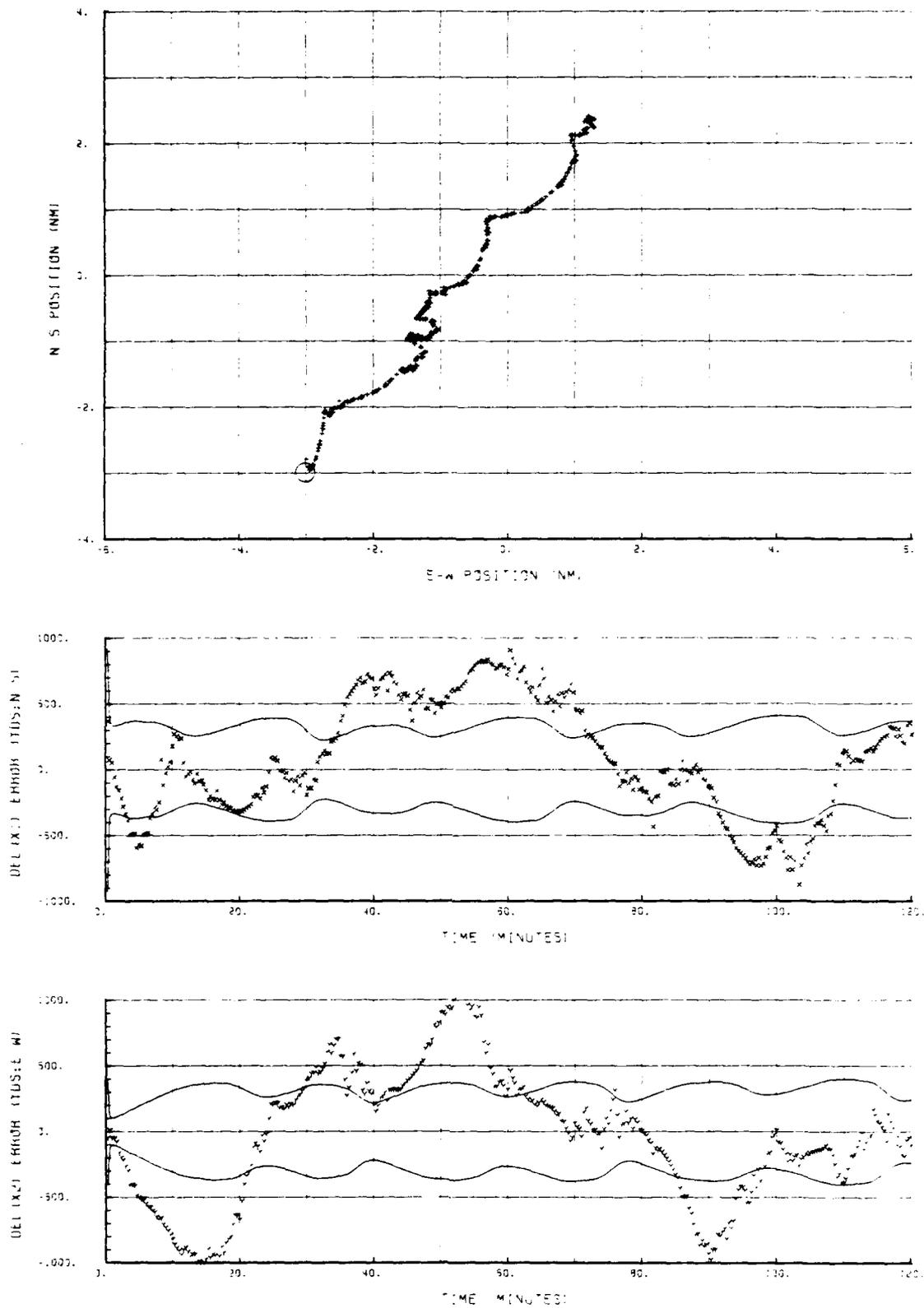


Figure 26. Est. position and relative errors for the two-state system using square pattern at 15NM. DR=(a/c 000-2.5 + Schuler, b 090-2), $\Delta t=20s$, Alt=3000'

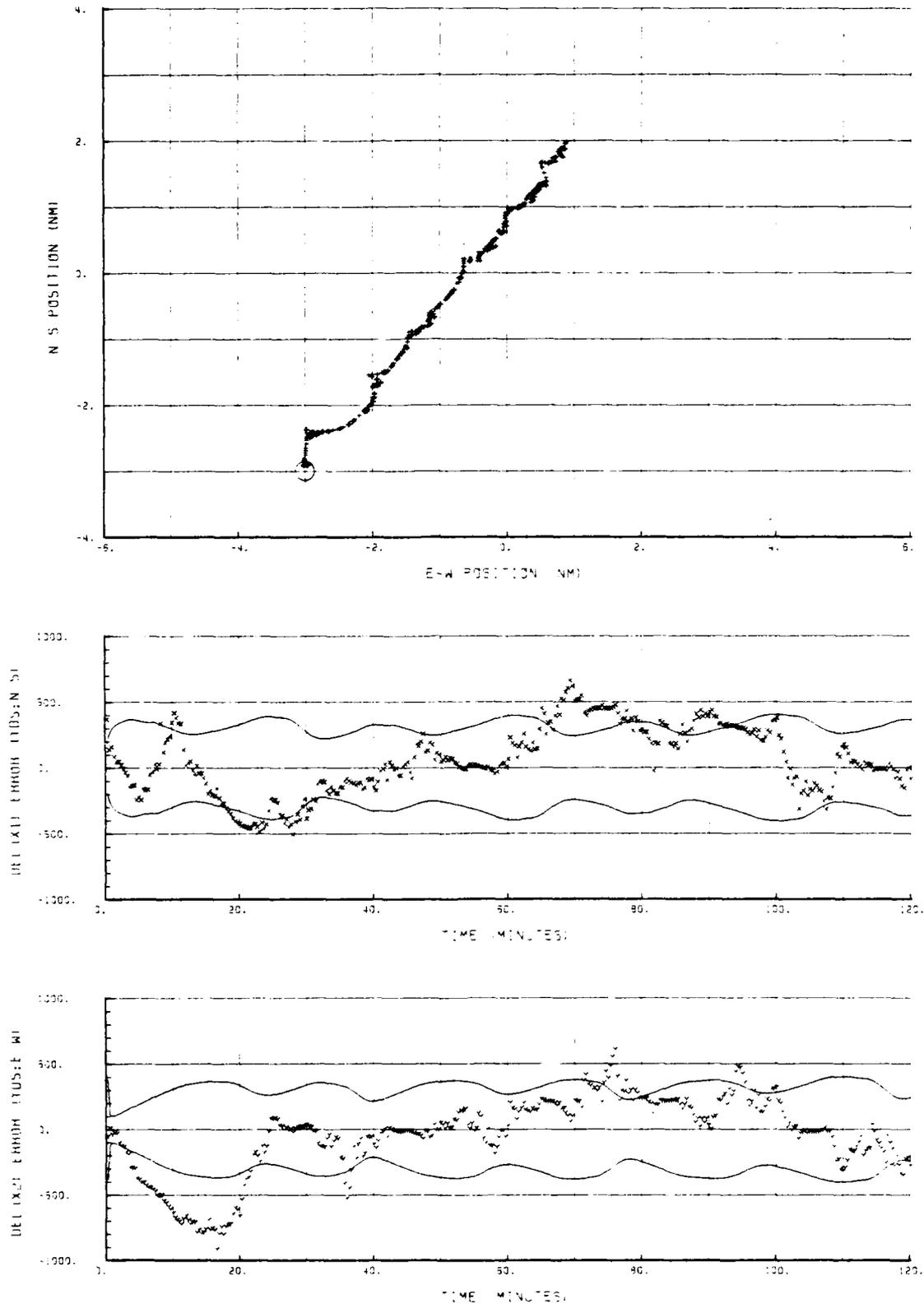


Figure 27. Est. position and relative errors for the two-state system using square pattern at 15NM. DR=(a/c 000-2.5 + No Sch., b 090-2), $\Delta t=20s$, Alt=3000'

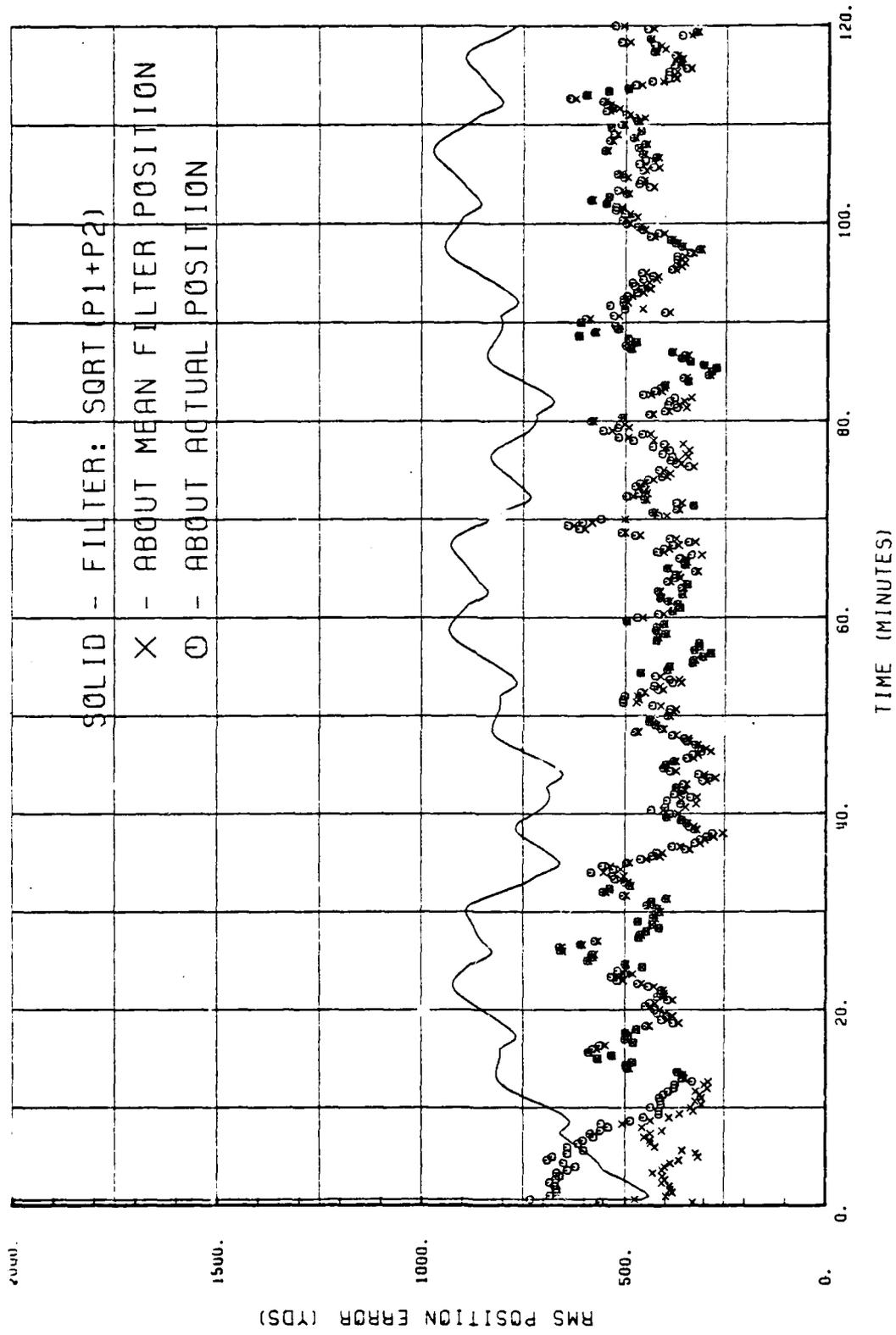


Figure 28. RMS position errors for the six-state system while flying a 15 NM square pattern around the sonobuoy. DR=(a/c 000-2.5 + Schuler, b 090-2), t=20s, Alt=3000'.

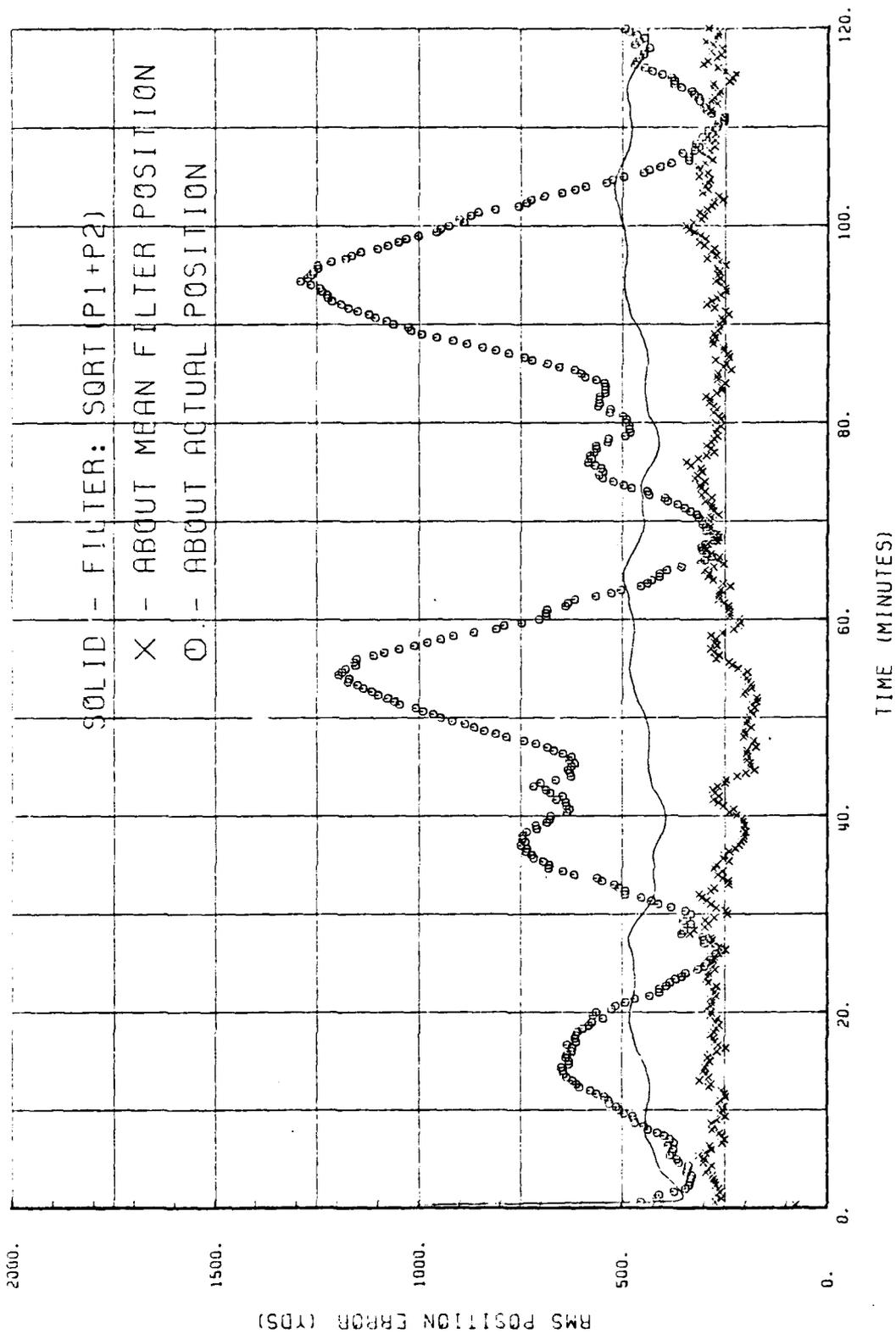


Figure 29. RMS position errors for the two-state system while flying a 15 NM square pattern around the sonobuoy. DR=(a/c 00-2.5 + Schuler, b 090-2), At=20s, Alt=3000'.

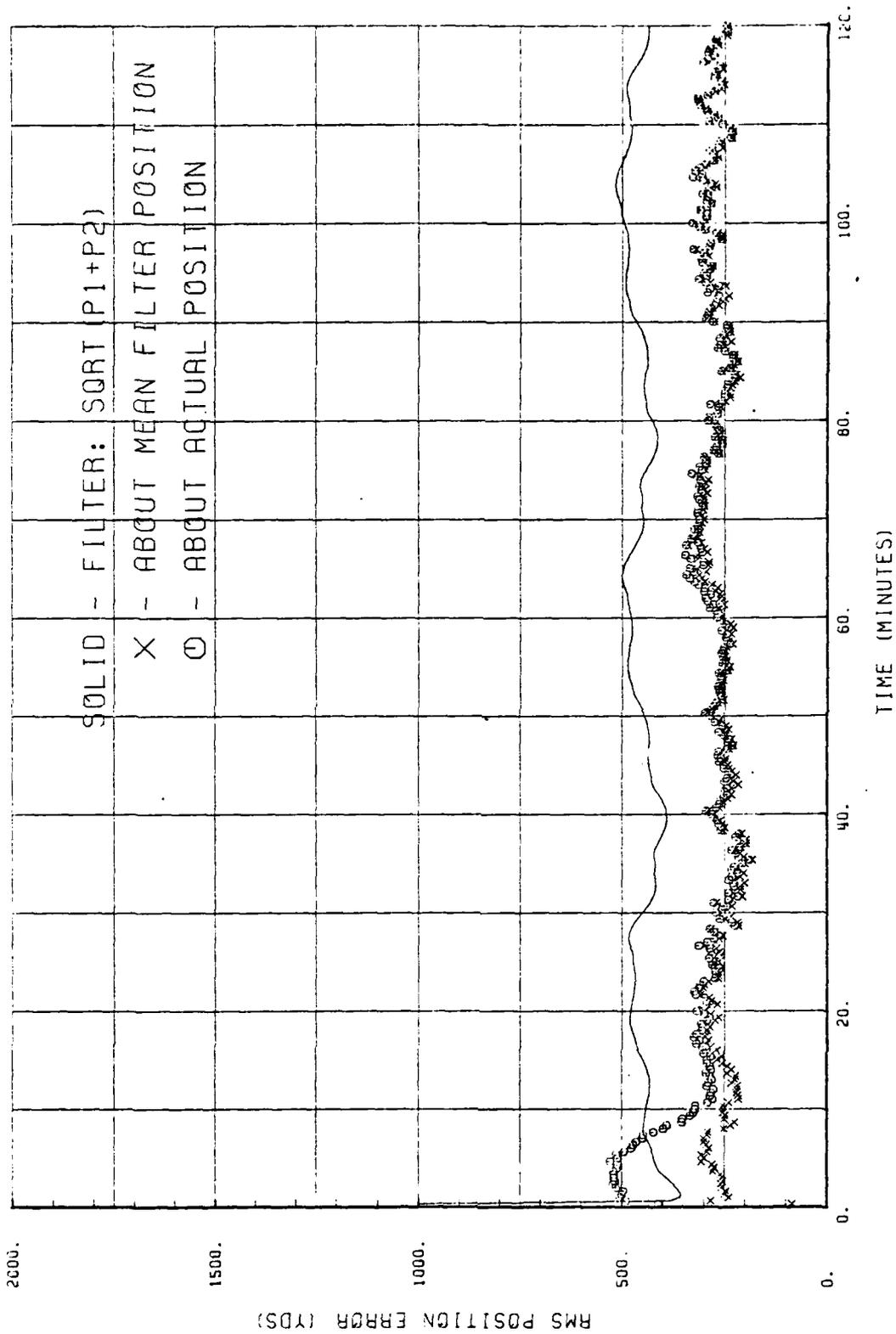
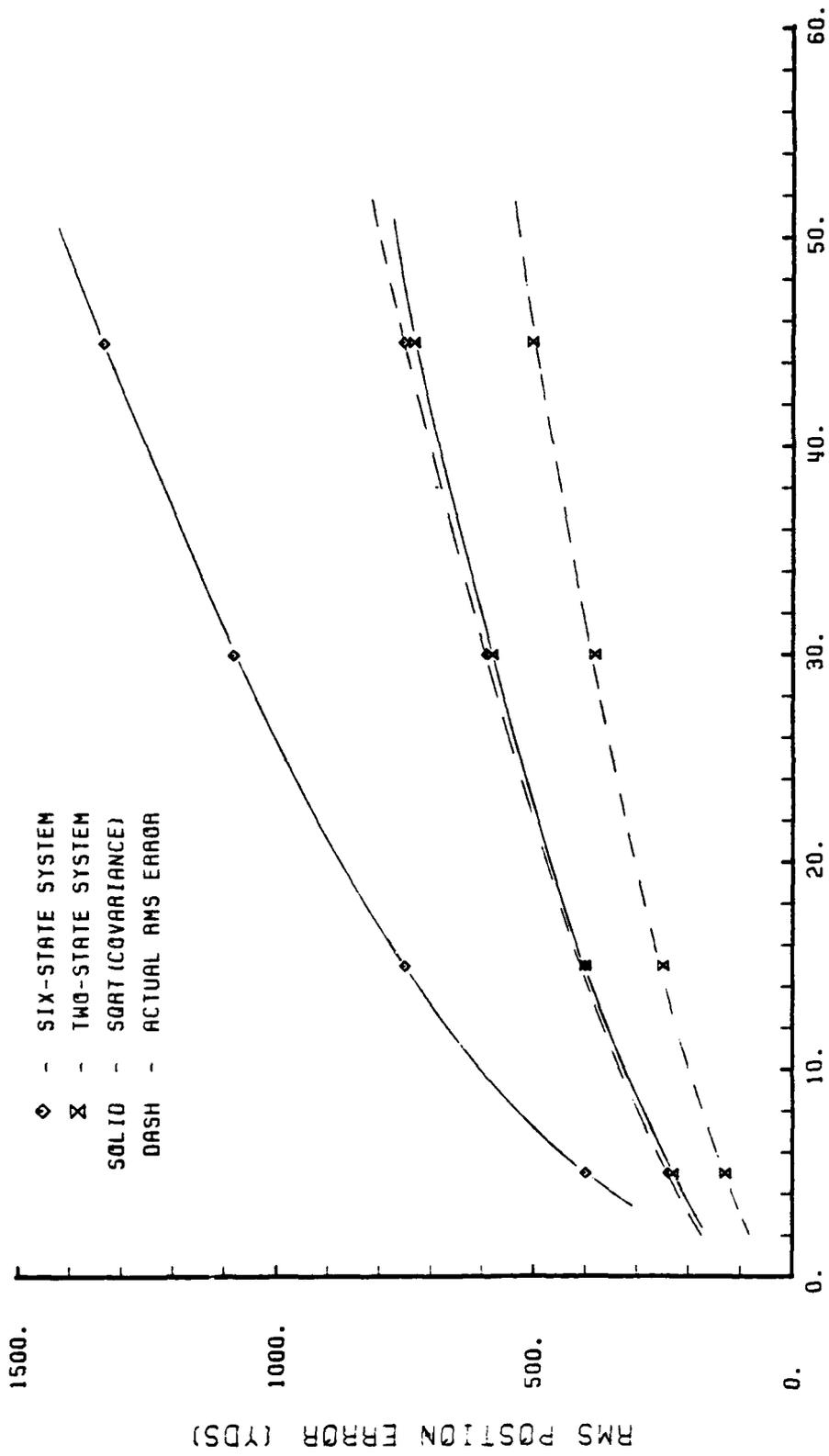
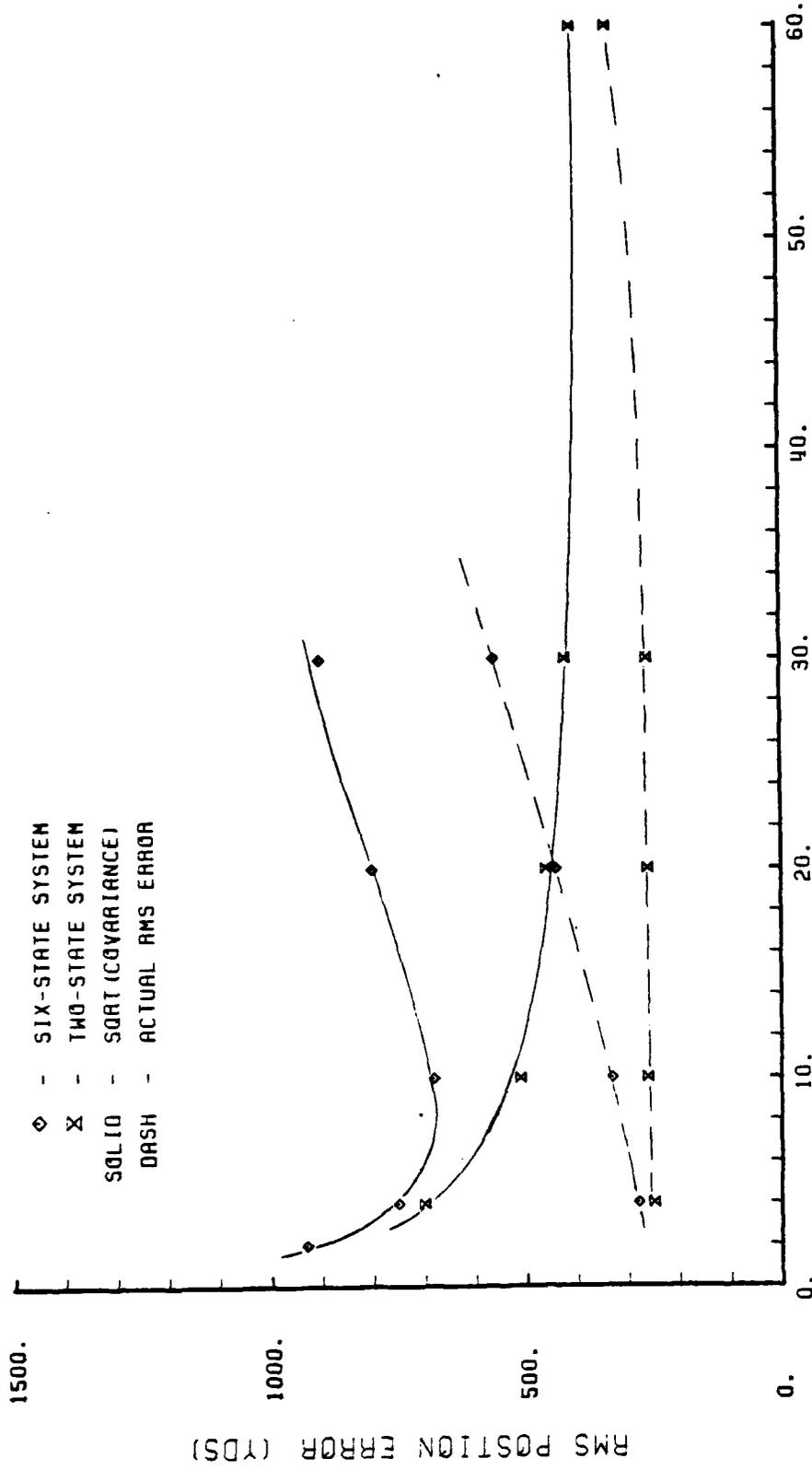


Figure 30. RMS position errors for the two-state system while flying a 15 NM square pattern around the sonobucy. DR(a/c 000-2.5 + No Sch., b 090-2), $\Delta t=20s$, Alt=3000'.



AIRCRAFT-TO-SONOBUOY RANGE (NM)

Figure 31. RMS position error as a function of aircraft-to-sonobuoy range, Δx , for the six-state and two-state systems.



MEASUREMENT INTERVAL - ONE SONOBUOY (SEC)

Figure 32. RMS position error as a function of measurement interval, Δt , for the six-state and two-state systems.

V. CONCLUSION

The following are the conclusions reached at the end of this research:

1. The six-state algorithm developed in this report produced estimates of relative sonobuoy position with a standard deviation in the error of 500 yards at a range of 20 NM. Aircraft drift, which included the Schuler cycle, had no effect on this system since the states were defined relative to the aircraft.
2. The two-state algorithm developed in this report produced estimates of relative sonobuoy position with a standard deviation of 300 yards plus an amount equal to the amplitude of the Schuler cycle error at a range of 20 NM. When no Schuler cycle navigational errors were present, the two-state system out-performed the six-state system.
3. Both algorithms showed an increase in steady state RMS error with an increase in range.
4. The six-state system showed an increase in actual steady state RMS error for an increase in the time interval between measurements; however, the covariance of this system indicated minimum error occurred with a 10 second interval on one sonobuoy at a range of 15 NM and an aircraft speed of 180 kts. (This translates to a $1\frac{1}{2}$ degree change in relative bearing.) The two-state

system showed no change in the actual steady state RMS error for intervals between 4 and 30 seconds and only a slight increase in these errors as a 60 second interval was approached. The covariance of the system actually decreased with increasing interval.

5. Neither system showed any dependence on altitudes within the operational limits at a range of 15 NM.
6. No adverse effect was observed while flying straight tracks as opposed to curved ones.
7. Steady state RMS error was affected by two conditions as measurement intervals changed. The longer the interval the more the error tended to increase, especially for the six-state algorithm, as a result of system propagation. The shorter the interval the more the error tended to increase as the systems encroached upon their non-observability condition. An optimum measurement interval existed for each system which was a function of range.

VI. SUMMARY

Both the six-state and the two-state algorithms were able to successfully track the sonobuoy. However, the six-state system was the best choice given the navigational system in use by the Orion. It was capable of providing estimated sonobuoy positions at a range of 20 NM such that the errors had a standard deviation of 500 yards. The computer time required to process one measurement was on the order of 200 milliseconds. On the other hand, the much simpler two-state system required only 28 milliseconds or one seventh ($\frac{1}{7}$) the amount of computer time. However, its estimate of sonobuoy position was degraded by Schuler cycle errors in the navigational system. Without these, this system was subject to errors having a standard deviation of 300 yards at a range of 20 NM. The Schuler cycle periodically increased these errors by an amount equal to its amplitude.

The algorithms were very much affected by the aircraft's flight path including the relative location of the sonobuoy, the range to the sonobuoy, and the amount of angular change in the bearing to the sonobuoy as measurements were taken. Since the sonobuoy position was to be estimated in two-dimensions, the aircraft was required to maneuver in such a way so as to provide information in both dimensions. For example, as the aircraft crossed north or south of the

sonobuoy the E-W errors decreased while the N-S errors increased. The trend reversed when the aircraft flew east or west of the sonobuoy. It was also observed that steady state RMS errors grew with range at a slightly decreasing rate. They were twice as large at 45 NM as they were at 15 NM. Finally, the sonobuoy position was not observable if the relative bearing to the sonobuoy did not change implying that two lines of bearing must cross in order to obtain an estimated position. If the amount of angular change between each measurement was too small it began to affect the errors because this condition was being approached. Anytime the aircraft flew in such a way as to reduce the rate of bearing change, or the frequency at which measurements were made was too high, this condition was prevalent. It was observed that at least $1\frac{1}{2}$ degrees of change was required to avoid this problem.

The six-state system showed no tendency toward divergence as long as initial conditions were reasonable. The two-state system, on the other hand, did have a tendency to diverge when the gains on the drift filter were not adjusted correctly. For example, if the gain was too high the estimated drift was heavily dependent on the last few measurements and was too quickly affected by changes in the position filter. But, it was also necessary for the drift filter to have non-zero gains in the steady state in order to provide some insurance against an initially wrong estimate of sonobuoy drift. With these modifications made

the algorithm operated correctly during all tests and, in fact, had less error in some cases than did the six-state system.

It is the opinion of the author that either one of these algorithms is an improvement on the historical method of "marking-on-top" the sonobuoy. Further, it adds a great deal more uncertainty as to exactly how much accuracy could be obtained in the old way. Non-linear navigational effects were not even considered and the number of measurements made to determine sonobuoy position was on the order of 10 or less for any one sonobuoy pattern. Nevertheless, in an operational sense the accuracies achieved here would need to be improved. Without the non-linear navigational errors the accuracies could have been much better. However, the results of this research do provide another perspective for the problem at hand.

APPENDIX A
THE COMPUTER PROGRAMS

MAIN

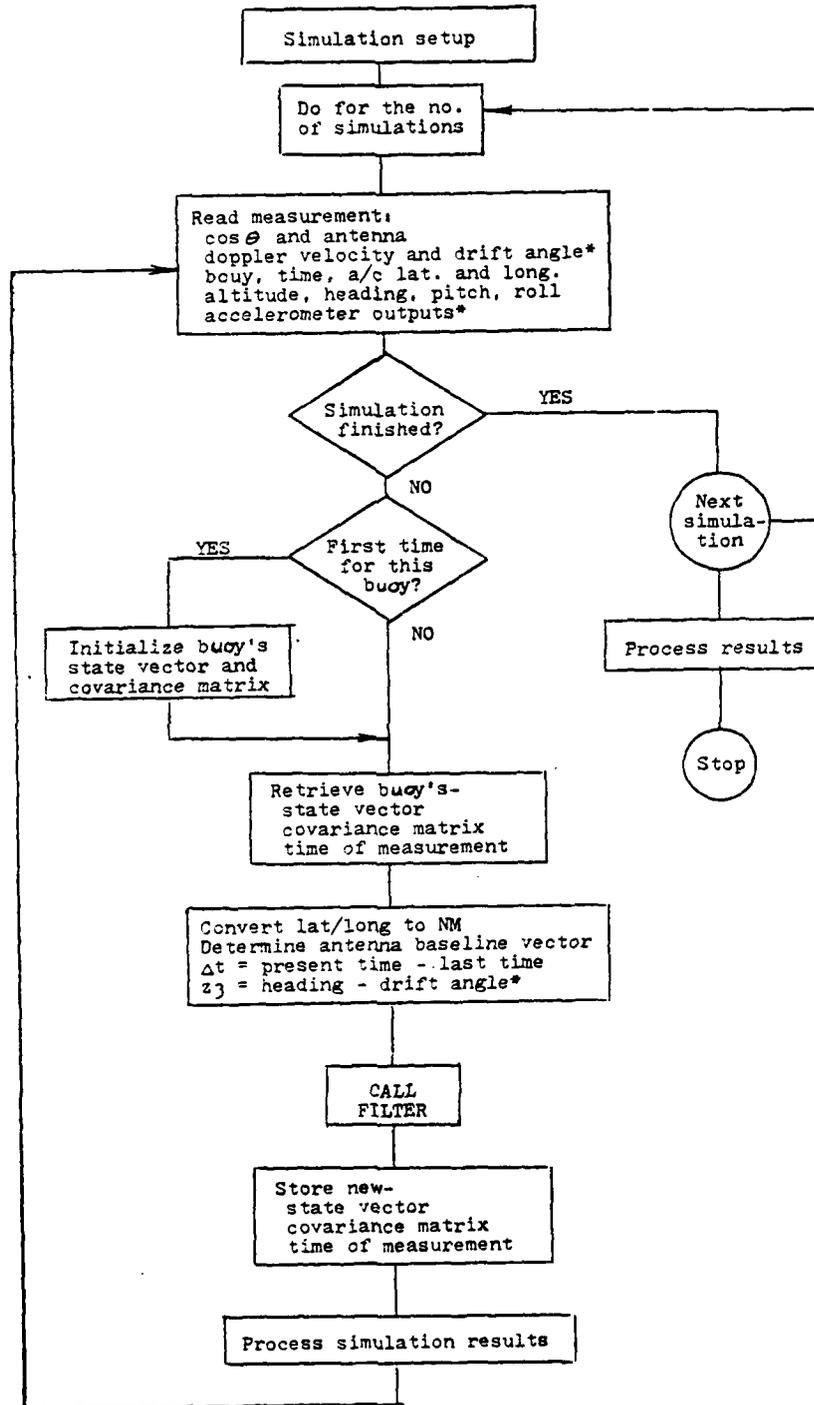


Figure 34. Flow chart:
main program

* Required for six-state system only.

SUBROUTINE FILTER
SIX-STATE

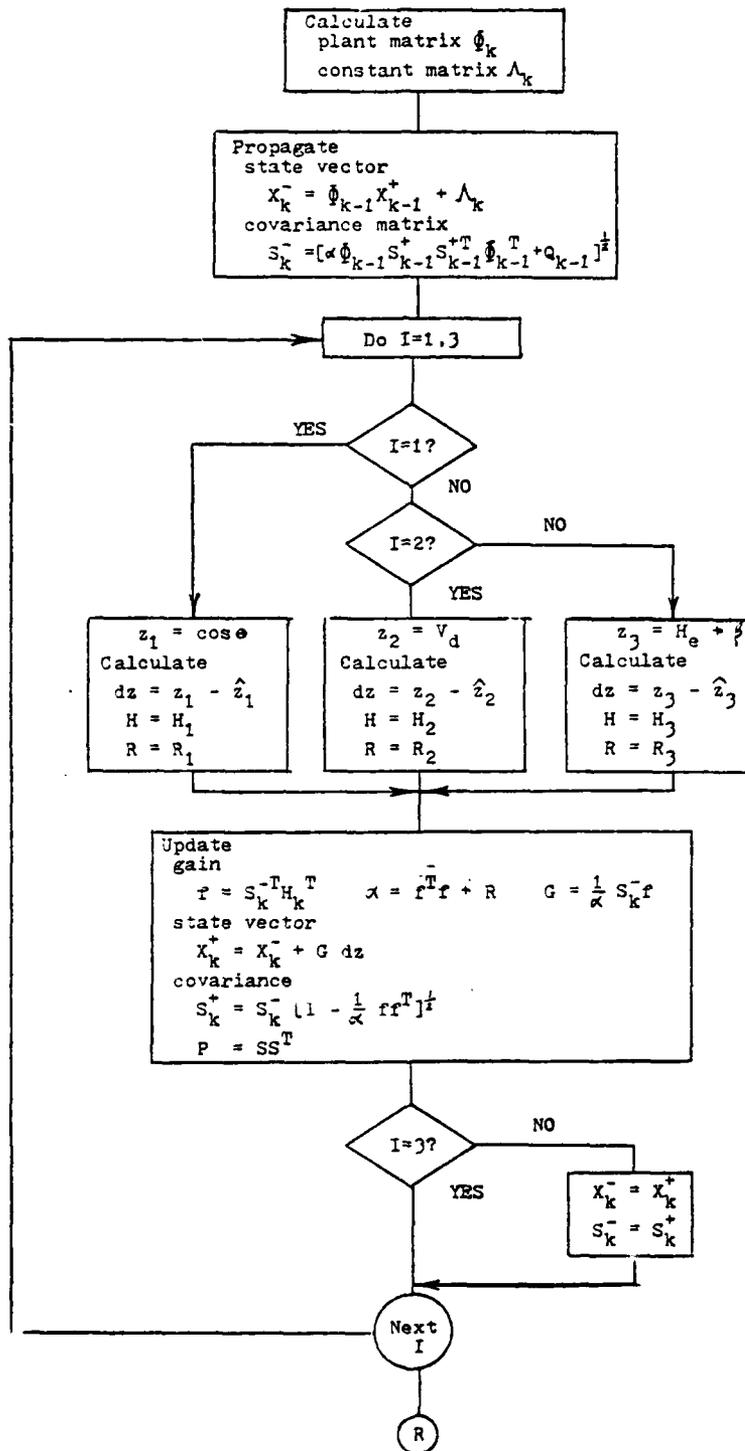


Figure 35. Flow chart: Subroutine FILTER six-state

SUBROUTINE FILTER
TWO-STATE

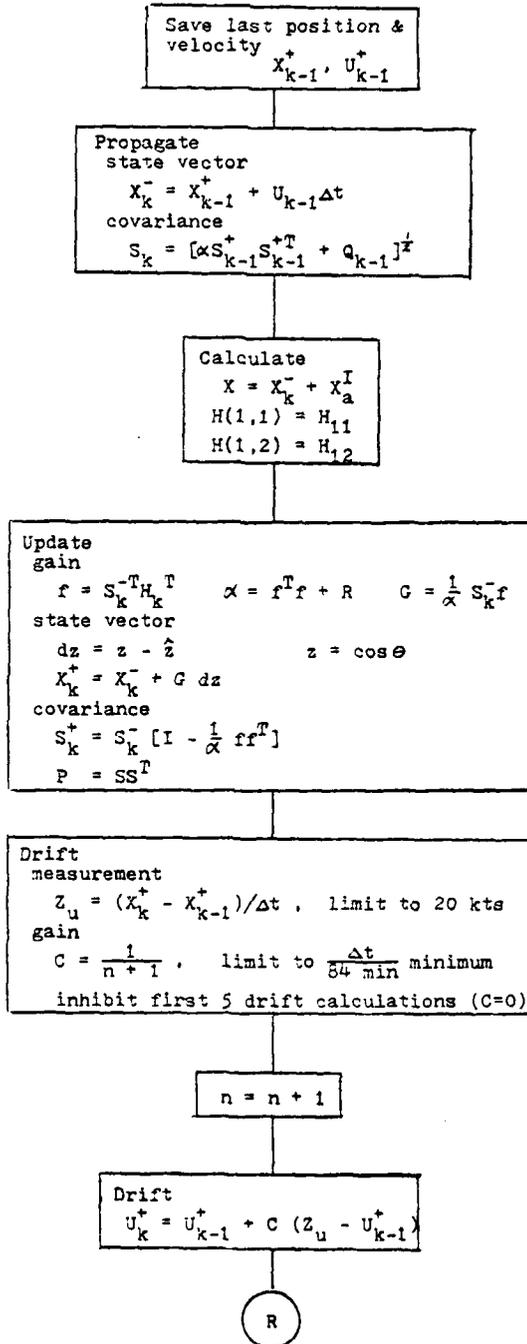


Figure 36. Flow chart: Subroutine FILTER two-state

GLOSSARY OF COMPUTER VARIABLES

SIMULATION PROGRAM

A	Vector of aircraft accelerations
ALPHA1	Result of $\frac{1}{f^T f + R}$ or $\frac{1}{\alpha}$
ALT	Altitude
ANGLE	Direction of sonobuoy drift from North
ANT	Antenna
AVEL	Initial aircraft velocity
AS	Matrix result of $I - \frac{ff^T}{\alpha}$
B	Matrix result of $\{eSS^T + Q\}^{\frac{1}{2}}$
BOUY	Sonobuoy RF channel or number
C	Gain of drift filter
CA	Result of $\Delta t / (84 \text{ min.})$
CA	Constant matrix in six-state system's equations
CKL	Rate of change of nautical miles for a change in degrees of longitude
D	Array which holds all items requiring storage from one measurement to the next
DELT	Interval between measurements, Δt
DELX1	Relative position, Δx_1
DELX2	Relative position, Δx_2
DELX3	Relative position, Δx_3
DRIFT	Magnitude of sonobuoy drift in yards
DSEED	Seed for the random number generator
DTIME	Array of initial drop times for the sonobuoy
DZ	Residual, dz
EDELX1	Error in the est. relative position of sonobuoy, Δx_1

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NAVAL POSTGRADUATE SCHOOL MONTEREY CA
ESTIMATION OF SONOBUOY POSITION RELATIVE TO AN AIRCRAFT USING E--ETC(U)
SEP 79 N M BROWNSBERGER

F/8 17/7

UNCLASSIFIED

NL

2 1/2 2
AD
2000000

END
DATE
FILMED
1 - 80
DOR

EDELX2 Error in the est. relative position of sonobuoy, ΔX_2
 EOSSO Matrix result of $\epsilon \hat{D}^T S^T S \hat{D}$
 EPSLN Age weighting factor,
 ESST Matrix result of $\epsilon S S^T$
 EX1 Error in est. geographical position of sonobuoy, ΔX_1
 EX2 Error in est. geographical position of sonobuoy, ΔX_2
 F f vector from Carlson square root technique
 FFT Matrix result of ff^T
 FFTA Matrix result of $\frac{ff^T}{\alpha}$
 FT Transpose of f
 G Gain matrix
 GAMMA Aircraft pitch angle, γ
 GDZ Vector result of $G dz$
 GR Gamma in radians
 H Aircraft heading
 H The H matrix when used in subroutine FILTER
 HOUR Time in hours
 HR Heading in radians
 HT Transpose of the H matrix
 I Index
 I Identity matrix when used in subroutine FILTER
 IEND Indicator of EOF
 IFFTA Matrix result of $I - \frac{ff^T}{\alpha}$
 IN File from which data is read
 INRES File from which true sonobuoy positions are read
 IOUT Device on which print is made
 IOUPT File in which results are written

IOUTZ File in which random vectors are inspected
 IRUN Index for run number
 IZ Index for measurement, $z = 1,2,3$
 K Incremental measure of time
 LAT Latitude
 LATI The latitude of the origin of the local earth fixed coordinate system
 LONG Longitude
 LONGI The longitude of the origin of the local earth fixed coordinate system
 MARK Indicates whether information is to be stored or retrieved
 MEAN1 Mean estimated sonobuoy position, μ_1
 MEAN2 Mean estimated sonobuoy position, μ_2
 MIN Time in minutes
 N Total number of measurements made on all sonobuoys
 N Index when used in subroutines INITAL and DATA
 NRUNS Integer equivalent of RUNS
 O Plant matrix, Φ
 OS Matrix result of ΦS
 OST Transpose of ΦS^T
 OSSO Matrix result of $\Phi^T S^T S \Phi$
 OX Matrix result of ΦX
 P Covariance matrix
 PHI Aircraft roll angle, ϕ
 PI π
 PR Phi in radians
 Q Variance of plant noise from $N(0, Q)$
 R Variance of measurement noise from $N(0, R)$

RAD $\pi / 180$
 RANGE Range to sonobuoy from aircraft
 RANGE2 Squared value of range
 RB1 Unit vector component of antenna baseline
 RB2 Unit vector component of antenna baseline
 RB3 Unit vector component of antenna baseline
 RBX Result of $(RB_1 \Delta x_1 + RB_2 \Delta x_2 + RB_3 \Delta x_3)$
 RNO Normally distributed random number from $N(0,1)$
 RUNS Number of simulation runs to be made
 S Array in Carlson's square root technique which represents square root of covariance matrix
 S1 Square root of variance, $\sqrt{P_{11}}$
 S2 Square root of variance, $\sqrt{P_{22}}$
 SEC Time in seconds
 SF Matrix result of S f
 SIGMA RMS error in the estimate about the actual location
 SIGMAF RMS value from the Kalman filter, $\sqrt{P_{11} + P_{22}}$
 SIGMAX RMS error in the estimate about the mean estimate
 SKL Rate of change of nautical miles for a change in degrees of latitude
 SNEXT Propagated matrix of S
 SST Matrix result of $S S^T$
 ST Transpose of S matrix
 SUMP1 Array for the sum of $P(1,1)$ over NRUNS
 SUMP2 Array for the sum of $P(2,2)$ over NRUNS
 SUMSQ1 Array for sum of $\tilde{\Delta x}_1^2$ over NRUNS
 SUMSQ2 Array for sum of $\tilde{\Delta x}_2^2$ over NRUNS
 SUMX1 Array for sum of $\tilde{\Delta x}_1$ over NRUNS

SUMX2 Array for sum of Δx_2 over NkUNS

T Last time in seconds the sonobuoy was processed by the filter

TIME Time from aircraft clock in seconds

U Sonobuoy drift vector

U1 Old value of U(1)

U2 Old value of U(2)

VAR1 Variance of the estimated sonobuoy position about the actual location σ_f^2

VAR2 Variance of the estimated sonobuoy position about the actual location σ_f^2

VARX1 Variance of the estimated sonobuoy position about the mean estimated position, σ_{x1}^2

VARX2 Variance of the estimated sonobuoy position about the mean estimated position, σ_{x2}^2

VELCT Relative velocity, ΔV

VELCT2 Squared value of relative veclocity

X State vector

XAI Aircraft inertial position, X_a^I

XAT Aircraft true position, X_a

X1 Old value of X(1)

X2 Old value of X(2)

XF Vector including XB1, XB2, velocity of sonobuoy, and two zeros used in the six-state system

XB1 True geographical location of sonobuoy, x_1

XB2 True geographical location of sonobuoy, x_2

XB1I Array of initial sonobuoy positions

XB2I Array of initial sonobuoy positions

XNEXT Propagated value of X

Z Measurement vector

ZHAT Estimate of z
ZM Normally distributed random number from $N(z,R)$
ZU Vector of velocity measurement for drift filter

PAY01000
 PAY01001
 PAY01002
 PAY01003
 PAY01004
 PAY01005
 PAY01006
 PAY01007
 PAY01008
 PAY01009
 PAY01010
 PAY01011
 PAY01012
 PAY01013
 PAY01014
 PAY01015
 PAY01016
 PAY01017
 PAY01018
 PAY01019
 PAY01020
 PAY01021
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 PAY01028
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 PAY01030
 PAY01031
 PAY01032
 PAY01033
 PAY01034
 PAY01035
 PAY01036
 PAY01037
 PAY01038
 PAY01039
 PAY01040
 PAY01041
 PAY01042
 PAY01043
 PAY01044
 PAY01045
 PAY01046
 PAY01047
 PAY01048
 PAY01049
 PAY01050

DIMENSION C(100), X(100), Y(100), Z(100), XH(100), YH(100), ZH(100)
 DATA C(100) / 1.0, 2.0, 3.0, 4.0, 5.0, 6.0, 7.0, 8.0, 9.0, 10.0 /
 DATA X(100) / 1.0, 2.0, 3.0, 4.0, 5.0, 6.0, 7.0, 8.0, 9.0, 10.0 /
 DATA Y(100) / 1.0, 2.0, 3.0, 4.0, 5.0, 6.0, 7.0, 8.0, 9.0, 10.0 /
 DATA Z(100) / 1.0, 2.0, 3.0, 4.0, 5.0, 6.0, 7.0, 8.0, 9.0, 10.0 /
 DATA XH(100) / 1.0, 2.0, 3.0, 4.0, 5.0, 6.0, 7.0, 8.0, 9.0, 10.0 /
 DATA YH(100) / 1.0, 2.0, 3.0, 4.0, 5.0, 6.0, 7.0, 8.0, 9.0, 10.0 /
 DATA ZH(100) / 1.0, 2.0, 3.0, 4.0, 5.0, 6.0, 7.0, 8.0, 9.0, 10.0 /

DO 10 I=1,100

10 CONTINUE
 X(I)=X(I)+C(I)
 Y(I)=Y(I)+C(I)
 Z(I)=Z(I)+C(I)
 XH(I)=XH(I)+C(I)
 YH(I)=YH(I)+C(I)
 ZH(I)=ZH(I)+C(I)
 X(I)=X(I)+C(I)
 Y(I)=Y(I)+C(I)
 Z(I)=Z(I)+C(I)
 XH(I)=XH(I)+C(I)
 YH(I)=YH(I)+C(I)
 ZH(I)=ZH(I)+C(I)

10 CONTINUE
 X(I)=X(I)+C(I)
 Y(I)=Y(I)+C(I)
 Z(I)=Z(I)+C(I)
 XH(I)=XH(I)+C(I)
 YH(I)=YH(I)+C(I)
 ZH(I)=ZH(I)+C(I)

PAT 03150
 PAT 03160
 PAT 03170
 PAT 03180
 PAT 03190
 PAT 03200
 PAT 03210
 PAT 03220
 PAT 03230
 PAT 03240
 PAT 03250
 PAT 03260
 PAT 03270
 PAT 03280
 PAT 03290
 PAT 03300
 PAT 03310
 PAT 03320
 PAT 03330
 PAT 03340
 PAT 03350
 PAT 03360
 PAT 03370
 PAT 03380
 PAT 03390
 PAT 03400
 PAT 03410
 PAT 03420
 PAT 03430
 PAT 03440
 PAT 03450
 PAT 03460
 PAT 03470
 PAT 03480
 PAT 03490
 PAT 03500

```

    C(1,1)=0.0
    C(2,1)=0.2
    C(3,1)=0.5
    C(4,1)=0.8
    C(5,1)=0.0
    C(6,1)=0.0
    C(7,1)=0.8
    C(8,1)=1.0
    C(9,1)=1.0

    PIVOT=KATIK

    C(1,1)=C(1,1)
    C(1,2)=SIN(NT)/X
    C(1,3)=COS(NT)/X
    C(1,4)=SIN(NT)/X
    C(1,5)=COS(NT)/X
    C(1,6)=SIN(NT)/X
    C(1,7)=COS(NT)/X
    C(1,8)=SIN(NT)/X
    C(1,9)=COS(NT)/X
    C(2,1)=C(2,1)
    C(2,2)=SIN(NT)/X
    C(2,3)=COS(NT)/X
    C(2,4)=SIN(NT)/X
    C(2,5)=COS(NT)/X
    C(2,6)=SIN(NT)/X
    C(2,7)=COS(NT)/X
    C(2,8)=SIN(NT)/X
    C(2,9)=COS(NT)/X
    C(3,1)=C(3,1)
    C(3,2)=SIN(NT)/X
    C(3,3)=COS(NT)/X
    C(3,4)=SIN(NT)/X
    C(3,5)=COS(NT)/X
    C(3,6)=SIN(NT)/X
    C(3,7)=COS(NT)/X
    C(3,8)=SIN(NT)/X
    C(3,9)=COS(NT)/X
    C(4,1)=C(4,1)
    C(4,2)=SIN(NT)/X
    C(4,3)=COS(NT)/X
    C(4,4)=SIN(NT)/X
    C(4,5)=COS(NT)/X
    C(4,6)=SIN(NT)/X
    C(4,7)=COS(NT)/X
    C(4,8)=SIN(NT)/X
    C(4,9)=COS(NT)/X
    C(5,1)=C(5,1)
    C(5,2)=SIN(NT)/X
    C(5,3)=COS(NT)/X
    C(5,4)=SIN(NT)/X
    C(5,5)=COS(NT)/X
    C(5,6)=SIN(NT)/X
    C(5,7)=COS(NT)/X
    C(5,8)=SIN(NT)/X
    C(5,9)=COS(NT)/X
    C(6,1)=C(6,1)
    C(6,2)=SIN(NT)/X
    C(6,3)=COS(NT)/X
    C(6,4)=SIN(NT)/X
    C(6,5)=COS(NT)/X
    C(6,6)=SIN(NT)/X
    C(6,7)=COS(NT)/X
    C(6,8)=SIN(NT)/X
    C(6,9)=COS(NT)/X
    C(7,1)=C(7,1)
    C(7,2)=SIN(NT)/X
    C(7,3)=COS(NT)/X
    C(7,4)=SIN(NT)/X
    C(7,5)=COS(NT)/X
    C(7,6)=SIN(NT)/X
    C(7,7)=COS(NT)/X
    C(7,8)=SIN(NT)/X
    C(7,9)=COS(NT)/X
    C(8,1)=C(8,1)
    C(8,2)=SIN(NT)/X
    C(8,3)=COS(NT)/X
    C(8,4)=SIN(NT)/X
    C(8,5)=COS(NT)/X
    C(8,6)=SIN(NT)/X
    C(8,7)=COS(NT)/X
    C(8,8)=SIN(NT)/X
    C(8,9)=COS(NT)/X
    C(9,1)=C(9,1)
    C(9,2)=SIN(NT)/X
    C(9,3)=COS(NT)/X
    C(9,4)=SIN(NT)/X
    C(9,5)=COS(NT)/X
    C(9,6)=SIN(NT)/X
    C(9,7)=COS(NT)/X
    C(9,8)=SIN(NT)/X
    C(9,9)=COS(NT)/X
  
```


1. TABLE (10,30,20,5) (10,30,20,5) (10,30,20,5) (10,30,20,5)
 2. TABLE (10,30,20,5) (10,30,20,5) (10,30,20,5) (10,30,20,5)
 3. TABLE (10,30,20,5) (10,30,20,5) (10,30,20,5) (10,30,20,5)
 4. TABLE (10,30,20,5) (10,30,20,5) (10,30,20,5) (10,30,20,5)
 5. TABLE (10,30,20,5) (10,30,20,5) (10,30,20,5) (10,30,20,5)

PAT 34900
 PAT 34901
 PAT 34902
 PAT 34903
 PAT 34904
 PAT 34905
 PAT 34906
 PAT 34907
 PAT 34908
 PAT 34909
 PAT 34910
 PAT 34911
 PAT 34912
 PAT 34913
 PAT 34914
 PAT 34915
 PAT 34916
 PAT 34917
 PAT 34918
 PAT 34919
 PAT 34920

SUBROUTINE TABLE (PSEUDO, Z, STP)
 DOUBLE PRECISION PSEUDO
 LOG Z
 PSEUDO = PSEUDO * STP
 Z = Z * STP
 WRITE (10,20) PSEUDO, Z, STP
 20. FORMAT (3F10.8)
 PRINT
 END

APPENDIX B
COMPUTER PROGRAM FOR DATA GENERATION

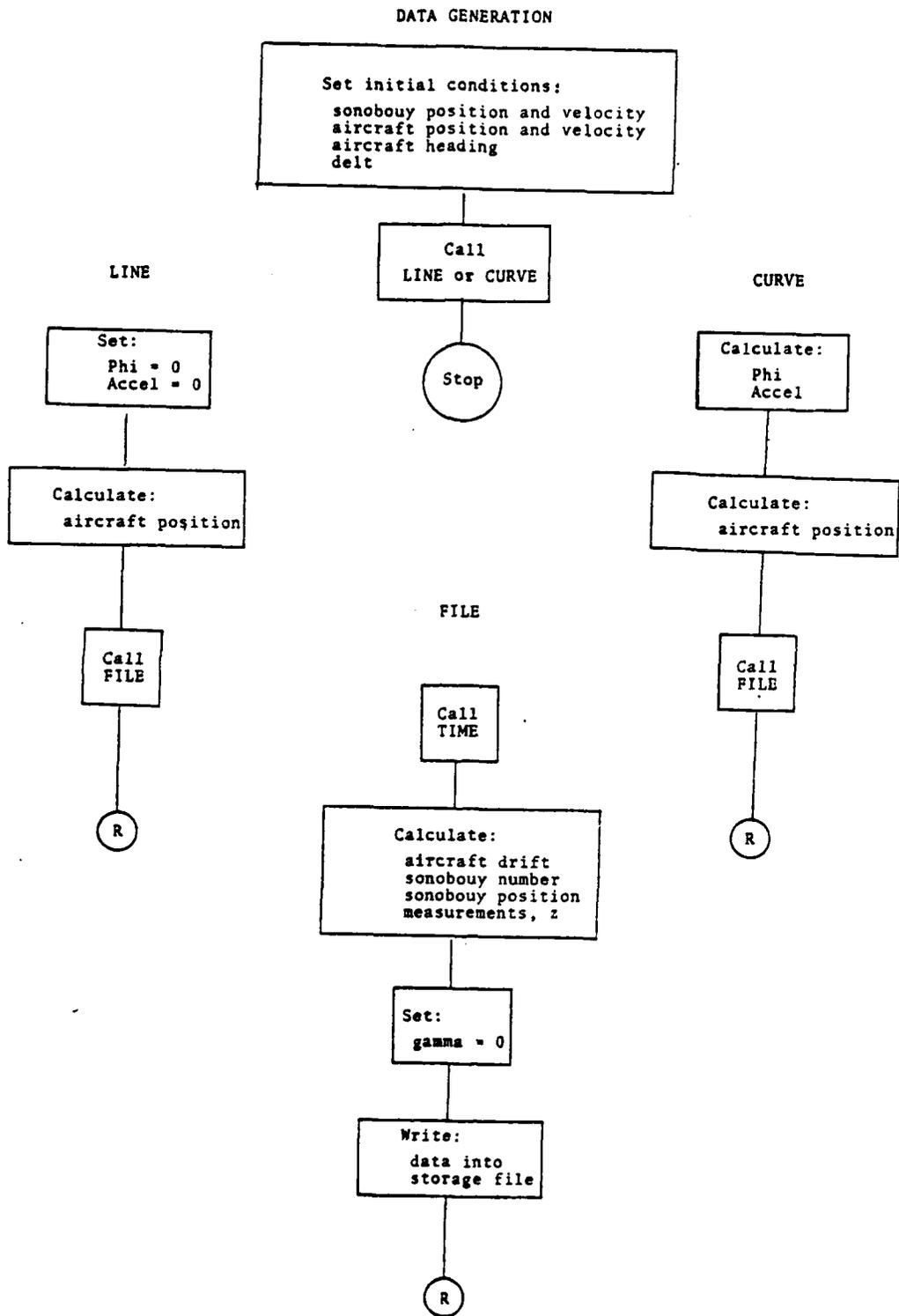


Figure 37. Flow chart: data generation program.

GLOSSARY OF COMPUTER VARIABLES
DATA GENERATION PROGRAM

A	Vector of aircraft accelerations
AL	Altitude in nautical miles
ALT	Altitude
ANT	Antenna
AVEL	Initial aircraft velocity
AX1	Amplitude of Schuler cycle, A_{x1}
AX2	Amplitude of Schuler cycle, A_{x2}
B	Intercept on N-S axis of straight track
BANG	Sonobuoy drift direction
BOUY	Sonobuoy RF channel or number
BVEL	Sonobuoy drift direction
CKL	Rate of change of nautical miles for a change in degrees of longitude
D	Length of straight track
DELT	Interval between measurements, Δt
DELV1	Relative velocity, Δv_1
DELV2	Relative velocity, Δv_2
DELX1	Relative position, Δx_1
DELX2	Relative position, Δx_2
DH	Total change in heading while on curved track
DSEED	Seed for the random number generator
END	Indicator of EOF
G	Gravitational acceleration
GAMMA	Aircraft pitch angle, γ

GR	Gamma in radians
H	Aircraft heading
HNEW	Final heading on curved track
HOUR	Time in hours
HR	Heading in radians
IOUT	Device on which print is made
IOUFP	File in which results are written
IOUFR	File in which true sonobuoy positions are written
IOUTZ	File in which random vectors are inspected
ITIME	Time from aircraft clock in seconds
KX1	Rate of linear navigational drift, K_{x1}
KX2	Rate of linear navigational drift, K_{x2}
LAT	Latitude
LATI	The latitude of the origin of the local earth fixed coordinate system
LONG	Longitude
LONGI	The longitude of the origin of the local earth fixed coordinate system
LTIME	Array of initial drop times for the sonobuoy
M	Slope of straight track
MARK	Indicates the first time a buoy is to be processed
MIN	Time in minutes
N	Index
NBOUY	Total number of sonobuoys
NSTOP	End of line or curve series
PHI	Aircraft roll angle, ϕ
PI	π
PI2	2π

PR	Phi in radians
R	Radius of curved track
RB1	Unit vector component of antenna baseline
RB2	Unit vector component of antenna baseline
RB3	Unit vector component of antenna baseline
S	Distance aircraft travels in time Δt
SEC	Time in seconds
SKL	Rate of change of nautical miles for a change in degrees of latitude
TRACK	Direction of aircraft travel over the ground
U	Direction to center of curved track
V	Direction from center of curved track to aircraft
VA1	Velocity of aircraft
VA2	Velocity of aircraft
VB1	Sonobuoy drift
VB2	Sonobuoy drift
WANG	Wind direction
WVEL	Wind velocity
XAI	Aircraft inertial position, X_a^I
XAT	Aircraft true position, X_a
XB1	True geographical location of sonobuoy, x_1
XB2	True geographical location of sonobuoy, x_2
XC1	Location of the center of the curved track
XC2	Location of the center of the curved track
XX1	Incremental change in X_1
XX2	Incremental change in X_2
Z	Measurement vector

CR100520
 CR100530
 CR100540
 CR100550
 CR100560
 CR100570
 CR100580
 CR100590
 CR100600
 CR100610
 CR100620
 CR100630
 CR100640
 CR100650
 CR100660
 CR100670
 CR100680
 CR100690
 CR100700
 CR100710
 CR100720
 CR100730
 CR100740
 CR100750
 CR100760
 CR100770
 CR100780
 CR100790
 CR100800
 CR100810
 CR100820
 CR100830
 CR100840
 CR100850
 CR100860
 CR100870
 CR100880
 CR100890
 CR100900
 CR100910
 CR100920
 CR100930
 CR100940
 CR100950
 CR100960
 CR100970
 CR100980
 CR100990

5 CALL CUFVEL(XAT,3,5,H,00C,,S,MARK)
 6 CALL LINF(XAT,H,21,,S,MARK)
 7 CALL CUFVEL(XAT,3,5,H,27C,,S,MARK)
 8 CALL LINF(XAT,H,21,,S,MARK)
 9 CALL CUFVEL(XAT,3,5,H,18C,,S,MARK)
 10 CALL LINF(XAT,H,22,,S,MARK)
 11 CALL CUFVEL(XAT,3,5,H,09C,,S,MARK)
 12 CALL LINF(XAT,H,22,,S,MARK)
 13 CALL CUFVEL(XAT,3,5,H,0,,S,MARK)
 14 CALL LINF(XAT,H,23,,S,MARK)
 15 CALL CUFVEL(XAT,3,5,H,27C,,S,MARK)
 16 CALL LINF(XAT,H,23,,S,MARK)
 17 CALL CUFVEL(XAT,3,5,H,18C,,S,MARK)
 18 CALL LINF(XAT,H,24,,S,MARK)
 19 CALL CUFVEL(XAT,3,5,H,09C,,S,MARK)
 20 CALL LINF(XAT,H,24,,S,MARK)
 21 CALL CUFVEL(XAT,3,5,H,00C,,S,MARK)
 22 CALL LINF(XAT,H,25,,S,MARK)
 23 CALL CUFVEL(XAT,3,5,H,27C,,S,MARK)
 24 CALL LINF(XAT,H,25,,S,MARK)
 25 CALL CUFVEL(XAT,3,5,H,18C,,S,MARK)
 26 CALL LINF(XAT,H,26,,S,MARK)
 27 CALL CUFVEL(XAT,3,5,H,09C,,S,MARK)
 28 CALL LINF(XAT,H,26,,S,MARK)

```

00101000
00101010
00101020
00101030
00101040
00101050
00101060
00101070
00101080
00101090
00101100
00101110
00101120
00101130
00101140
00101150
00101160
00101170
00101180
00101190
00101200
00101210
00101220
00101230
00101240
00101250
00101260
00101270
00101280
00101290
00101300
00101310
00101320
00101330
00101340
00101350
00101360
00101370
00101380
00101390
00101400

```

```

9000 UNDE=99999.9
9001 UNDE=95
WRITE(CUCL,9001) UNDE
WRITE(CUCLP,9000) UNDE
9002 FORMAT(I10,F10.4)
9003 FORMAT(I2)

STOP
END

SUBROUTINE LIT (XAT,F,D,S,MARK)
DIMENSION XAT(2),A(2)
REAL M
NSTICP=FIX(D/5)
PHI=C.
A(1)=0.
A(2)=C.
IF(P.EC.D. J GO TO 20
IF(D.EC.180.J GO TO 40
HR=0+3.1415926/180.
NECCTAN(HR)
N=XAT(1)-M*XAT(2)
XAL=5+SIGN(F)

10 L=1,NSTICP
XAT(2)=XAT(2)+XAL
XAT(1)=M*XAT(2)+R
CALL FILE (XAT,D,PHI,A,MARK)
GO TO 55
20 N=30 F=1,HRCTOP
XAT(1)=XAT(1)+S
CALL FILE (XAT,D,PHI,A,MARK)
GO TO 55
30 D=50 N=1,NSTICP
XAT(1)=XAT(1)-S
CALL FILE (XAT,D,PHI,A,MARK)
GO TO 55
40 N=100
CALL FILE (XAT,D,PHI,A,MARK)
GO TO 55
50
55
END

```

```

SUBROUTINE CURVE1(XAT,P,H,HEF,S,MARK)
      DIMENSION XAT(2),P(2)
      DATA PI,PI2,CG /3.141592653,9.8696044,0.28183829/
      AVEL=1.0
      U=(H-S)*PI/180.
      XC1=COS(U)+XAT(1)
      XC2=1+SIGN(U)*XAT(2)
      V=(H+S)*PI/180.
      UC1=COS(V)-HEF
      UC2=1+SIGN(V)*HEF
      PSTOP=1+1X(UC1+PI/180.*P/S)
      PFI=-ATAN(AVEL*1.6878)*2/(CG*XC1*XC2*(115))*.180./PI
      GO TO 1,IF=1,IFSTOP
      V-V-S/S
      IF(V.GT.PI/2) V-V-PI
      IF(V.LT.C) V=V+PI/2
      R(1)=-COS(V)*AVEL*2/4
      R(2)=-SIGN(V)*AVEL*2/4
      HEV=180./PI-90.
      IF(HEV.LT.C) HEV=HEV+360.
      XC1(1)=R(1)*S(V)+XC1
      XC2=SIGN(R(2)*S(V)-XAT(1)-XC1(1)-2)
      IF(V.LT.PI) XC1(2)=XC2+XX2
      IF(V.GT.PI) XC1(2)=XC2-XX2
      CALL SILE(XAT,P,H,HEF,S,MARK)
      HEF=1.5
      HEF=H
      IF=1
      END

```

```

SUBROUTINE CURVE1(XAT,P,H,HEF,S,MARK)
      DIMENSION XAT(2),P(2)
      DATA PI,PI2,CG /3.141592653,9.8696044,0.28183829/
      AVEL=1.0
      U=(H-S)*PI/180.
      XC1=COS(U)+XAT(1)
      XC2=1+SIGN(U)*XAT(2)
      V=(H+S)*PI/180.
      UC1=COS(V)-HEF
      UC2=1+SIGN(V)*HEF
      PSTOP=1+1X(UC1+PI/180.*P/S)
      PFI=-ATAN(AVEL*1.6878)*2/(CG*XC1*XC2*(115))*.180./PI
      GO TO 1,IF=1,IFSTOP
      V-V-S/S
      IF(V.GT.PI/2) V-V-PI
      IF(V.LT.C) V=V+PI/2
      R(1)=-COS(V)*AVEL*2/4
      R(2)=-SIGN(V)*AVEL*2/4
      HEV=180./PI-90.
      IF(HEV.LT.C) HEV=HEV+360.
      XC1(1)=R(1)*S(V)+XC1
      XC2=SIGN(R(2)*S(V)-XAT(1)-XC1(1)-2)
      IF(V.LT.PI) XC1(2)=XC2+XX2
      IF(V.GT.PI) XC1(2)=XC2-XX2
      CALL SILE(XAT,P,H,HEF,S,MARK)
      HEF=1.5
      HEF=H
      IF=1
      END

```

```

SUBROUTINE CURVE1(XAT,P,H,HEF,S,MARK)
      DIMENSION XAT(2),P(2)
      DATA PI,PI2,CG /3.141592653,9.8696044,0.28183829/
      AVEL=1.0
      U=(H-S)*PI/180.
      XC1=COS(U)+XAT(1)
      XC2=1+SIGN(U)*XAT(2)
      V=(H+S)*PI/180.
      UC1=COS(V)-HEF
      UC2=1+SIGN(V)*HEF
      PSTOP=1+1X(UC1+PI/180.*P/S)
      PFI=-ATAN(AVEL*1.6878)*2/(CG*XC1*XC2*(115))*.180./PI
      GO TO 1,IF=1,IFSTOP
      V-V-S/S
      IF(V.GT.PI/2) V-V-PI
      IF(V.LT.C) V=V+PI/2
      R(1)=-COS(V)*AVEL*2/4
      R(2)=-SIGN(V)*AVEL*2/4
      HEV=180./PI-90.
      IF(HEV.LT.C) HEV=HEV+360.
      XC1(1)=R(1)*S(V)+XC1
      XC2=SIGN(R(2)*S(V)-XAT(1)-XC1(1)-2)
      IF(V.LT.PI) XC1(2)=XC2+XX2
      IF(V.GT.PI) XC1(2)=XC2-XX2
      CALL SILE(XAT,P,H,HEF,S,MARK)
      HEF=1.5
      HEF=H
      IF=1
      END

```



```

CPI000740
CPI000750
CPI000760
CPI000770
CPI000780
CPI000790
CPI001000
CPI001100
CPI001200
CPI001300
CPI001400
CPI001500
CPI001600
CPI001700
CPI001800
CPI001900
CPI002000
CPI002100
CPI002200
CPI002300
CPI002400
CPI002500
CPI002600
CPI002700
CPI002800
CPI002900
CPI003000
CPI003100
CPI003200
CPI003300
CPI003400
CPI003500
CPI003600
CPI003700
CPI003800
CPI003900
CPI004000
CPI004100
CPI004200
CPI004300
CPI004400
CPI004500
CPI004600
CPI004700
CPI004800
CPI004900
CPI005000
CPI005100

```

```

*****
3     DATA GENERATION
4     FCRA
5     SQUARE PATTERN
*****

```

```

C     IC ALPH, (FFLK)
C     IOUT=7
C     IOUTP=10
C     IOUTZ=4
C     XAT(1)=-3.
C     XAT(2)=-5.
C     AVFL=180.
C     F=186.
C     DELT=50.
C     ALT=3000.
C     GAMMA=C.
C     LEVELD=7654221.000
C     AX1,XX2,KX1,KX2 IN ACDFIF
C     WVFL=0.
C     WANGE=0.
C     RVILLEZ.
C     RANG=0.50.
C     CURV CALLS
C     LINE AND BUOYS-ADJUST '098' IN BUDYNU
C     FOR MULTIPLE BUOYS-ADJUST '098' IN BUDYNU
C     AND INITIAL BUOY POSITIONS IN BUDYPU
C     WARNING: H CANNOT BE GREATER THAN 300. **
C     EXTENSION XAT(?)
C     IOUT=7
C     IOUTP=10
C     IOUTZ=4
C     MARK=1
C     XAT(1)=-2.
C     XAT(2)=-5.
C     H=180.
C     AVFL=100.
C     DELT=50.
C     SEV=1.0E+17/3000.
1     CALL LIFE(XAT,H,7.5,C,MARK)
2     CALL CURVE(XAT,7.5,H,09C,7.5,MARK)
3     CALL LIFE(XAT,H,10,5,MARK)

```


APPENDIX C
Range Analysis

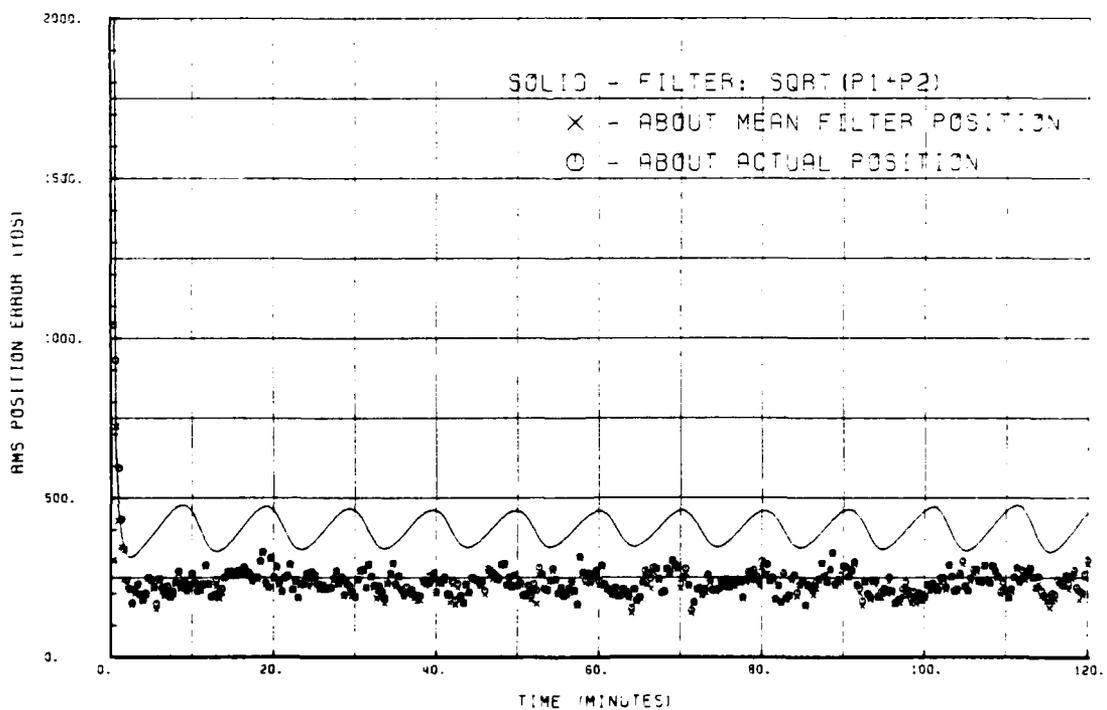
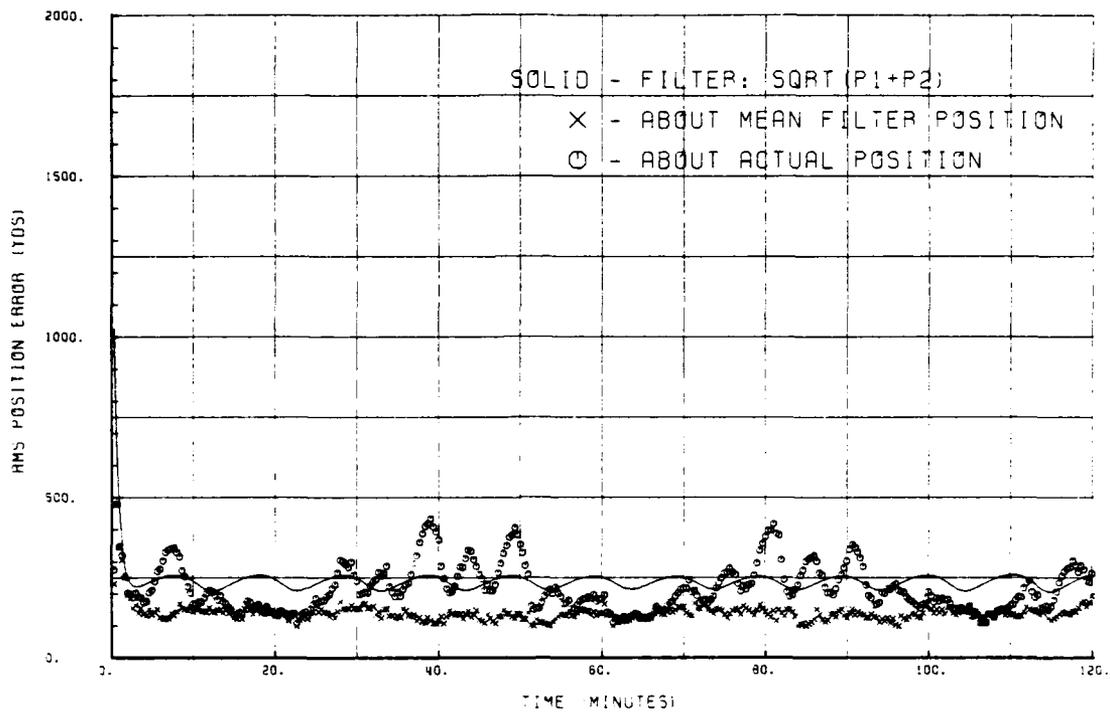


Figure 38. RMS errors for two-state system (top) and six-state system (bottom) using circular pattern at 5 NM.

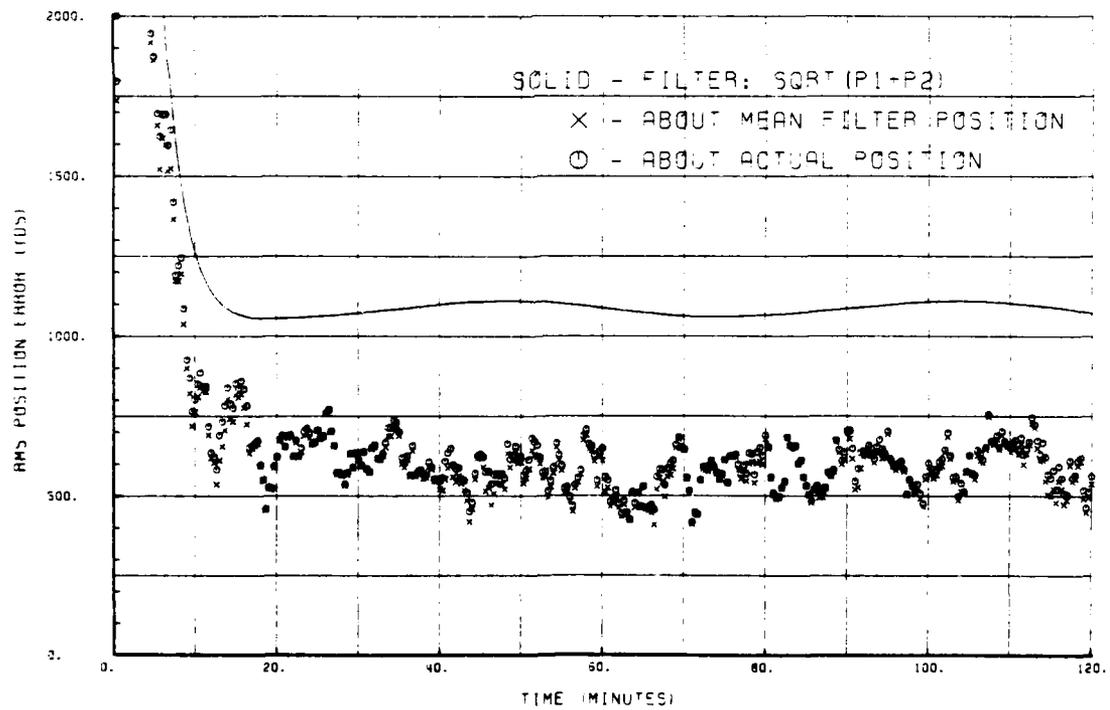
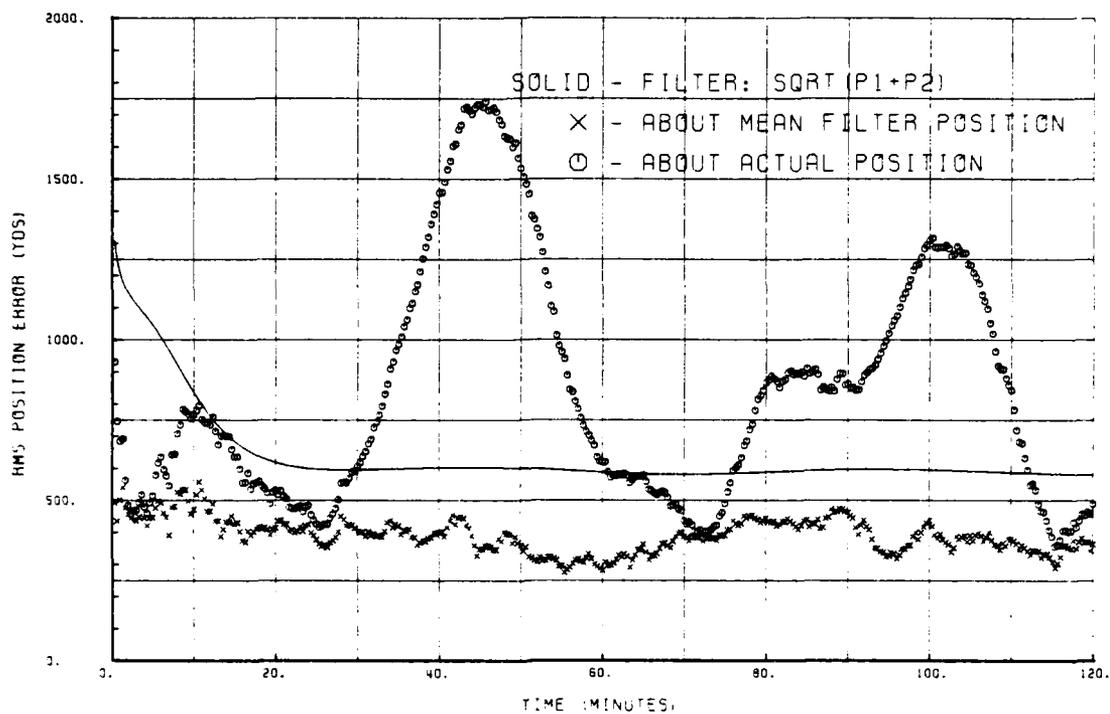


Figure 39. RMS errors for two-state system (top) and six-state system (bottom) using circular pattern at 30 NM.

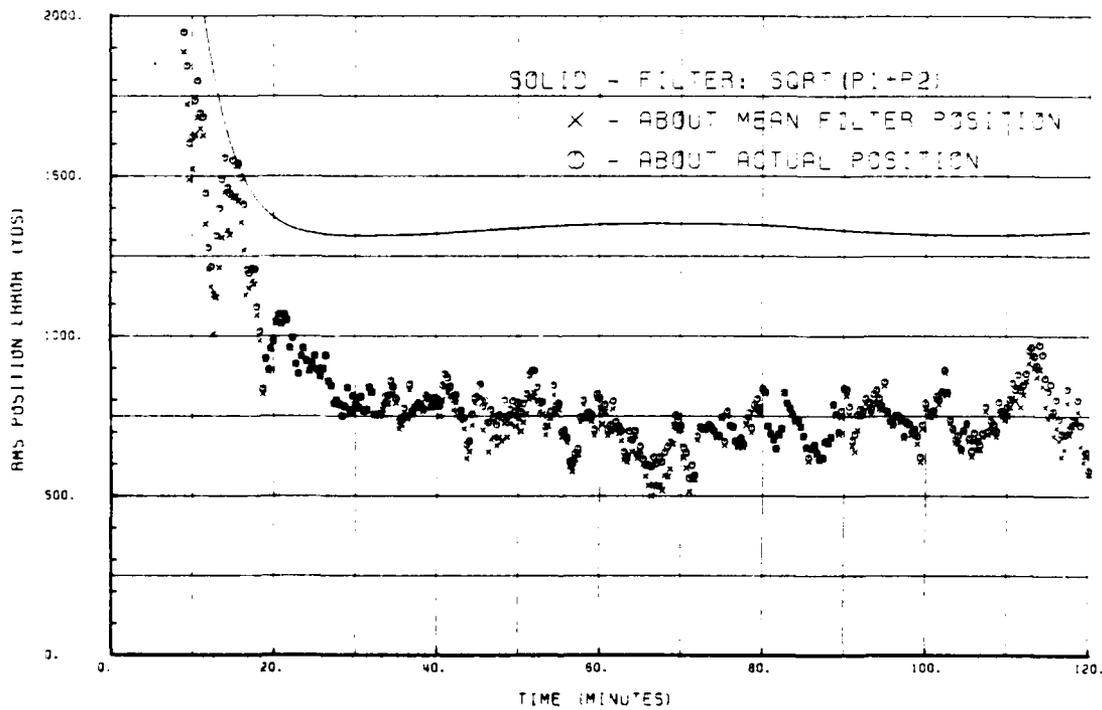
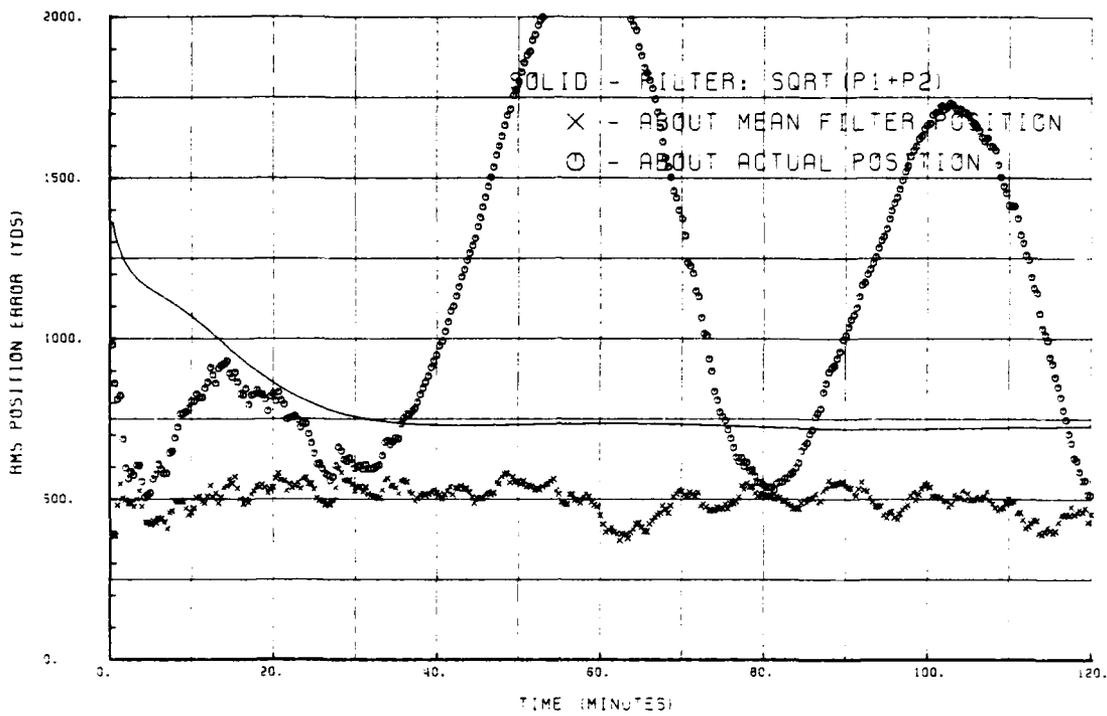


Figure 40. RMS errors for two-state system (top) and six-state system (bottom) using circular pattern at 45 NM.

Frequency Analysis

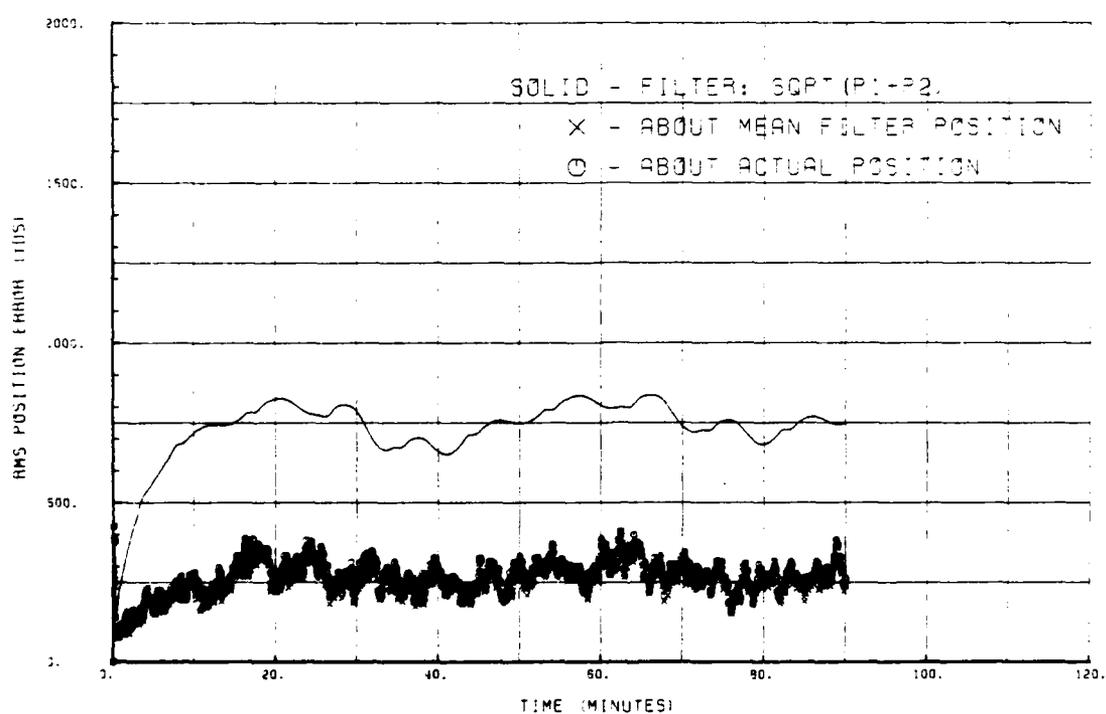
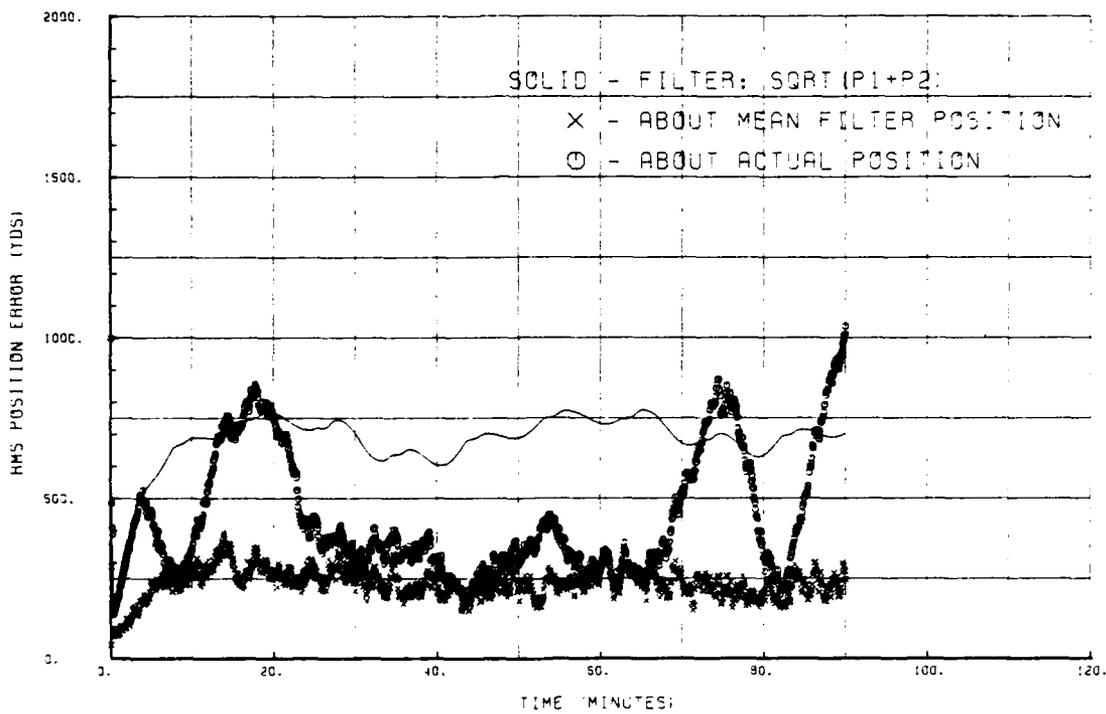


Figure 41. RMS errors for two-state system (top) and six-state system (bottom) using square pattern with $t=4$ sec.

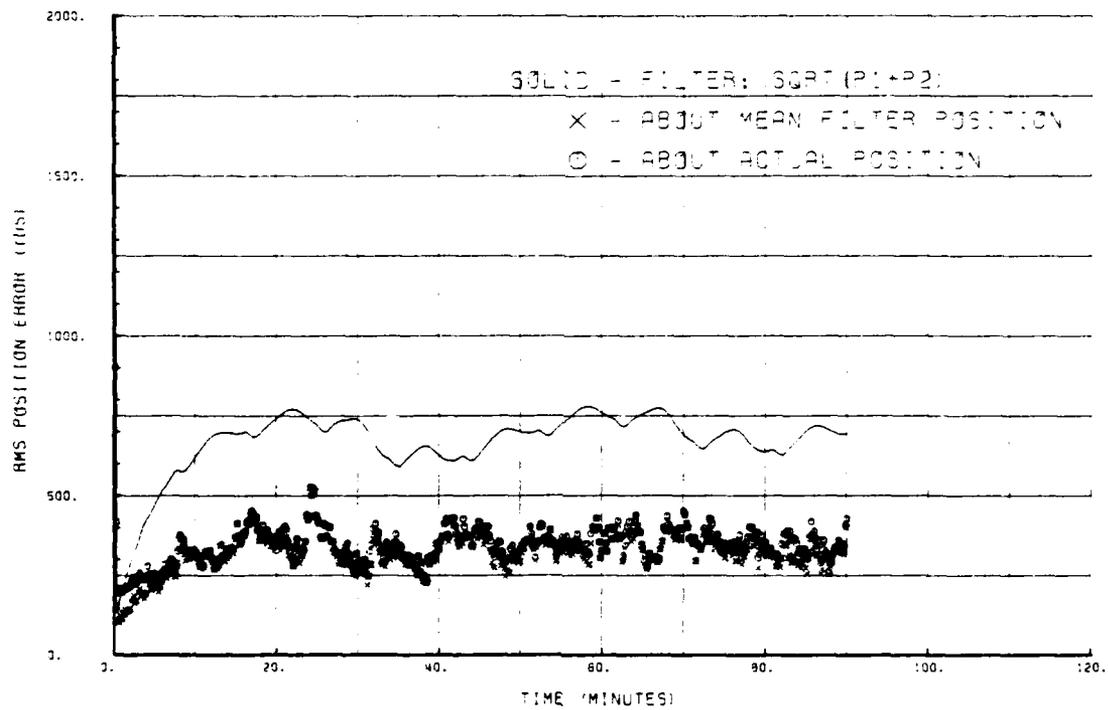
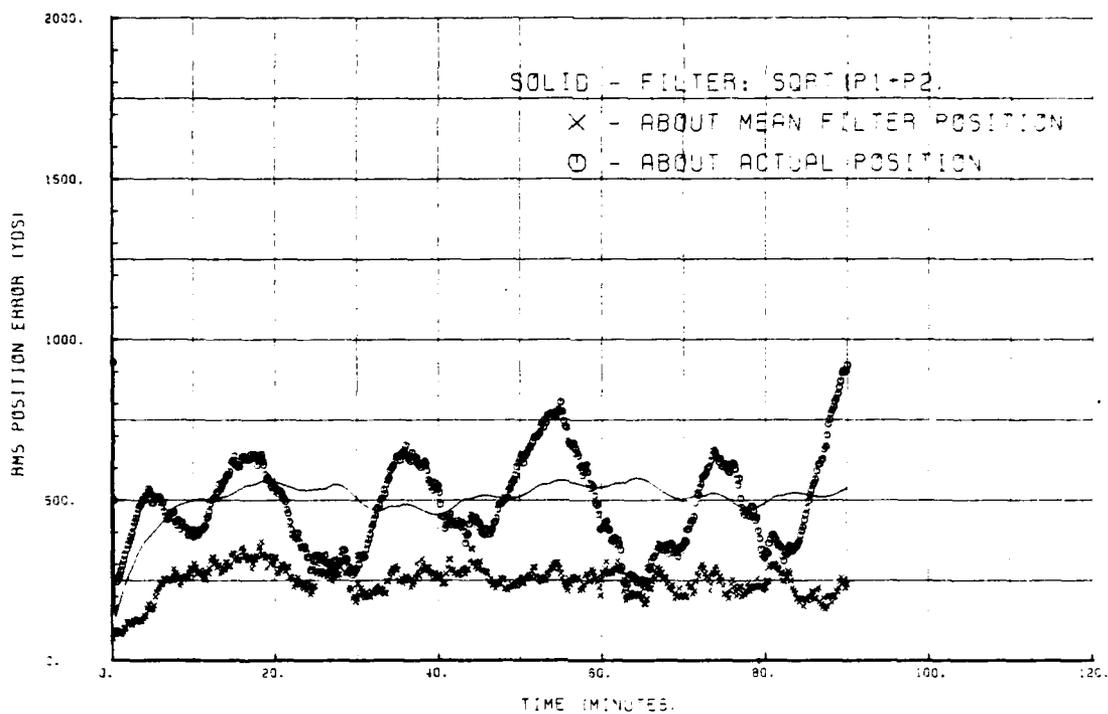


Figure 42. RMS errors for two-state system (top) and six-state system (bottom) using square pattern with $t=10$ sec.

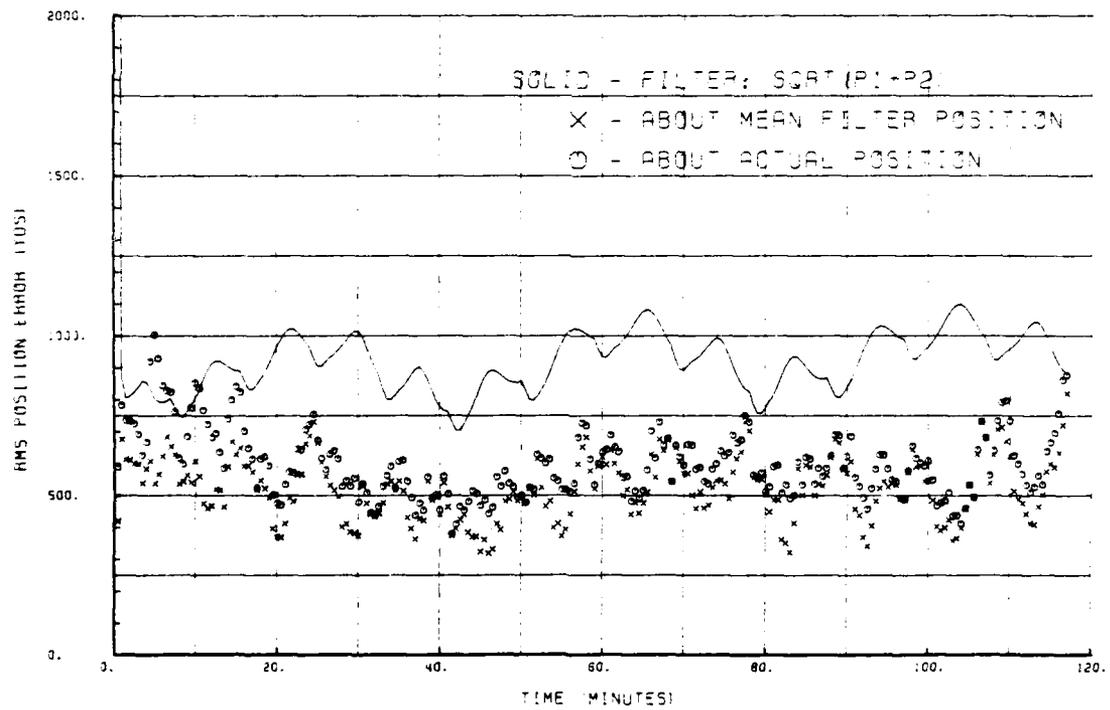
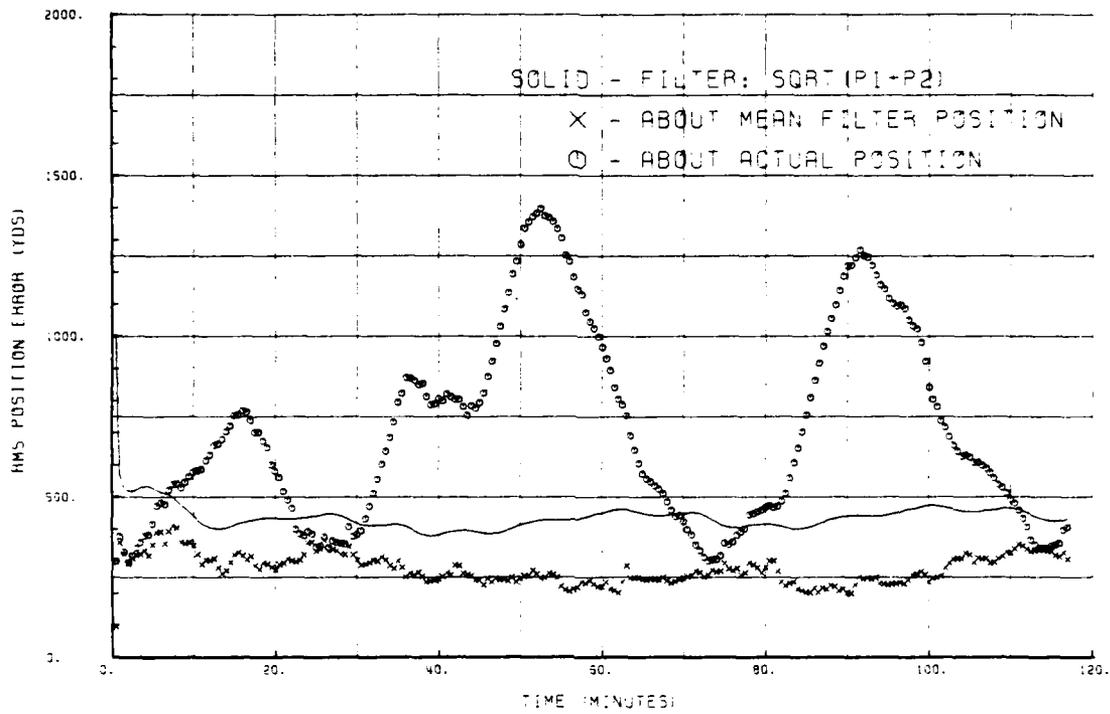


Figure 43. RMS errors for two-state system (top) and six-state system (bottom) using square pattern with $t=30$ sec.

Altitude Analysis

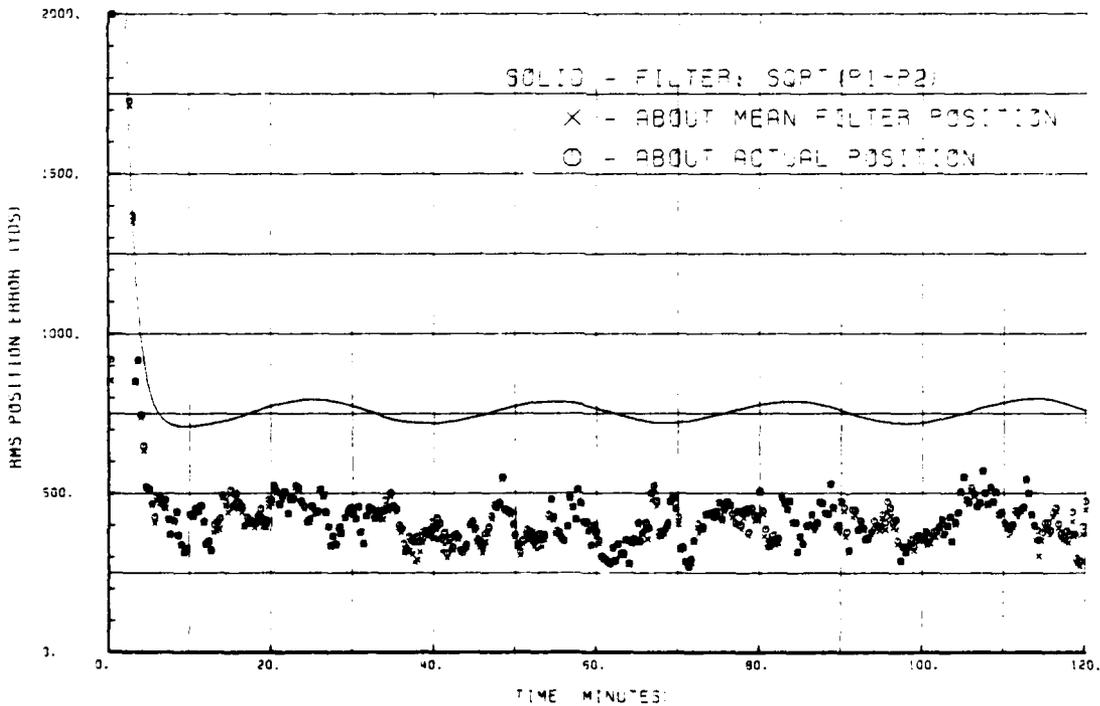
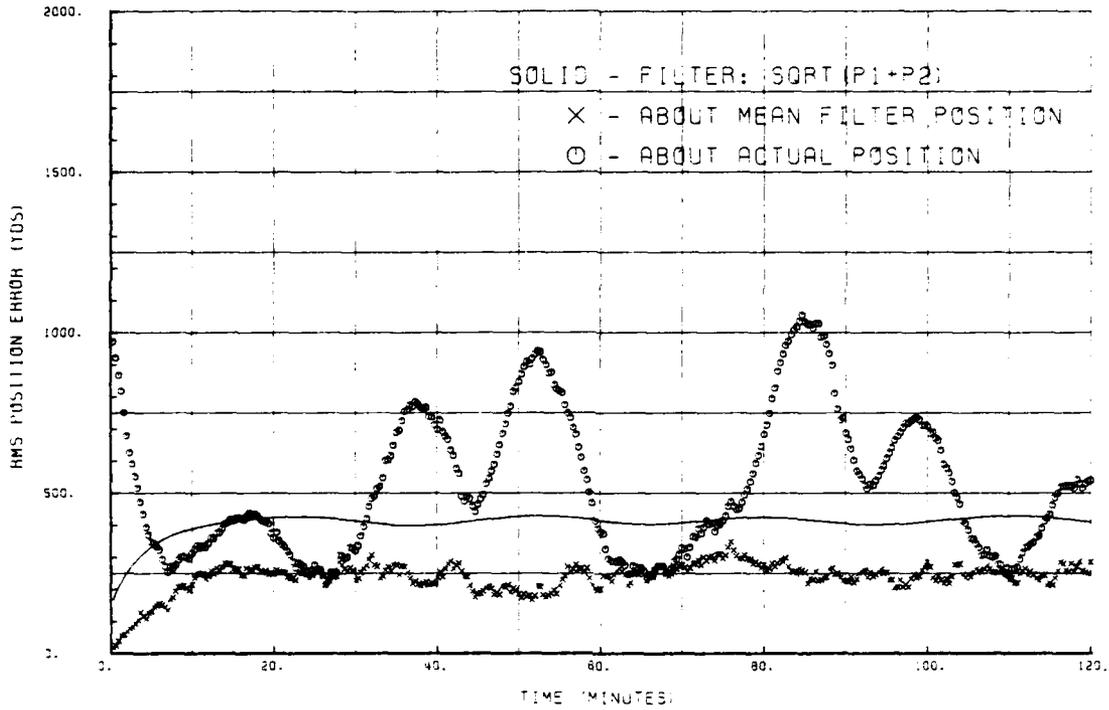


Figure 43. RMS errors for two-state system (top) and six-state system (bottom) using circular pattern, Alt=300'.

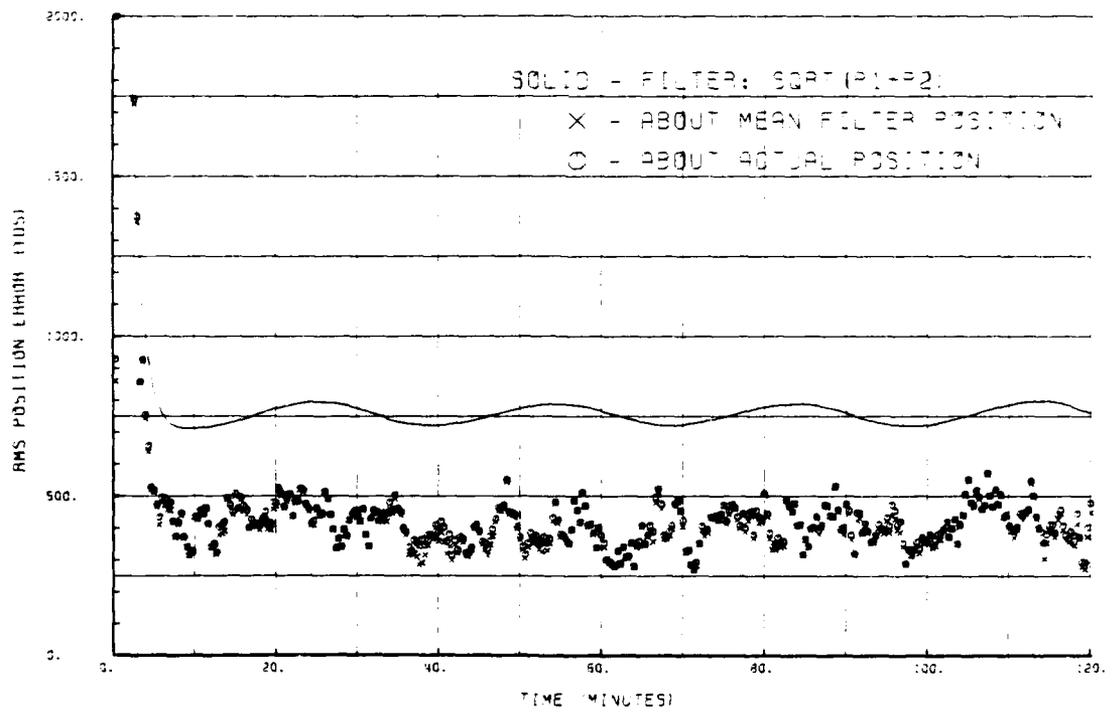
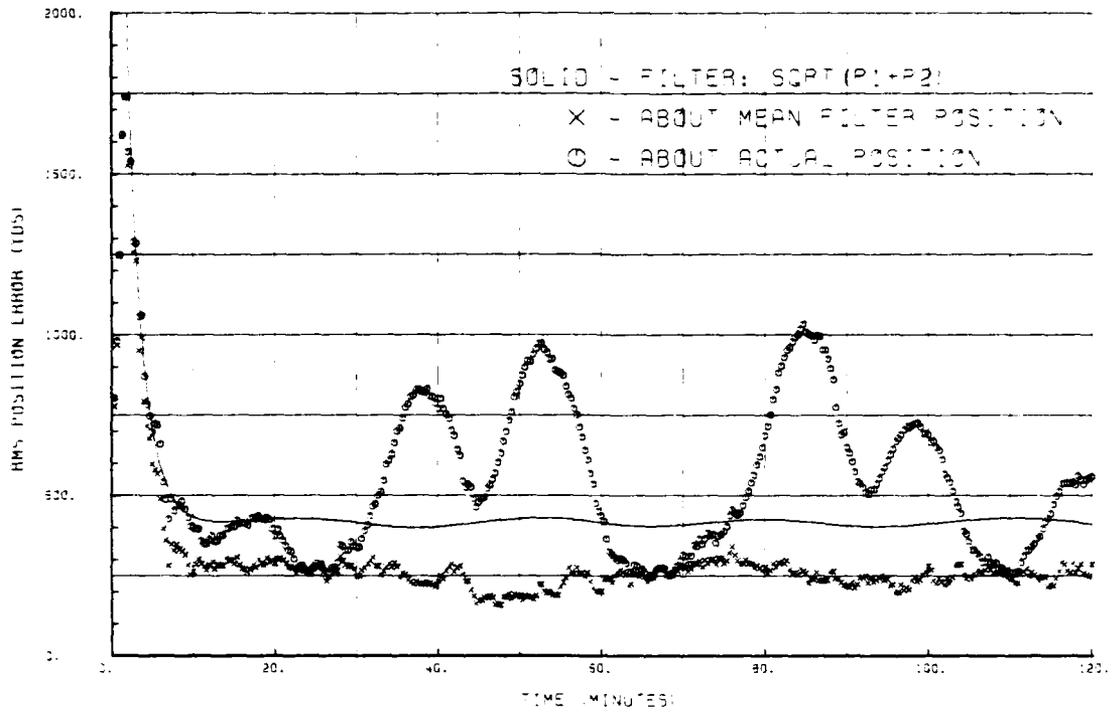


Figure 45. RMS errors for two-state system (top) and six-state system (bottom) using circular pattern, Alt=10,000'.

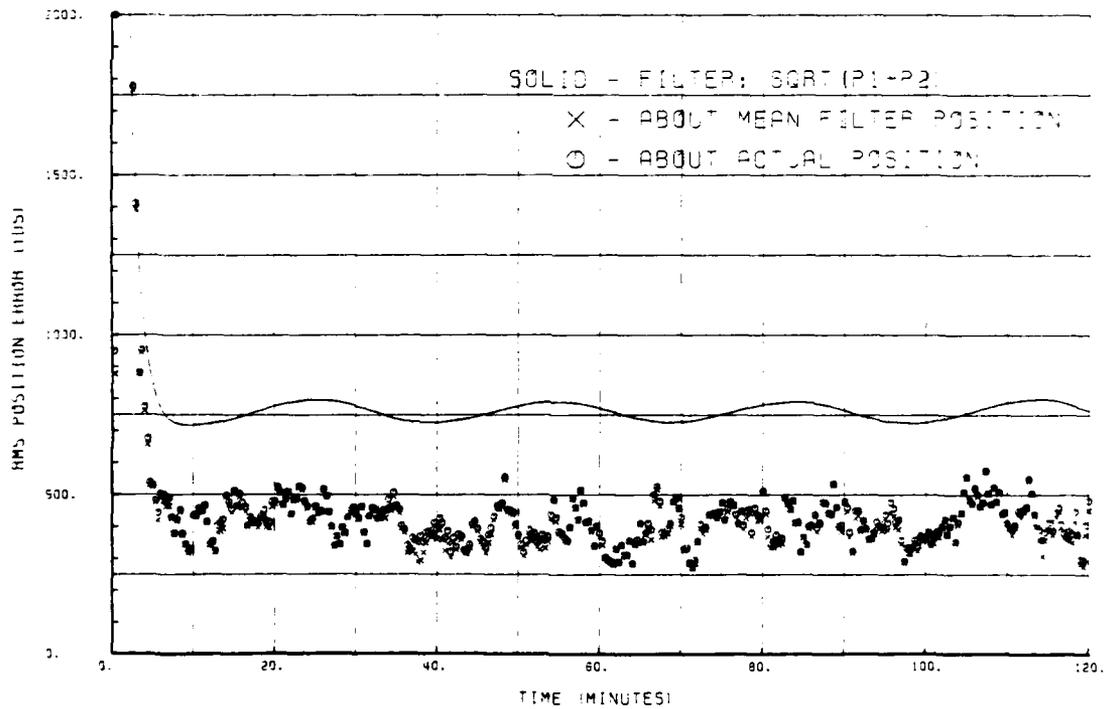
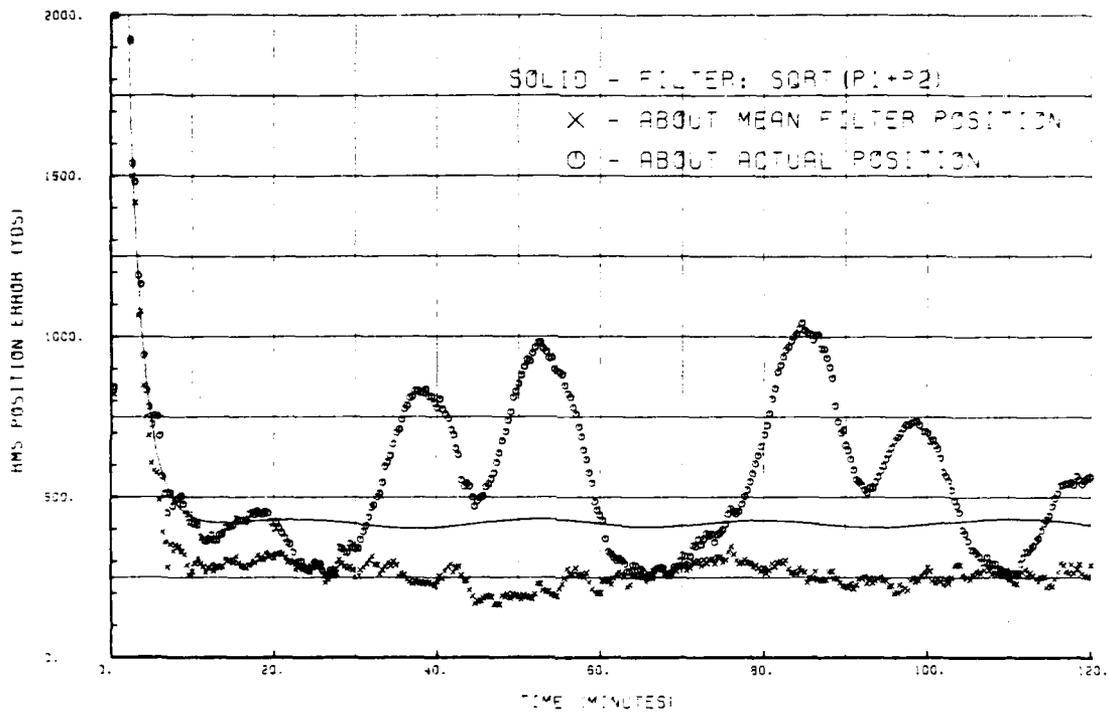


Figure 46. RMS errors for two-state system (top) and six-state system (bottom) using circular pattern, Alt=20,000'.

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